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	Vendor Issue Number	1203025				
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Product	Standards	
Part No.	AN30888A	
Package Code No.	SSOP016-P-0225E	

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# AN30888A High Brightness LED Driver IC

### Overview

AN30888A is a Boost/Buck-Boost/Buck DCDC controller that drives an external power NMOS switch. It is suitable for driving high brightness LED for LED lighting applications.

### Features

- Battery operation : 3 V to 20 V
- 0 A to a few Amperes depending on rating of external NMOS and mode of operation • Output current range :
- Current mode control architecture
- LED dimming function available by using PWM signal
- 30 mV / 200 mV reference voltage
- Low standby current
- Configurable as either Boost/Buck-Boost/ Buck mode converter
- Built-in various protection circuit : Under voltage lock out

Over voltage protection

Soft start function

### Applications

- LED lighting module
- LED lantern applications
- White LED backlighting for LCD panel
- White LED flash light driving applications
- General LED back lighting

### Package

16 pin Plastic Shrink Small Outline Package (SSOP Type)

- Туре
  - Bi-CMOS IC

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- Application Circuit Example (Block Diagram)
  - Boost Mode



• This application circuit is an example. The operation of mass production set is not guaranteed. You should perform enough evaluation and verification on the design of mass production set. You are fully responsible for the incorporation of the above application circuit and information in the design of your equipment.

- Use external resistor with  $\pm 1\%$  accuracy at CS pin.
- Use ceramic type capacitor (Typ. 1  $\mu F,$  Min. 0.5  $\mu F)$  at VREG pin.
- Use schottky diode at VOUT.
- This block diagram is for explaining functions. The part of the block diagram may be omitted, or it may be simplified.

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- Application Circuit Example (Block Diagram) (continued)
  - Buck-Boost Mode



• This application circuit is an example. The operation of mass production set is not guaranteed. Perform enough evaluation and verification on the design of mass production set.

- Use external resistor with  $\pm 1\%$  accuracy at CS pin.
- Use ceramic type capacitor (Typ. 1  $\mu F,$  Min. 0.5  $\mu F)$  at VREG pin.
- Use schottky diode at VOUT.
- This block diagram is for explaining functions. The part of the block diagram may be omitted, or it may be simplified.

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- Application Circuit Example (Block Diagram) (continued)
  - Buck Mode



• This application circuit is an example. The operation of mass production set is not guaranteed. Perform enough evaluation and verification on the design of mass production set.

- Use external resistor with  $\pm 1\%$  accuracy at CS pin.
- Use ceramic type capacitor (Typ. 1  $\mu F,$  Min. 0.5  $\mu F)$  at VREG pin.
- Use schottky diode at VOUT.

• This block diagram is for explaining functions. The part of the block diagram may be omitted, or it may be simplified.

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### Pin Descriptions

Pin No.	Pin name	Туре	Description
1	GND	Ground	Ground
2	N.C.		
3	N.C.		—
4	VIN	Power Supply	Power Supply of IC
5	VFB_SEL	Input	Feedback voltage select
6	CS	Input	Current Sense
7	N.C.		—
8	OVP	Input	Over Voltage Protection input pin for Boost mode; Connect to GND for Buck mode
9	GNDP	Ground	Power Ground
10	N.C.		—
11	SW	Output	External NMOS Transistor Gate Drive
12	PWM	Input	PWM Dimming Control
13	N.C.		
14	ENB	Input	Standby On/Off Control
15	N.C.		
16	VREG	Output	Regulator Output

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### Absolute Maximum Ratings

Note) Absolute maximum ratings are limit values which do not result in damages to this IC, and IC operation is not guaranteed at these limit values.

A No.	Parameter	Symbol	Rating	Unit	Notes
1	Supply voltage	V <sub>DD</sub>	21	V	*1
2	GND pin current	I <sub>GND</sub>	—	А	
3	Power dissipation	P <sub>D</sub>	135	mW	*2
4	Operating ambient temperature	T <sub>opr</sub>	-25 to +85	°C	*3
5	Storage temperature	T <sub>stg</sub>	-55 to +125	°C	*3

Notes)\*1: The values under the condition not exceeding the above absolute maximum ratings and the power dissipation.

\*2 : The power dissipation shown is the value at  $T_a = 85^{\circ}C$  for the independent (unmounted) IC package without a heat sink. When using this IC, refer to the  $P_D$ - $T_a$  diagram of the package standard and design the heat radiation with sufficient margin so that the allowable value might not be exceeded based on the conditions of power supply voltage, load, and ambient temperature.

\*3 : Except for the power dissipation, operating ambient temperature, and storage temperature, all ratings are for  $T_a = 25^{\circ}C$ .

### Operating Supply Voltage Range

Parameter	Symbol	Range	Unit	Notes
Supply voltage range	V <sub>IN</sub>	3.0 to 20	V	*1
Supply voltage range (Boost Mode/Buck-Boost Mode)	V <sub>IN1</sub>	3.0 to 12	V	*1
Supply voltage range (Buck Mode)	V <sub>IN2</sub>	3.0 to 20	V	*1

Note) \*1: The values under the condition not exceeding the above absolute maximum ratings and the power dissipation.

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### Allowable Voltage Range

• Allowable current and voltage ranges are limit ranges which do not result in damages to this IC, and IC operation is not guaranteed within these limit ranges.

- Voltage values, unless otherwise specified, are with respect to GND.
- $V_{IN}$  is voltage for VIN pin.
- Do not apply external currents or voltages to any pin not specifically mentioned.

Pin No.	Pin name	Rating	Unit	Note
4	VIN	-0.3 to 20	V	*1
5	VFB_SEL	-0.3 to 5.5	V	_
6	CS	$-0.3$ to $V_{REG}$	V	*2
8	OVP	$-0.3$ to $V_{REG}$	V	*2
11	SW	-0.3 to V <sub>REG</sub>	V	*2
12	PWM	-0.3 to 5.5	V	*2
14	ENB	$-0.3$ to $V_{\rm IN}$	V	*1
16	VREG	-0.3 to 4.3	V	*2

- Notes) \*1 :  $V_{\rm IN}\,$  must not exceed 20 V.
  - \*2 :  $V_{REG}$  must not exceed 4.3 V.

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## ■ Electrical Characteristics at V<sub>IN</sub> = 6 V, ENB = 6 V, PWM = V<sub>REG</sub>

Note)  $T_a = 25^{\circ}C \pm 2^{\circ}C / T_a = 25^{\circ}C \pm 2^{\circ}C$  unless otherwise specified.

в			Test	<b>a</b>		Limits		Linit	
No.	Parameter	Symbol	circuits	Conditions	Min	Тур	Max	Unit	Notes
Circu	it Current Consumption								
1	Standby Current	I <sub>STB</sub>	1	ENB = 0 V	_		10	μΑ	
2	Operating Quiescent Current	I <sub>CC</sub>	1	$ENB = V_{IN}$ No load condition			1	mA	
ENA	BLE (ENB), VFB_SEL and PWM C	Control Fund	ction	-					
3	ENB High Input Logic	V <sub>ENBH</sub>	1	—	3		V <sub>IN</sub>	V	—
4	ENB Low Input Logic	V <sub>ENBL</sub>	1	—	0		0.3	V	—
5	VFB_SEL High Input Logic	V <sub>VFBSELH</sub>	1		$0.7 \times V_{REG}$	_	5	V	
6	VFB_SEL Low Input Logic	V <sub>VFBSELL</sub>	1		0		$0.3 \times V_{REG}$	V	
7	PWM High input Logic	V <sub>PWMH</sub>	1		$0.7 \times V_{REG}$		5	V	
8	PWM Low input Logic	V <sub>PWML</sub>	1		0		$0.3 \times V_{REG}$	V	
Input	Pin Current Consumption								
9	Enable Pin Current	I <sub>ENB</sub>	1	ENB = 6 V			25	μΑ	
Outp	ut Driver								
10	SW High Output Logic	V <sub>SWH</sub>	1	SW output High logic; MOSFET ON condition	$0.7 \times V_{REG}$	_	V <sub>REG</sub> +0.2	v	
11	SW Low Output Logic	V <sub>SWL</sub>	1	SW output Low logic; MOSFET OFF condition	-0.2		0.2	v	
Unde	er Voltage Lock Out (UVLO)			1					
12	Under Voltage protection on value	V <sub>UVLOON</sub>	1	$V_{IN}$ Falling SW OFF; $V_{REG}$ = no load	1.9	2.1	2.3	V	
13	Under voltage protection Hysteresis	V <sub>UVLOHYS</sub>	1	$V_{IN}$ Rising SW ON – $V_{IN}$ Falling SW OFF; $V_{REG}$ = no load	0.1	0.3	0.5	V	

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### Electrical Characteristics (Reference values for design) at $V_{IN} = 6 V$

Note)  $T_a = 25^{\circ}C \pm 2^{\circ}C / T_a = 25^{\circ}C \pm 2^{\circ}C$  unless otherwise specified.

The characteristics listed below are reference values derived from the design of the IC and are not guaranteed by inspection. If a problem does occur related to these characteristics, we will respond in good faith to user concerns.

в	Devementer	Cumhal	Test	Conditions	Refe	erence va	lues	les	
No.	Parameter	Symbol	circuits	Conditions	Min	Тур	Max	Unit	Notes
Refe	erence Voltage Control								
14	VFB Reference Voltage 1	$V_{VFB1}$	1	VFB_SEL = High OVP = 0 V (Buck mode)	196	202	208	mV	
15	VFB Reference Voltage 2	V <sub>VFB2</sub>	1	VFB_SEL = Low OVP = 0 V (Buck mode)	24	32	40	mV	
Over	Over Voltage Protection (Boost Mode Only)								
16	Over Voltage Protection Threshold	V <sub>OVP</sub>	_	$R1 = 470 \text{ k}\Omega, R2 = 30 \text{ k}\Omega$	18	21	24	v	
Outp	ut Driver	·							
17	Driver Off Time	T <sub>OFF</sub>	1	Fix off time at SW pin	0.5	1	2	μs	
18	Maximum Operating Frequency	F <sub>Max</sub>	1	_	_		1.5	MHz	
Regu	lator Voltage (VREG)		1						
19	VREG Output Voltage	V <sub>REG</sub>	1	$\begin{array}{l} 4 \ V \leq V_{IN} \leq 15 \ V \\ \text{No Load Condition,} \\ \text{CVREG} = 1 \ \mu\text{F} \end{array}$	3.45	3.65	3.85	v	
Effici	ency	·							
20	Efficiency	Eff		$V_{IN} = 6 V$ 1 LED of V <sub>F</sub> = 3.7 V I <sub>LED</sub> = 400 mA VFB_SEL = High OVP = 0 V (Buck mode)	_	90		%	

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### Control Pin Mode Table

Note) See parameters B No. 3, 4, 5, 6, 7 and 8 in the Electrical Characteristics for control voltage ranges.

Pin No.	Description		Pin vo	oltage	Remarks
PIII NO.	Desci	iption	Low	High	Remarks
5	VFB_SEL	ON/OFF	$V_{FB} = 32 \text{ mV}$	$V_{FB} = 202 \text{ mV}$	Feedback voltage selection
12	PWM	ON/OFF	PWM OFF	PWM ON	When PWM is not used, the pin is left floating
14	ENB	ON/OFF	STANDBY	OPERATING	Standby / Operating mode control

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### Test Circuit Diagram

1. Test Circuit 1



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### Electrical Characteristics Test Procedures

С	Demonster		Input		Output		Switch					
No.	Parameter	Pin No.	Conditions	Pin No.	Conditions	S1	S2	S3	S4	S5	S6	
Circui	t Current Consumption		·									
1	Standby Current	14 12 5 8 6 11	$ENB = 0 V$ $PWM = 0 V$ $VFB\_SEL = 0 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	4	Standby current consumption	5	5	5	2	2	1	
2	Operating Quiescent Current	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = 0 V$ $VFB\_SEL = 0 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	4	Current consumption	2	5	5	2	2	1	
ENAB	LE (ENB), VFB_SEL and	PWM	Control Function	1	-	1	1		1	1		
3	ENB High Input Logic	14 12 5 8 6 11	$ENB = 0.30 V$ $PWM = Hi-Z$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	16	Output DC voltage	3	1	2	2	2	1	
4	ENB Low Input Logic	14 12 5 8 6 11	$ENB = 3.0 V$ $PWM = Hi-Z$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	16	Output DC voltage	4	1	2	2	2	1	
5	VFB_SEL High Input Logic	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = Hi-Z$ $VFB\_SEL = 0.70 \times V_{REG}$ $OVP = 0 V$ $CS = 100 \text{ mV}$ $SW = Hi-Z$	11	Output DC voltage / 1 MHz	2	1	3	2	3	1	
6	VFB_SEL Low Input Logic	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = Hi-Z$ $VFB\_SEL = 0.30 \times V_{REG}$ $OVP = 0 V$ $CS = 100 \text{ mV}$ $SW = Hi-Z$	11	Output DC voltage / 1 MHz	2	1	4	2	3	1	
7	PWM High input Logic	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = 0.70 \times V_{REG}$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	11	Output DC voltage / 1 MHz	2	3	2	2	2	1	

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### ■ Electrical Characteristics Test Procedures (continued)

С			Input		Output		Switch					
No.	Parameter	Pin No.	Conditions	Pin No.	Conditions	S1	S2	S3	S4	S5	S6	
8	PWM Low input Logic	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = 0.30 \times V_{REG}$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	11	Output DC voltage / 1 MHz	2	4	2	2	2	1	
Input F	Pin Current Consumption											
9	Enable Pin Current	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = Hi-Z$ $VFB\_SEL = 0 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	14	Current Consumption	2	1	5	2	2	1	
Outpu	t Driver											
10	SW High Output Logic	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = 3.65 V$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = 0 A$	11	Output DC voltage / 1 MHz	2	2	2	2	2	2	
11	SW Low Output Logic	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = 0 V$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = 0 A$	11	Output DC voltage / 1 MHz	2	5	2	2	2	2	

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### Electrical Characteristics Test Procedures (continued)

С	_		Conditions		Output		Switch					
No.	Parameter	Pin No.			Conditions	S1	S2	S3	S4	S5	S6	
Under	voltage Lock Out (UVLO)											
12	Under voltage protection on value	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = Hi-Z$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	11	Output DC voltage / 1 MHz	2	1	2	2	2	1	
13	Under voltage protection Hysteresis	14 12 5 8 6 11	$ENB = V_{IN}$ $PWM = Hi-Z$ $VFB\_SEL = 3.65 V$ $OVP = 0 V$ $CS = 0 V$ $SW = Hi-Z$	11	Output DC voltage / 1 MHz	2	1	2	2	2	1	

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### Technical Data

• I/O block circuit diagrams and pin function descriptions

Note) The characteristics listed below are reference values derived from the design of the IC and are not guaranteed.

Pin No.	Waveform and voltage	Internal circuit	Impedance	Description
1	GND		—	Signal Ground
2		_		No connection
3		_		No connection
4	DC (3 V to 20 V)		Z : Low	VIN Power Supply of IC
5	DC (0 V to 5 V)	VREG (16 VFB_SEL (3) (5) (5) (7) (6) (6) (6) (6) (6) (7) (6) (6) (6) (6) (6) (6) (7) (6) (7) (6) (7) (7) (7) (7) (7) (7) (7) (7	Z : Hi-Z	VFB_SEL Feedback voltage select pin
6	DC (0 V to 250 mV)	VREG (16 CS 6 500 7000 7000 7000 7000 7000 7000	Z : Hi-Z	CS Current Sense Pin

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### Technical Data (continued)

• I/O block circuit diagrams and pin function descriptions (continued)

Note) The characteristics listed below are reference values derived from the design of the IC and are not guaranteed.

Pin No.	Waveform and voltage	Internal circuit	Impedance	Description
7		_	_	No connection
8	DC (0 V to 1.26 V)	VREG 16 0VP 8 1k	Z : Hi-Z	OVP Over Voltage Protection input pin for Boost and Buck-Boost mode Connect to GND for Buck mode
9	GNDP	_		Power Ground
10		—	—	No connection
11	Pulse (0 V to 3.65 V)	VREG	Z : Hi-Z	SW External NMOS Transistor Gate Driving Pulse
12	Pulse (0 V to 5 V)	$PWM \xrightarrow{167k} 4$	Ζ : 170 kΩ	PWM PWM Dimming Control

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### Technical Data (continued)

• I/O block circuit diagrams and pin function descriptions (continued)

Note) The characteristics listed below are reference values derived from the design of the IC and are not guaranteed.



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### Technical Data (continued)

· Functions and properties descriptions

### (1) Overview

AN30888A is a constant current LED driver. The IC works as a Boost /Buck-Boost/ Buck mode DCDC controller with external MOSFET.

Operating input voltages ranges from 3 V to 20 V. The mode of operation depends on the number of LEDs to be driven and the supply voltage level.

In general, please adhere to the following:

If total LED voltage drop is more than supply voltage, boost mode is adopted. If LED voltage drop is less than the supply voltage, buck mode is adopted . If supply voltage is close to the total LED voltage drop, the Buck-Boost mode can be used. Please note that the different mode of operation should be manually configured.

Output LED current can be designed ranges from 0 A and to a few amperes depending on the mode of operation, the external MOSFET characteristic and feasible  $R_{CS}$  value used. The control architecture uses current mode fix off time control. The  $V_{FB}$  reference voltage determines LED current by setting VFB\_SEL pin with values of 32 mV or 202 mV under buck mode. By applying V<sub>FB</sub> voltage of 32 mV, user can achieve higher efficiency with lower power dissipation in  $R_{CS}$  resistor. Applying 202 mV  $V_{FB}$  voltage achieves better LED current accuracy.

### (2) Standby enable function

AN30888A enters standby mode when ENB pin is pulled low. During standby, the IC draws a small current of value less than 10 µA from the power supply. This helps to achieve longer battery usage time. During Boost mode operation, although external MOSFET cannot be turned on at standby condition, there is still a DC current path between the input and the LEDs through the inductor and schottky diode. Thus it is important to make sure that during boost mode, the minimum forward voltage of the LED array must exceed the maximum input voltage to ensure the LEDs remain off during standby mode.

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• Functions and properties descriptions (continued)

(3) Internal regulator

An internal 3.65 V regulator is used as the power supply for internal core circuit of this IC.

This regulated voltage,  $V_{REG}$  will be provided when  $V_{IN}$  is approximately in the range of 4 V to 20V. For  $V_{IN}$  lower than 4 V, regulator will act as a  $V_{IN}$  voltage follower, with output voltage close to  $V_{IN}$ . The amount of drop voltage from  $V_{IN}$  during  $V_{IN}$  follower mode depends on load current of the regulator and also tolerance of the IC. In general, the regulator output voltage will be approximately 0.3 V lower than  $V_{IN}$  during this mode of operation.

This regulator requires a capacitor of 1  $\mu$ F to be connected to VREG pin. This capacitor helps to provide a stable regulated voltage to the IC. The regulator has a current ability of approximately 15 mA. However, it is not designed to provide as external power supply voltage. Hence an external load exceeding approximately 0.5 mA to the VREG pin is not allowed.

### (4) Output setting consideration

The output voltage,  $V_{\text{OUT}}$  is set using the following equations for both boost and buck mode:

$V_{OUT} = (V_F \times N_{LEDs} + V_D) \dots$	Eq[1]	(Boost mode)
$V_{OUT} = (V_{IN} - V_F \times N_{LEDs}) \dots$	Eq[2]	(Buck mode)
$V_{OUT} = (V_F \times N_{LEDs} + V_D + V_{IN}) \dots$	Eq[3]	(Buck-Boost mode)
V <sub>IN</sub> : Battery or Input power supply voltage		
V <sub>F</sub> : LED forward drop voltage		
N <sub>LEDs</sub> : Number of LEDs stacked in series		
V <sub>D</sub> : Schottky diode forward drop voltage		

For Boost mode or Buck-Boost mode operation,  $V_{OUT}$  setting should be lesser than Drain–Source breakdown voltage of external MOSFET as mention in (11). Also  $V_{OUT}$  should be lesser than OVP protection threshold as mentioned in (9).

For Buck mode operation,  $V_{OUT}$  setting should give sufficient voltage for external MOSFET to operate properly at the required output current setting.

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• Functions and properties descriptions (continued)

(5) Feedback voltage  $V_{FB}$  at CS pin

The  $V_{FB}$  voltage is generated internally in the IC and output at CS pin. This voltage allows users to fix the input peak current,  $I_{PK}$  as well as the LED output current,  $I_{LED}$ . This voltage will change according to the setting at VFB\_SEL pin.

For operation in boost mode/buck-boost mode,  $V_{FB}$  will be inversely proportionally to supply voltage,  $V_{IN}$ . When input supply voltage decreases,  $V_{FB}$  will increase. This ensure LED current remain accurate as supply voltage decreases. When operating in buck mode,  $V_{FB}$  voltage will remain at 202 mV or 32 mV depending on whether VFB\_SEL pin is high or low.

The following are some figures of  $V_{FB}$  voltage with respect to  $V_{IN}$ . For detail information, please refer to graph and data table information as the following.

For Boost mode and Buck-Boost mode :

 $\begin{array}{ll} V_{FB} = 116 \mbox{ mV} & (When \mbox{ VFB}SEL = High \ ; \ V_{IN} = 6 \ V) \\ V_{FB} = 50 \mbox{ mV} & (When \ VFB}SEL = Low \ ; \ V_{IN} = 6 \ V) \\ V_{FB} = 198.3 \mbox{ mV} & (When \ VFB}SEL = High \ ; \ V_{IN} = 3 \ V) \\ V_{FB} = 88 \mbox{ mV} & (When \ VFB}SEL = Low \ ; \ V_{IN} = 3 \ V) \\ \end{array}$ 

For Buck mode :

$V_{FB} = 202 \text{ mV}$	(When VFB_	SEL = High,	for all V <sub>IN</sub> level)
$V_{FB} = 32 \text{ mV}$	(When VFB_	SEL = Low,	for all V <sub>IN</sub> level)

To improve overall efficiency  $V_{FB}$  voltage can be set lower by switching VFB\_SEL = Low. On the other hand accuracy can be improved by using VFB\_SEL = High mode.

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### Technical Data (continued)

• Functions and properties descriptions (continued)

(5) Feedback voltage  $V_{FB}$  at CS pin (continued)





		mode/ ost mode	Buck mode	
V <sub>IN</sub> (V)	VFB_SEL = High	VFB_SEL = Low	VFB_SEL = High	VFB_SEL = Low
	V <sub>FB</sub> (mV)	V <sub>FB</sub> (mV)	V <sub>FB</sub> (mV)	V <sub>FB</sub> (mV)
3	198.3	88.0	202	32
4	161.0	71.0	202	32
5	132.3	57.7	202	32
6	116.0	50.0	202	32
7	98.3	43.0	202	32
8	86.3	38.0	202	32
9	77.3	34.0	202	32
10	70.0	31.0	202	32
11	64.0	28.3	202	32
12	59.0	26.3	202	32
13	N.A	N.A	202	32
14	N.A	N.A	202	32
15	N.A	N.A	202	32

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• Functions and properties descriptions (continued)

(6) Inductor selection

Inductor value, L is set by the required inductor ripple current desired.

The general trend for lower inductor value is a smaller inductor physical size, but a larger input ripple current. Similarly, an increase in inductor value will decease input ripple current.

Users are advice to choose an inductor that can handle the peak current  $I_{PK}$ , flowing across it without saturating. In addition, inductor with lower series resistance are preferred to provide better operating efficiency.

The following equation gives a general guideline in selection inductor value based on 30% peak to peak ripple current across the inductor.

L =	$(V_{OUT} - V_{IN}) \times T_{OFF}$		Eq[4]	(Boost mode, Buck-Boost mo	
	$0.3 \times I_{\rm IN}$		-4[.]	(20000 110 40)	
	$(V_{IN} - V_{OUT}) \times T_{OFF}$				
L =	$0.3  imes I_{LED}$	–	Eq[5]	(Buck mode)	
	V <sub>OUT</sub>		/ Output voltage		
	V <sub>IN</sub>		/ Input supply voltage		
	T <sub>OFF</sub>		/ Fixed off time design at 1 µs		
	I <sub>LED</sub>		/ LED output current		
	I <sub>IN</sub>		/ $I_{IN}$ is input current from	n supply voltage	e

Please note that the 0.3 factor can be altered if 30% peak to peak current is changed. i.e, if percentage of peak to peak current needed is 40%, this factor will be 0.4.

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• Functions and properties descriptions (continued)

(7) Setting output LED current and choosing current sense resistor  $R_{CS}$ 

The LED current in this IC can be set easily by selecting the appropriate R<sub>CS</sub> resistor to be used at CS pin of this chip.

For Boost Mode and Buck-Boost mode :

R<sub>CS</sub> resistor can be set in the following way :

First is to calculate input current  $I_{\mbox{\scriptsize IN}}$  at the required operating condition :

$I_{\rm IN} = (V_{\rm OUT} + V_{\rm D}) \times (I_{\rm LED} / V_{\rm IN}) \$	Eq[6]	(Boost mode,	Buck-Boost mode)
V <sub>OUT</sub>	/ Output voltage		
V <sub>D</sub>	/ Schottky diode forward drop voltage		
I <sub>LED</sub>	/ Required LED current		
$V_{IN}$	/ Input supply voltage		

After which the peak input current,  $I_{PK}$  can be determine by adding  $I_{IN}$  with half the peak to peak ripple current at the inductor.

	$(V_{OUT} - V_{IN}) \times T_{OFF}$			
$I_{PK} = I_{IN} + \dots$	2L	 Eq[7]	(Boost mode,	Buck-Boost mode)
T <sub>OFF</sub>		/ T <sub>OFF</sub> is fixed off time =	= 1 μs	
L		/ Inductor value found in	n part (6) induct	or selection
V <sub>OUT</sub>		/ Output voltage		
$V_{IN}$		/ Input supply voltage		
I <sub>IN</sub>		/ $I_{IN}$ is input current four	nd in Eq[6]	

Lastly,  $R_{CS}$  resistor can be determine by using :

$$R_{CS} = \frac{V_{FB}}{I_{PK}}$$
 (Boost mode, Buck-Boost mode)

 $V_{FB}$  is voltage at CS pin. Refer to data graphs for the  $V_{FB}$  voltage at different input voltage condition.  $I_{PK}$  is peak current found in Eq[7]

Using numeric example of operating condition :

$$V_{IN} = 6 \text{ V}, V_{OUT} = 10 \text{ V}, I_{LED} = 500 \text{ mA}, T_{OFF} = 1 \text{ } \mu\text{s}, L = 16 \text{ } \mu\text{H}, V_D = 0.4 \text{ V}, V_{FB} = 0.1 \text{ } V@V_{IN} = 6 \text{ V}$$
  
From Eq[6]:  $I_{IN} = (10 + 0.4) \times (0.5 / 6) = 0.8667 \text{ A}$   
From Eq[7]:  $I_{PK} = 0.8667 + \frac{(10 - 6) \times 1\mu}{2 \times 16\mu} = 0.9971 \text{ A}$   
From Eq[8]:  $R_{CS} = \frac{0.1}{0.9917} = 100.8 \text{ m}\Omega$ 

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• Functions and properties descriptions (continued)

(7) Setting output LED current and choosing current sense resistor R<sub>CS</sub> (continued)

For Buck Mode :

R<sub>CS</sub> resistor can be set in the following way :

First is to calculate the peak current  $I_{PK}$  using Eq[9]. During buck mode, peak current sense correspond to the average output LED current plus half of actual current ripple through the inductor.

$I_{PK} = I_{LED} + \dots$	$(V_{\rm IN} - V_{\rm OUT}) \times T_{\rm OFF}$ $2L$	 Eq[9]	(Buck mode)
$f V_{OUT} \ V_{IN}$		/ Output voltage / Input supply voltage	
T <sub>OFF</sub>		$/ T_{OFF} =$ Fixed off time d	lesign at 1 µs
I <sub>LED</sub> L		/ LED output current / L = Inductor value four	nd in part (6) inductor selection

Lastly, R<sub>CS</sub> resistor can be determine by using :

$R_{CS} = \frac{V_{FB}}{\dots}$	Eq[10]	(Buck mode)
I <sub>PK</sub>		

 $V_{FB}$  is voltage at CS pin. Refer to data graphs for the  $V_{FB}$  voltage at different input voltage condition.  $I_{PK}$  is peak current found in Eq[9].

Using numeric example of operating condition :

$$V_{IN} = 12 \text{ V}, V_{OUT} = 2 \text{ V}, I_{LED} = 500 \text{ mA}, T_{OFF} = 1 \text{ } \mu\text{s}, L = 66 \text{ } \mu\text{H}, V_{FB} = 0.2 \text{ V}$$
  
From Eq[9]:  $I_{PK} = 0.5 + \frac{(12 - 2) \times 1\mu}{2 \times 66\mu} = 0.575 \text{ A}$   
From Eq[10]:  $R_{CS} = \frac{0.2}{0.575} = 348 \text{ m}\Omega$ 

Please note that for component deviation such as inductor, diodes, etc, these deviation can cause the designed  $I_{PK}$  to be higher or lower than the calculated value.

Users may need to fine tune the value of  $R_{CS}$  from the calculated values in order to obtain accurate  $I_{LED}$  measurement. Please take note of total impedance including parasitic impedance of PCB trace at CS pin to ground when designing the required  $R_{CS}$  value. This is especially important if the designed  $I_{LED}$  is high as  $R_{CS}$  value will be small and in turn making parasitic impedance significant to the total impedance seen at CS pin

• Functions and properties descriptions (continued)

### (8) Soft start

Soft start circuit is incorporated into this IC to avoid high in-rush current during start-up.

After the device is enabled (ENB = High), the output inductor current and output voltage will rise slowly from initial condition. This slow start-up time ensure smooth start-up as well as minimize in-rush current.

### (9) Over Voltage Protection (OVP)

When operating in Boost mode or Buck-Boost mode, over voltage protection is needed to prevent damages to IC or external component damages in cases of open LED condition.

OVP switches off external power MOSFET to prevent output from rising over a designed OVP voltage. Output voltage should be limited to the rating of external component used. (for example Drain Source voltage rating of the external MOSFET or the output capacitor)

OVP compares the internal reference voltage of 1.26 V with output voltage through resistor network.

OVP threshold is set using the following equation:

 $V_{OVP} = \frac{1.262 \text{ V} \times (\text{R1} + \text{R2})}{\text{R2}} \dots \text{Eq[11]}$ (Boost mode, Buck-Boost mode)

If  $R1 = 470 \text{ k}\Omega$ ,  $R2 = 30 \text{ k}\Omega$ , OVP threshold will be designed at around 21 V.

When OVP is triggered, output voltage will be clamped at this threshold voltage (with hysteresis of around 1 V to 2 V) until the fault (e.g open LED condition) has been removed.

When operating in buck mode, the OVP pin must be short to ground to disable this function as OVP function is not necessary in this mode.

• Functions and properties descriptions (continued)

(10) Under Voltage Lock Out (UVLO)

Under Voltage lock out prevents IC from operation at supply voltage lower than 2.1 V.

This function prevent IC from abnormal operation when supply voltage  $V_{IN}$  drops below our recommended input range. When input voltage is lower than this lock out value of 2.1 V, external MOSFET will be switched off. When input voltage rises to 2.4 V or more, device operation starts again. This means a hysteresis voltage of about 0.3 V.

### (11) Power MOSFET consideration

When selecting the power MOSFET, it is important to consider parameters such as gate-source, drain-source breakdown voltage, total gate capacitance, ON resistance and the drain current rating.

When power is turned on for IC operating in boost mode, output voltage needed to drive LED will be reflected to Drain-Source voltage of the power MOSFET. Thus it is recommended to select a MOSFET that can handle this output voltage. Alternatively, output and Drain-Source voltage can be protected and clamped by OVP circuit as mentioned in point (9).

Gate capacitance of the MOSFET chosen should ideally to be smaller than 3 nF.

### (12) PWM operation

PWM signal can be generated externally and input into PWM pin of this IC. This PWM signal will turn on and off the output driver, giving an average output LED current that is proportional to the duty cycle of the PWM signal.

 $I_{LED(avg.)} = I_{LED} \times Duty...$  Eq[12] (Boost / Buck-Boost / Buck mode)

 $I_{\text{LED(avg.)}=}$  The average output LED current after PWM is input  $I_{\text{LED}} =$  The nominal LED current set in part (7)

Duty = The ratio of on pulse time compared to total period time of the PWM signal.

A PWM frequency of 1 kHz or lower is recommended to minimize error due to rise and fall time of the converter output.

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• Functions and properties descriptions (continued)

### (13) Maximum duty operation

Maximum Duty limitation is needed when operating in Boost or Buck- Boost mode. This prevents the output voltage from having abnormal operation. For Buck mode, there is no need for maximum duty limit as SW pin is able to switch to 100% duty. Please refer to the graph below for maximum duty vs  $V_{IN}$  data for Boost and Buck-Boost mode operation.

Boost/Buck-boost mode				
V <sub>IN</sub> (V) Duty Limit (%)				
3.0	88.73			
3.5	87.09			
4.0	85.27			
4.5	83.58			
5.0	81.99			
5.5	79.79			
6.0	78.40			
6.5	77.38			
7.0	76.25			
7.5	75.19			
8.0	73.92			
8.5	72.89			
9.0	71.78			
9.5	70.83			
10.0	69.91			
10.5	68.90			
11.0	67.97			
11.5	67.09			
12.0	66.33			



### (14) Minimum duty operation

Parasitic circuit capacitance, inductance and external MOSFET gate drive current can create spike in the current sense, CS pin voltage at the point when external MOSFET is switched on. In order to prevent this spike to terminate the ON time prematurely, an internal filter of time constant, 100 ns is designed in chip. This time constant of 100 ns translates to a minimum duty of around 9% for all modes of operation. To further reduce the spike in the CS voltage especially when operating in low  $I_{LED}$  condition (example:  $R_{CS}$  is more than 0.8  $\Omega$  or more), external RC filter can be used in between  $V_{FB}$  node to CS pin which act as a low pass filter to filter spike noise from entering CS pin. This RC filter time constant should be long enough to reduce the parasitic spike without significantly affecting the shape of CS voltage.

The recommended RC value range from :  $R = 10 \Omega$  to 1 k $\Omega$  and C = 100 pF to 500 pF depending on mode of operation and spike level.

Product \$	Standards
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### Usage Notes

- Special attention and precaution in using
  - 1. This IC is intended to be used for general electronic equipment [LED Lighting Devices].
    - Consult our sales staff in advance for information on the following applications:
    - Special applications in which exceptional quality and reliability are required, or if the failure or malfunction of this IC may directly jeopardize life or harm the human body.
    - Any applications other than the standard applications intended.
      - (1) Space appliance (such as artificial satellite, and rocket)
      - (2) Traffic control equipment (such as for automobile, airplane, train, and ship)
      - (3) Medical equipment for life support
      - (4) Submarine transponder
      - (5) Control equipment for power plant
      - (6) Disaster prevention and security device
      - (7) Weapon
      - (8) Others : Applications of which reliability equivalent to (1) to (7) is required

It is to be understood that our company shall not be held responsible for any damage incurred as a result of or in connection with your using the IC described in this book for any special application, unless our company agrees to your using the IC in this book for any special application.

- 2. Pay attention to the direction of LSI. When mounting it in the wrong direction onto the PCB (printed-circuit-board), it might smoke or ignite.
- 3. Pay attention in the PCB (printed-circuit-board) pattern layout in order to prevent damage due to short circuit between pins. In addition, refer to the Pin Description for the pin configuration.
- 4. Perform a visual inspection on the PCB before applying power, otherwise damage might happen due to problems such as a solderbridge between the pins of the semiconductor device. Also, perform a full technical verification on the assembly quality, because the same damage possibly can happen due to conductive substances, such as solder ball, that adhere to the LSI during transportation.

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### ■ Usage Notes (continued)

### • Special attention and precaution in using (continued)

- Take notice in the use of this product that it might break or occasionally smoke when an abnormal state occurs such as output pin-V<sub>DD</sub> short (Power supply fault), output pin-GND short (Ground fault), or output-to-output-pin short (load short). And, safety measures such as an installation of fuses are recommended because the extent of the above-mentioned damage and smoke emission will depend on the current capability of the power supply.
- 6. When designing your equipment, comply with the range of absolute maximum rating and the guaranteed operating conditions (operating power supply voltage and operating environment etc.). Especially, please be careful not to exceed the range of absolute maximum rating on the transient state, such as power-on, power-off and mode-switching. Otherwise, we will not be liable for any defect which may arise later in your equipment.

Even when the products are used within the guaranteed values, take into the consideration of incidence of break down and failure mode, possible to occur to semiconductor products. Measures on the systems such as redundant design, arresting the spread of fire or preventing glitch are recommended in order to prevent physical injury, fire, social damages, for example, by using the products.

- 7. When using the LSI for new models, verify the safety including the long-term reliability for each product.
- 8. When the application system is designed by using this LSI, be sure to confirm notes in this book. Be sure to read the notes to descriptions and the usage notes in the book.

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	PACK	AGE S	STAND	ARDS	,
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	Package Co	de	SSOP016-F	2-0225E	
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			,		
	H.Shidooka	H.Yoshida	M.Okajima	M.Itoh	

	PACKAGE STANDARDS		
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## 1. Outline Drawing

### Unit:mm



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## 2. Package Structure (Technical Report : Reference Value)

Chip Material		Si	(1)	
Leadframe material		Cu alloy	2	
Inner lead surface		Ag plating	3	
Outer lead surface		SnBi plating	4	
Chip mount	Method	Resin adhesive method	5	
	Material	Adhesive material		
Wirebond	Method	Thermo-compression bonding	6	
Wilebond	Material	Au		
Molding	Method	Transfer molding		
Molding	Material	Epoxy resin	ש ך	
Mass		72 mg		



PACKAGE STANDARDS		
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3. Mark Drawing



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## 4. Power Dissipation (Technical Report)



### 5. Power Dissipation (Supplementary Explanation)

#### [Experiment environment]

Power Dissipation (Technical Report) is a result in the experiment environment of SEMI standard conformity. (Ambient air temperature (Ta) is 25 degrees C)

#### [Supplementary information of PWB to be used for measurement]

The supplement of PWB information for Power Dissipation data (Technical Report) are shown below.

Indication	Total Layer	Resin Material	
Glass-Epoxy	1-layer	FR-4	
4-layer	4-layer	FR-4	

#### [Notes about Power Dissipation (Thermal Resistance)]

Power Dissipation values (Thermal Resistance) depend on the conditions of the surroundings, such as specification of PWB and a mounting condition, and a ambient temperature. (Power Dissipation (Thermal Resistance) is not a fixed value.)

The Power Dissipation value (Technical Report) is the experiment result in specific conditions (evaluation environment of SEMI standard conformity), and keep in mind that Power Dissipation values (Thermal resistance) depend on circumference conditions and also change.

#### [Definition of each temperature and thermal resistance]

Ta : Ambient air temperature

The temperature of the air is defined at the position where the convection, radiation, etc. don't affect the temperature value, and it's separated from the heating elements.

- Tc : It's the temperature near the center of a package surface. The package surface is defined at the opposite side if the PWB.
- Tj : Semiconductor element surface temperature (Junction temperature.)
- Rth(j-c) : The thermal resistance (difference of temperature of per 1 Watts) between a semiconductor element junction part and the package surface
- Rth(c-a) : The thermal resistance (difference of temperature of per 1 Watts) between the package surface and the ambient air

Rth(j-a) : The thermal resistance (difference of temperature of per 1 Watts) between a semiconductor element junction part and the ambient air



Recommended		
Coldoring Conditions	Total pages	page
Soldering Conditions	2	1

# Product name : AN30888A-VF Package : SSOP016-P-0225E

# 1. Recommended Soldering Conditions

In case that the semiconductor packages are mounted on the PCB, the soldering should be performed under the following conditions.

## 1 Reflow soldering

# Reflow peak temp. : max. 260 °C



No.	mark	contents	value
1	T1	Pre-heating temp.	150 °C∼180 °C
2	t1	Pre-heating temp. hold time	60 s∼120 s
3	а	Rising rate	2 °C/s~5 °C/s
4	Тр	Peak temp.	255 °C+5 °C、-0 °C
5	tp	Peak temp. hold time	10 s±3 s
6	tw	High temp. region hold time	within 60 s $~(\geqq220~^\circ\!C)$
7	b	Down rate	2 °C/s~5 °C/s
8	-	Number of reflow	within 2 times

\* Peak temperature : less than 260  $^\circ\!\mathrm{C}$ 

\*Temperature is measured at package surface point

## ② Wave soldering (Flow soldering)

\*Temp. of solder : 260 °C or less

- \*Soak time : within 5 s
- \*Number of flow : only 1 time

## ③ Manual soldering

- \* Iron Temperature : 350 °C or less (Device lead temperature : 270 °C, 10 s max.)
- \* Soldering time : within 3 s
- \*Number of manual soldering : only 1 time

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Soldering Conditions	2	2

# 2. Storage environment after dry pack opening



**\*** Because the taping and the magazine materials are not the heat-resistant materials, the bake at 125°C cannot be done. Therefore, please solder everything or control everything in the rule time. Please keep them in an equal environment with the moisture-proof packaging or dry box.

(Temperature: room temperature, relative humidity: 30% or less.)

To control storage time, when bake in the taping and the magazine is necessary, it is necessary for each type to set a bake condition. Please inquire of our company.

### ☆ AN30888A-VF limitation, low temperature bake condition : 40 °C / 25 %RH or less / 192 h

# 3. Note

- (1) Storage environment conditions: keep the following conditions Ta=5  $^{\circ}C \sim 30 ^{\circ}C$ , RH=30 %  $\sim 70 ^{\circ}N$ .
- 2 Storage period before opening dry pack shall be 1 year from a shipping day under Ta=5  $^{\circ}C \sim 30 ^{\circ}C_{3}$ RH=30 %  $\sim$  70 %. When the storage exceeds, Bake at 125 °C with 15 h to 25 h.
- 3 Baking cycle should be only one time.

Please be cautious of solderability at baking.

- (4) In case that use reflow two times, 2nd reflow must be finished within 336 hours.
- (5) Remove flux sufficiently from product in the washing process.

(Flux : Chlorineless rosin flux is recommended.)

6 In case that use ultrasonic for product washing,

There is the possibility that the resonance may occur due to the frequency and shape of PCB.

It may be affected to the strength of lead. Please be cautious of this matter.

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Prepared	Revised

Recommended		
Land Dattern	Total pages	page
Land Pattern	1	1

Integrated circuits

The land width for surface-mount semiconductor devices is related to on the amount of solder applied. It is necessary to optimize the land dimensions together with the method of solder application. Previously, EIAJ defined recommended land dimensions by package type, but because of the above reasons, currently only the positions where the terminals should be (the terminal land area) are standardized. For reference, the land design guidelines previously defined by EIAJ are shown below. During the actual design of the printed circuit board, the solder application method, etc. should be thoroughly considered before deciding.







$$l_2 \ge L + \beta_1 + \beta_2$$
$$b \le b_2 \le e - \gamma$$

- Y : Solder resistance bridge (Y = 0.3 mm)
- b : Terminal width
- $\beta_1$  : Soldering strength ( $\beta_1 = 0.3 \text{ mm}$ )
- $\beta_2$  : Solder mask pattern accuracy or soldering guideline ( $\beta_2 = 0.2 \text{ mm}$ )
- Note: Values within parentheses ( ) are recommended values.

The above size is calculated based on the experiment results by Matsushita Electric Industrial Co., Ltd., and is not intended as a guarantee of mounting reliability.
Mounting reliability can vary depending on factors such as the equipment specifications and conditions material specifications and properties, and environmental conditions.
To ensure satisfactory results, your company should evaluate and confirm actual mounting performance.

Packing Specification	Total pages page	
	3	1

Specifications of packing by the embossment tape (Specifications for dampproof packing of the reel without the inner carton)



2009.03.09
Prepared Revi



Reel	2000 PCS	Reel	X	IPCS
Packing case	10000 Pcs	Reel	×	5Pcs



Unit : mm



# Panasonic

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