

## Product Overview

The NSI1305 is a high performance  $\Sigma\Delta$  modulator with output separated from input based on the NOVOSENSE capacitive isolation technology. The device has a linear differential input signal range of  $\pm 50\text{mV}$  ( $\pm 64\text{mV}$  full-scale) or  $\pm 250\text{mV}$  ( $\pm 320\text{mV}$  full-scale). The differential input is ideally suited to shunt resistor-based current sensing in high voltage applications where isolation is required.

The analog input is amplified and continuously sampled by a second-order  $\Sigma\Delta$  modulator and converted to a high speed, single bit data stream. The output data is synchronous to the external clock with a frequency range from 5MHz to 21MHz. By using an appropriate digital filter (such as sinc3 filter) to decimate the bitstream, the device can achieve 16 bits resolution and an 86dB/82.5dB signal to noise ratio (SNR) at 78.125kSPS with a 20MHz master clock.

The fail-safe functions including input common-mode overvoltage detection and missing AVDD detection simplify system-level design and diagnostics.

## Key Features

- Up to 5000V<sub>RMS</sub> Insulation Voltage
- Clock frequency: 5MHz to 21MHz
- $\pm 50\text{mV}$  or  $\pm 250\text{mV}$  Linear Input Voltage Range
- Excellent DC Performance:
  - Offset Error:  $\pm 50\mu\text{V}$  or  $\pm 100\mu\text{V}$  (Max)
  - Offset Drift:  $-0.5\text{~}1.5\mu\text{V}/^\circ\text{C}$  (Max)
  - Gain Error:  $\pm 0.2\%$  (Max)
  - Gain Drift:  $\pm 40\text{ppm}/^\circ\text{C}$  (Max)
- SNR: 82.5dB or 86dB (Typ)
- High CMTI: 150kV/ $\mu\text{s}$  (Typ)
- System-Level Diagnostic Features:
  - AVDD monitoring
  - Input common-mode overvoltage detection

- Operation Temperature:  $-40^\circ\text{C}\text{~}125^\circ\text{C}$
- RoHS-Compliant Packages:
  - SOP8(300mil)
  - SOP16(300mil)

## Safety Regulatory Approvals

- UL recognition: up to 5000V<sub>rms</sub> for 1 minute per UL1577
- CQC certification per GB4943.1-2011
- CSA component notice 5A approval IEC60950-1 standard
- DIN VDE V 0884-11:2017-01

## Applications

- Shunt current monitoring
- AC motor controls
- Power and solar inverters
- Uninterruptible Power Suppliers
- Automotive onboard chargers

## Device Information

Part Number	Package	Body Size
NSI1305x-DSWVR	SOP8(300mil)	5.85mm $\times$ 7.50mm
NSI1305x-DSWR	SOP16(300mil)	10.30mm $\times$ 7.50mm

## Functional Block Diagrams

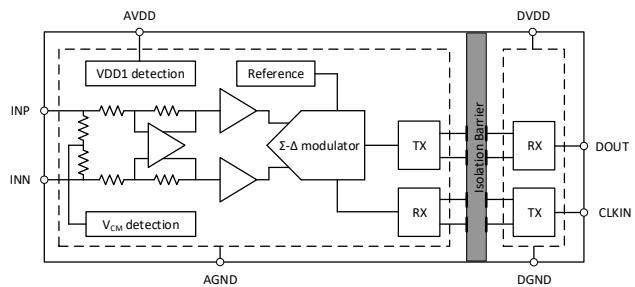


Figure 1. NSI1305 Block Diagram

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## 1. Pin Configuration and Functions

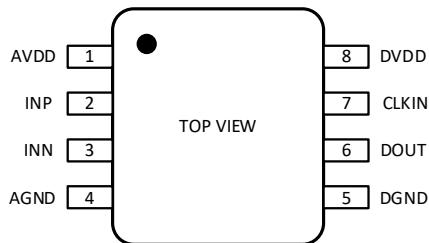


Figure 1.1 NSi1305 Package (SOP8(300mil))

Table 1.1 NSi1305 Pin Configuration and Description

NSi1305 PIN NO.	SYMBOL	FUNCTION
1	AVDD	Power supply for analog side (3.0V to 5.5V)
2	INP	Positive analog input ( $\pm 250\text{mV}$ recommended for NSi1305M25 and $\pm 50\text{mV}$ recommended for NSi1305M05)
3	INN	Negative analog input
4	AGND	Analog ground reference
5	DGND	Digital ground reference
6	DOUT	Modulator data output
7	CLKIN	Modulator clock input: 5~21MHz
8	DVDD	Power supply for digital side (3.0V to 5.5V)

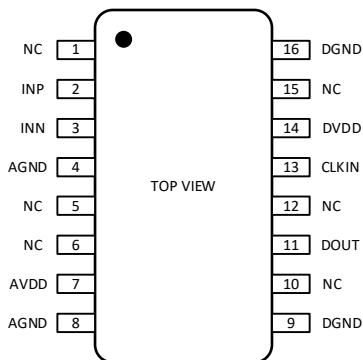


Figure 1.2 NSi1305 Package (SOP16(300mil))

Table 1.2 NSi1305 Pin Configuration and Description

<b>NSi1305 PIN NO.</b>	<b>SYMBOL</b>	<b>FUNCTION</b>
1	NC	Internally connected to AVDD, this pin can be left floating or tied to AVDD
2	INP	Positive analog input ( $\pm 250\text{mV}$ recommended for NSI1305M25 and $\pm 50\text{mV}$ recommended for NSI1305M05)
3	INN	Negative analog input
4	AGND/NC	Not internally connected. This pin can be tied to AGND, or leave floating.
5	NC	Not internally connected, this pin can be left floating or tied to AVDD, AGND
6	NC	Not internally connected, this pin can be left floating or tied to AVDD, AGND
7	AVDD	Power supply for analog side (3.0V to 5.5V)
8	AGND	Analog ground reference
9	DGND	Digital ground reference
10	NC	Not internally connected, this pin can be left floating or tied to DVDD, DGND
11	DOUT	Modulator data output
12	NC	Not internally connected, this pin can be left floating or tied to DVDD, DGND
13	CLKIN	Modulator clock input: 5~21MHz
14	DVDD	Power supply for digital side (3.0V to 5.5V)
15	NC	Not internally connected, this pin can be left floating or tied to DVDD, DGND
16	DGND	Digital ground reference

## 2. Absolute Maximum Ratings

Parameters	Symbol	Min	Typ	Max	Unit
Power Supply Voltage	AVDD, DVDD	-0.3		6.5	V
Analog Input Voltage	INP, INN	AGND-6		AVDD+0.5	V
Digital Input Voltage	CLKIN	DGND-0.5		DVDD+0.5	V
Digital Output Voltage	DOUT	DGND-0.5		DVDD+0.5	V
Output current per Output Pin	Io	-10		10	mA
Operating Temperature	T <sub>OPR</sub>	-40		125	°C
Junction Temperature	T <sub>J</sub>	-40		150	°C
Storage Temperature	T <sub>STG</sub>	-55		150	°C
Electrostatic discharge	HBM <sup>(1)</sup>	±2000			V
	CDM <sup>(2)</sup>	±1000			V

(1) Human body model (HBM), per AEC-Q100-002-RevD

(2) Charged device model (CDM), per AEC-Q100-011-RevB

## 3. Recommended Operating Conditions

Parameters	Symbol	Min	Typ	Max	Unit
Analog Side Power Supply	AVDD	3.0	5.0	5.5	V
Digital Side Power Supply	DVDD	3.0	3.3	5.5	V
NSI1305M05	Differential input voltage before clipping output	V <sub>Clipping</sub>		±64	mV
	Linear differential input full scale voltage	V <sub>FSR</sub>	-50	50	mV
	Operating common-mode input voltage	V <sub>CM</sub>	-0.032	0.8	V
NSI1305M25	Differential input voltage before clipping output	V <sub>Clipping</sub>		±320	mV
	Linear differential input full scale voltage	V <sub>FSR</sub>	-250	250	mV
	Operating common-mode input voltage	V <sub>CM</sub>	-0.16	0.8	V
Operating Ambient Temperature	T <sub>A</sub>	-40		125	°C

## 4. Thermal Information

Parameters	Symbol	SOP8(300mil)	SOP16(300mil)	Unit
Junction-to-ambient thermal resistance	R <sub>θJA</sub>	86	82	°C/W

Parameters	Symbol	SOP8(300mil)	SOP16(300mil)	Unit
Junction-to-case (top) thermal resistance	$R_{\theta JC(\text{top})}$	28	42	°C/W
Junction-to-board thermal resistance	$R_{\theta JB}$	42	46	°C/W
Junction-to-top characterization parameter	$\Psi_{JT}$	4	12	°C/W
Junction-to-board characterization parameter	$\Psi_{JB}$	42	46	°C/W

## 5. Specifications

### 5.1. Electrical Characteristics: NSI1305M05

(AVDD = 3.0V ~ 5.5V, DVDD = 3.0V ~ 5.5V, INP = -50mV to +50mV, and INN = AGND = 0V,  $T_A$  = -40 °C to 125 °C and sinc<sup>3</sup> filter with OSR=256. Unless otherwise noted, Typical values are at CLKIN=20MHz, AVDD = 5V, DVDD = 3.3V,  $T_A$  = 25 °C)

Parameters	Symbol	Min	Typ	Max	Unit	Comments
<b>Power Supply</b>						
Analog Side Supply Voltage	AVDD	3.0	5.0	5.5	V	
Digital Side Supply Voltage	DVDD	3.0	3.3	5.5	V	
Analog Side Supply Current	IAVDD		11.5	15	mA	
Digital Side Supply Current	IDVDD		2	3	mA	
AVDD undervoltage detection threshold voltage	AVDD <sub>UV</sub>	2	2.5	3	V	AVDD falling
<b>Analog Input</b>						
Common-mode overvoltage detection level	$V_{CMov}$	0.9			V	Detection level has a typical hysteresis of 96 mV
Common-mode rejection ratio	CMRR <sub>dc</sub>		-95		dB	INP = INN, $f_{IN} = 0$ Hz, $V_{CM \min} \leq VIN \leq V_{CM \max}$
	CMRR <sub>ac</sub>		-99		dB	INP = INN, $f_{IN} = 10$ kHz, $V_{CM \min} \leq VIN \leq V_{CM \max}$
Single-ended input resistance	$R_{IN}$		4.75		kΩ	INN = AGND
Differential input resistance	$R_{IND}$		4.9		kΩ	
Input capacitance	$C_I$		2		pF	
Input bias current	$I_{IB}$	-24	-23	-20	µA	INP = INN = AGND, $I_{IB} = (I_{IBP} + I_{IBN}) / 2$
Input bias current drift	$TCl_{IB}$		±2		nA/°C	
Common-mode transient immunity	CMTI	100	150		kV/µs	Common-mode transient immunity
<b>DC Accuracy</b>						
Differential nonlinearity	DNL	-0.99		0.99	LSB	
Integral nonlinearity	INL	-4	±1	4	LSB	

Parameters	Symbol	Min	Typ	Max	Unit	Comments
Offset error	$E_o$	-50	$\pm 2.5$	50	$\mu V$	$INP = INN = AGND$
Offset error thermal drift	$TCE_o$	-0.5	$\pm 0.15$	0.5	$\mu V/^\circ C$	
Gain error	$E_G$	-0.2%	$\pm 0.005\%$	0.2%		
Gain error thermal drift	$TCE_G$	-30	$\pm 10$	30	$ppm/^\circ C$	
Power supply rejection ratio	PSRR		-106		dB	PSRR vs AVDD, at DC
			-104		dB	PSRR vs AVDD, 100mV and 10kHz ripple
<b>AC Accuracy</b>						
Signal to noise ratio	SNR	78	82.5		dB	$f_{IN} = 1kHz$
Signal to noise and distortion	SINAD	77	82.5		dB	$f_{IN} = 1kHz$
Total harmonic distortion	THD		-96	-84	dB	$f_{IN} = 1kHz$
Spurious-free dynamic range	SFDR	86	100		dB	$f_{IN} = 1kHz$
<b>Digital Input / Output</b>						
Input current	$I_{IN}$	-9		1	$\mu A$	$DGND \leq V_{IN} \leq DVDD$
Input capacitance	$C_{IN}$		5		pF	
Output load capacitance	$C_{LOAD}$		30		pF	
High-level input voltage	$V_{IH}$	$0.7 \times DVDD$		$DVDD + 0.3$	V	
Low-level input voltage	$V_{IL}$	-0.3		$0.3 \times DVDD$	V	
High-level output voltage	$V_{OH}$	DVDD-0.1			V	$I_{OH} = -20\mu A$
		DVDD-0.4			V	$I_{OH} = -4mA$
Low-level output voltage	$V_{OL}$			0.1	V	$I_{OL} = 20\mu A$
				0.4	V	$I_{OL} = 4mA$
<b>Timing</b>						
CLKIN clock frequency	$f_{CLKIN}$	5		21	MHz	
CLKIN clock duty ratio	Duty	40	50	60	%	
DOUT rising time	$t_r$			5	ns	$C_{LOAD} = 15pF$
DOUT falling time	$t_f$			5	ns	$C_{LOAD} = 15pF$
DOUT hold time after rising edge of CLKIN	$t_H$	3.5			ns	$C_{LOAD} = 15pF$
Rising edge of CLKIN to DOUT valid delay	$t_D$			15	ns	$C_{LOAD} = 15pF$
Analog setting time	$t_{AS}$		0.5		ms	AVDD step to 3.0 V with $DVDD \geq 3.0 V$ , to DOUT valid, 0.1% settling

## 5.2. Electrical Characteristics: NSI1305M25

(AVDD = 3.0V ~ 5.5V, DVDD = 3.0V ~ 5.5V, INP = -250mV to +250mV, and INN = AGND = 0V, TA = -40°C to 125°C and sinc<sup>3</sup> filter with OSR=256. Unless otherwise noted, Typical values are at CLKIN=20MHz, AVDD = 5V, DVDD = 3.3V, TA= 25°C)

Parameters	Symbol	Min	Typ	Max	Unit	Comments
<b>Power Supply</b>						
Analog Side Supply Voltage	AVDD	3.0	5.0	5.5	V	
Digital Side Supply Voltage	DVDD	3.0	3.3	5.5	V	
Analog Side Supply Current	IAVDD		11.4	15.1	mA	
Digital Side Supply Current	IDVDD		1.78	3	mA	
AVDD undervoltage detection threshold voltage	AVDD <sub>UV</sub>	1.8	2.3	2.7	V	AVDD falling
<b>Analog Input</b>						
Common-mode overvoltage detection level	V <sub>CMov</sub>	0.9			V	Detection level has a typical hysteresis of 96 mV
Common-mode rejection ratio	CMRR <sub>dc</sub>		-106		dB	INP = INN, f <sub>IN</sub> = 0 Hz, V <sub>CM min</sub> ≤ VIN ≤ V <sub>CM max</sub>
	CMRR <sub>ac</sub>		-104		dB	INP = INN, f <sub>IN</sub> = 10 kHz, V <sub>CM min</sub> ≤ VIN ≤ V <sub>CM max</sub>
Single-ended input resistance	R <sub>IN</sub>		19		kΩ	INN = AGND
Differential input resistance	R <sub>IND</sub>		22		kΩ	
Input capacitance	C <sub>I</sub>		2		pF	
Input bias current	I <sub>IB</sub>	-24	-18	-12	μA	INP = INN = AGND, I <sub>IB</sub> = (I <sub>IBP</sub> + I <sub>IBN</sub> ) / 2
Input bias current drift	TCl <sub>IB</sub>		±2		nA/°C	
Common-mode transient immunity	CMTI	100	150		kV/μs	Common-mode transient immunity
<b>DC Accuracy</b>						
Differential nonlinearity	DNL	-0.99		0.99	LSB	
Integral nonlinearity	INL	-4	±1	4	LSB	
Offset error	E <sub>O</sub>	-100	±4.5	100	μV	INP = INN = AGND
Offset error thermal drift	TCE <sub>O</sub>	-1	±0.15	1	μV/°C	
Gain error	E <sub>G</sub>	-0.2%	±0.005%	0.2%		
Gain error thermal drift	TCE <sub>G</sub>	-40	±20	40	ppm/°C	
Power supply rejection ratio	PSRR		-100		dB	PSRR vs AVDD, at DC
			-90		dB	PSRR vs AVDD, 100mV and 10kHz ripple

Parameters	Symbol	Min	Typ	Max	Unit	Comments
<b>AC Accuracy</b>						
Signal to noise ratio	SNR	82	86		dB	$f_{IN} = 1\text{kHz}$
Signal to noise and distortion	SINAD	80	84		dB	$f_{IN} = 1\text{kHz}$
Total harmonic distortion	THD		-90	-80	dB	$f_{IN} = 1\text{kHz}$
Spurious-free dynamic range	SFDR	82	92		dB	$f_{IN} = 1\text{kHz}$
<b>Digital Input / Output</b>						
Input current	$I_{IN}$	-9		1	$\mu\text{A}$	$DGND \leq V_{IN} \leq DVDD$
Input capacitance	$C_{IN}$		5		pF	
Output load capacitance	$C_{LOAD}$		30		pF	
High-level input voltage	$V_{IH}$	$0.7 \times DVDD$		$DVDD+0.3$	V	
Low-level input voltage	$V_{IL}$	-0.3		$0.3 \times DVDD$	V	
High-level output voltage	$V_{OH}$	DVDD-0.1			V	$I_{OH} = -20\mu\text{A}$
		DVDD-0.4			V	$I_{OH} = -4\text{mA}$
Low-level output voltage	$V_{OL}$			0.1	V	$I_{OL} = 20\mu\text{A}$
				0.4	V	$I_{OL} = 4\text{mA}$
<b>Timing</b>						
CLKIN clock frequency	$f_{CLKIN}$	5		21	MHz	
CLKIN clock duty ratio	Duty	40	50	60	%	
DOUT rising time	$t_r$			5	ns	$C_{LOAD} = 15\text{pF}$
DOUT falling time	$t_f$			5	ns	$C_{LOAD} = 15\text{pF}$
DOUT hold time after rising edge of CLKIN	$t_h$	3.5			ns	$C_{LOAD} = 15\text{pF}$
Rising edge of CLKIN to DOUT valid delay	$t_d$			15	ns	$C_{LOAD} = 15\text{pF}$
Analog setting time	$t_{AS}$		0.5		ms	AVDD step to 3.0 V with $DVDD \geq 3.0\text{ V}$ , to DOUT valid, 0.1% settling

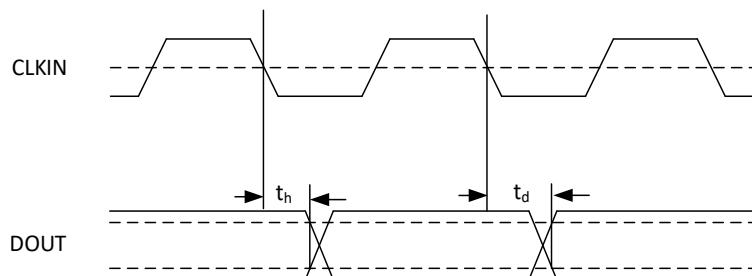


Figure 5.1 Digital Input and Output Timing

### 5.3. Typical Performance Characteristics

Unless otherwise noted, test at AVDD = 5V, DVDD = 3.3V, Vin = -250mV to 250mV (NSI1305M25) or -50mV to 50mV (NSI1305M05), CLKIN=20MHz, and sinc<sup>3</sup> filter with OSR=256.

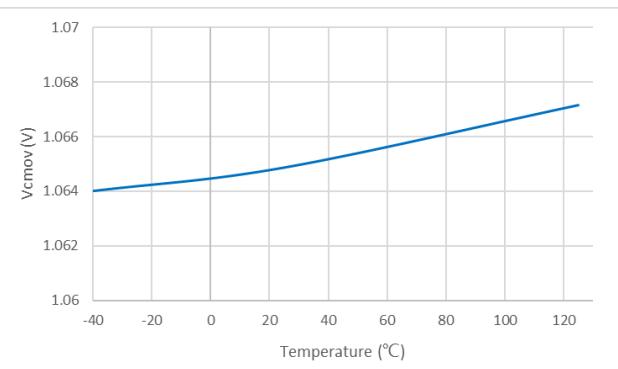


Figure 5.2 Common-Mode Overvoltage Detection Level vs Temperature

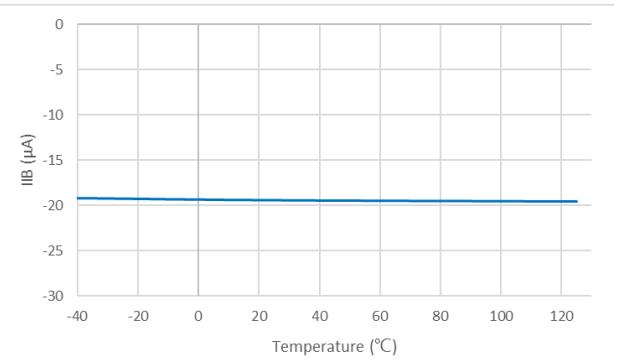


Figure 5.3 Input Bias Current vs Temperature

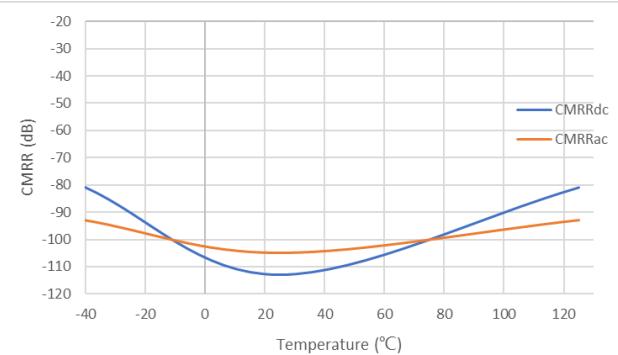


Figure 5.4 Common-Mode Rejection Ratio vs Temperature

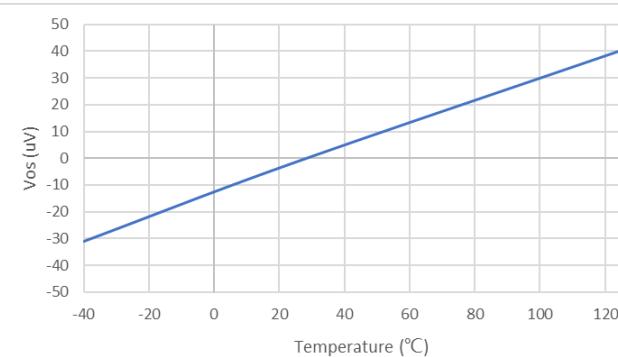


Figure 5.5 Input Offset Voltage vs Temperature

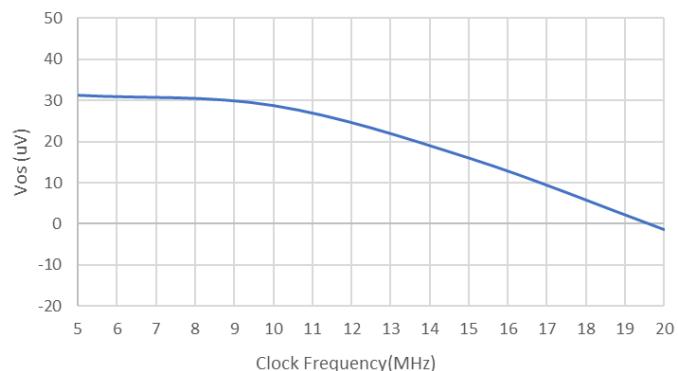


Figure 5.6 Input Offset Voltage vs Clock Frequency

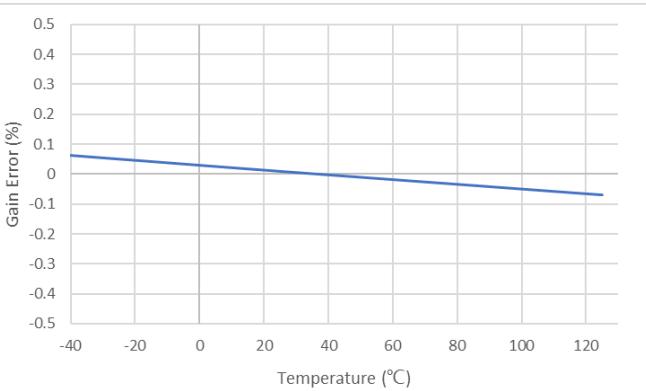


Figure 5.7 Gain Error vs Temperature

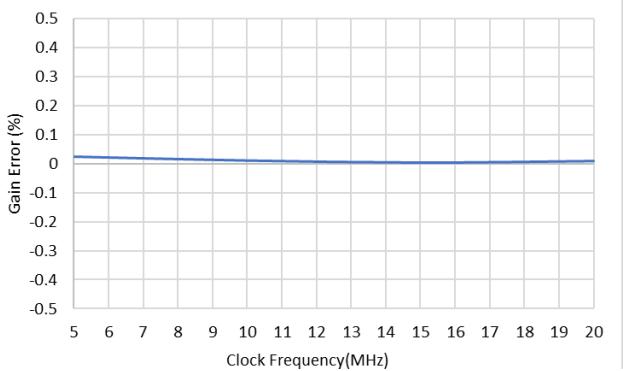


Figure 5.8 Gain Error vs Clock Frequency

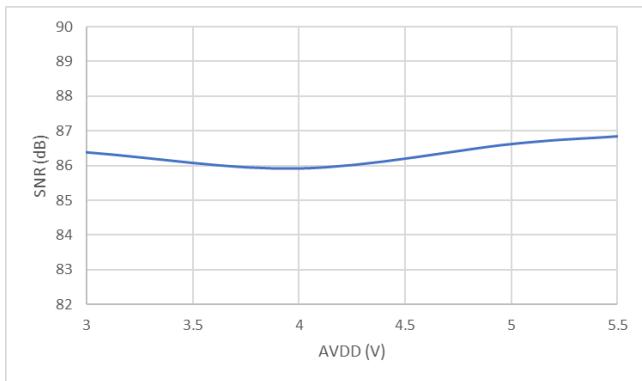


Figure 5.9 Signal-to-Noise Ratio vs Analog Side Supply Voltage

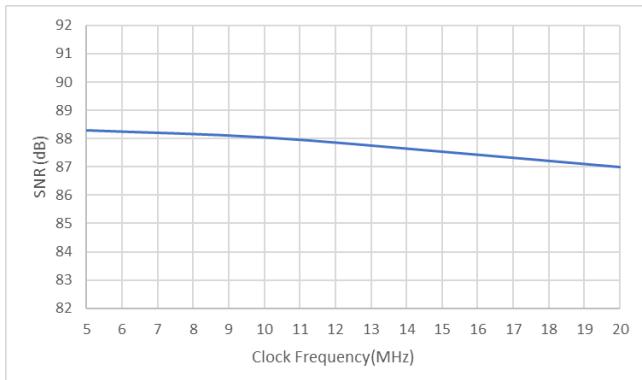


Figure 5.10 Signal-to-Noise Ratio vs Clock Frequency

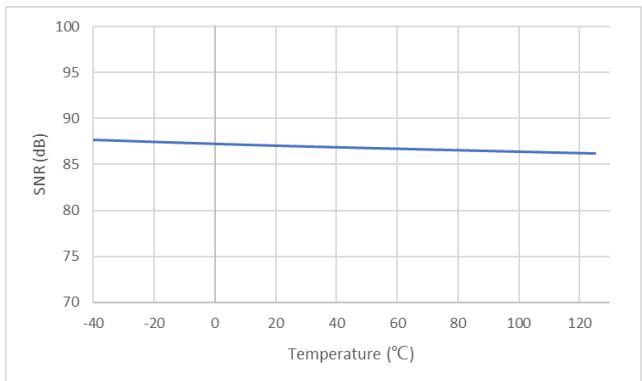


Figure 5.11 Signal-to-Noise Ratio vs Temperature

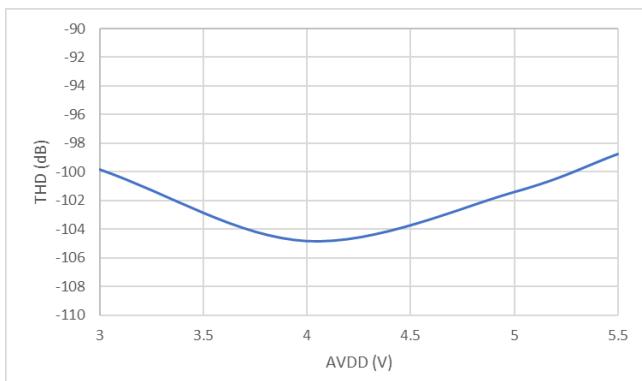


Figure 5.12 Total Harmonic Distortion vs Analog Side Supply Voltage

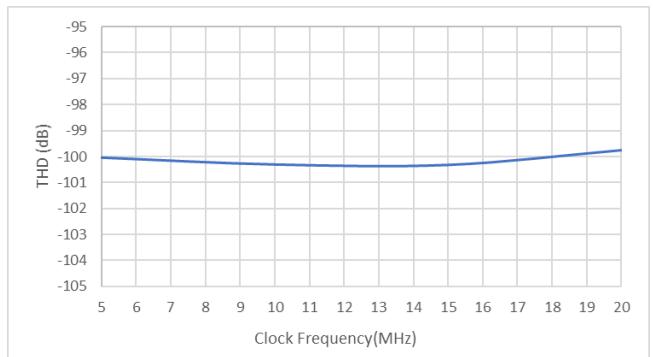


Figure 5.13 Total Harmonic Distortion vs Clock Frequency

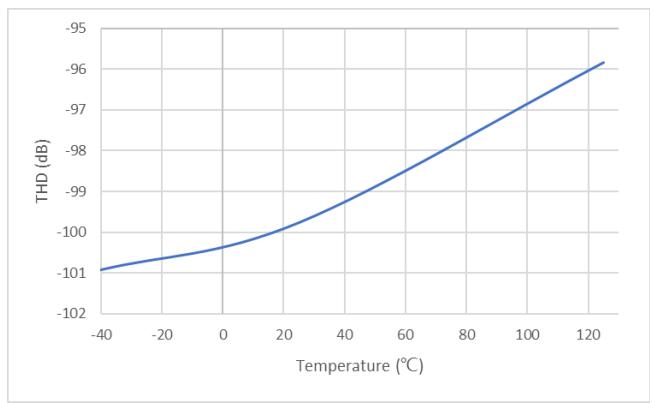


Figure 5.14 Total Harmonic Distortion vs Temperature

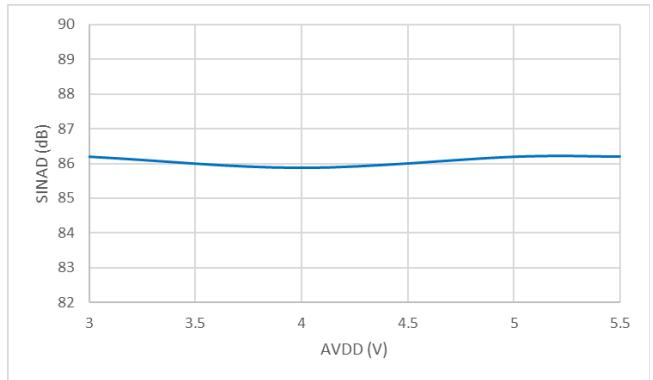


Figure 5.15 Signal-to-Noise + Distortion vs Analog Side Supply Voltage

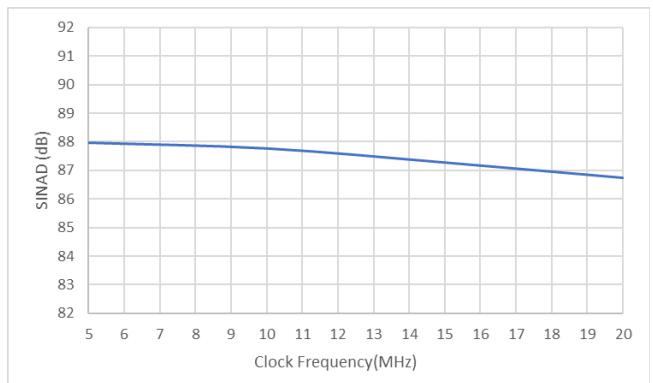


Figure 5.16 Signal-to-Noise + Distortion vs Clock Frequency

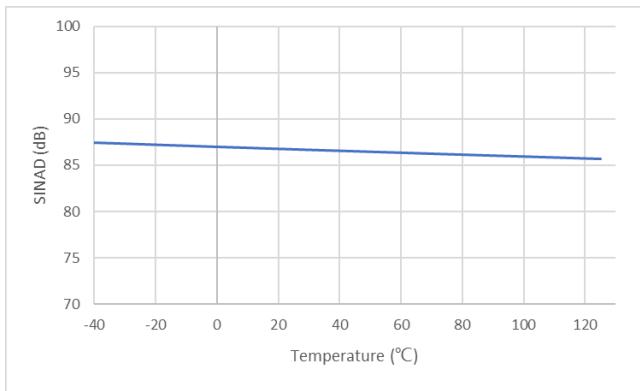


Figure 5.17 Signal-to-Noise + Distortion vs Temperature

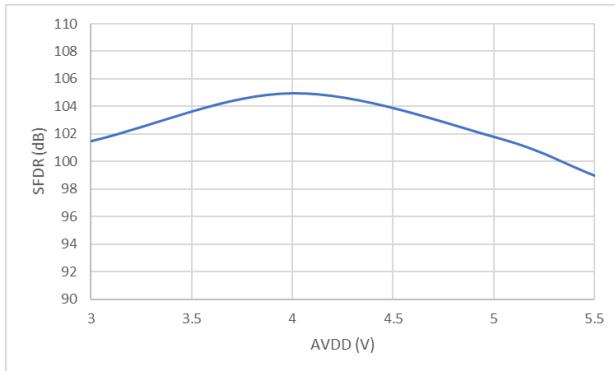


Figure 5.18 Spurious-Free Dynamic Range vs Analog Side Supply Voltage

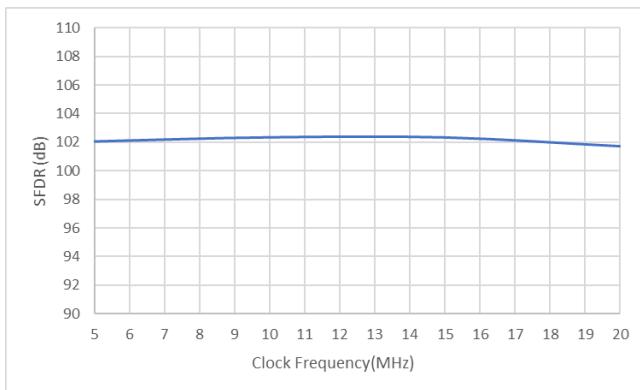


Figure 5.19 Spurious-Free Dynamic Range vs Clock Frequency

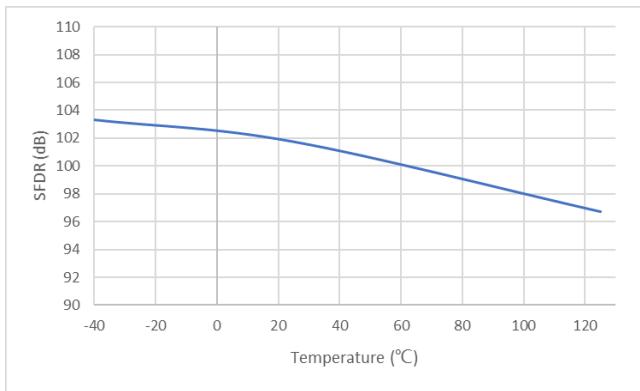


Figure 5.20 Spurious-Free Dynamic Range vs Temperature

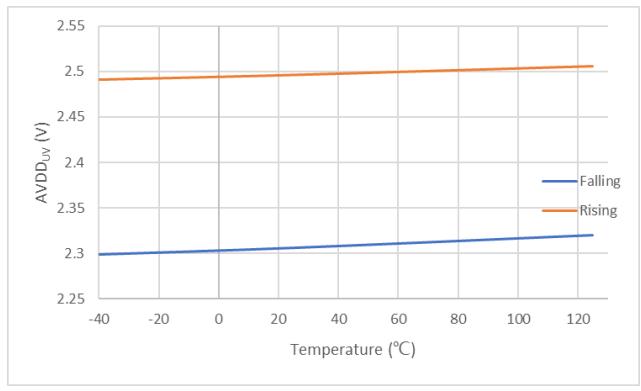


Figure 5.21 Analog Side Under-Voltage Detection Level vs Temperature

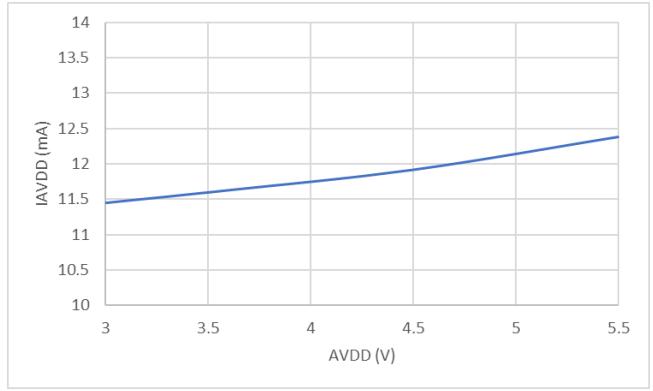


Figure 5.22 Analog Side Supply Current vs Supply Voltage

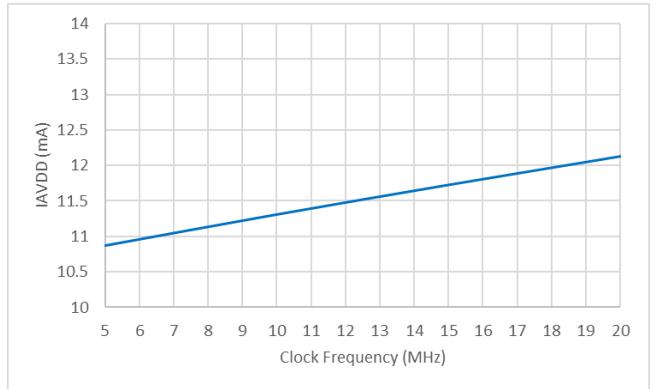


Figure 5.23 Analog Side Supply Current vs Clock Frequency

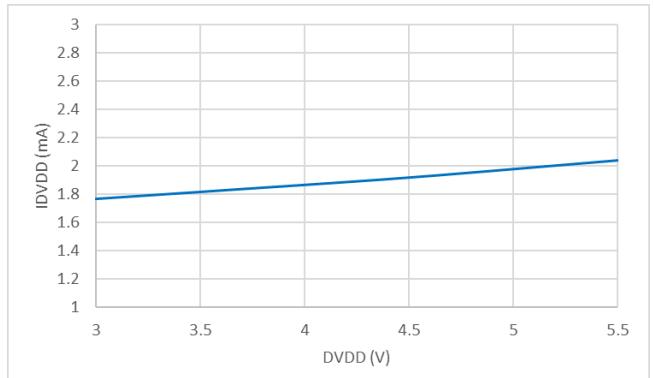


Figure 5.24 Digital Side Supply Current vs Supply Voltage

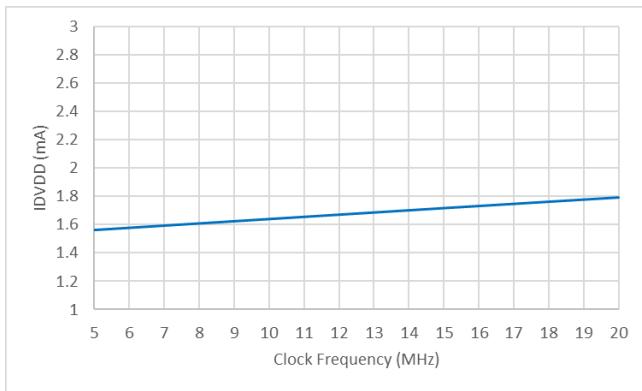


Figure 5.25 Digital Side Supply Current vs Clock Frequency

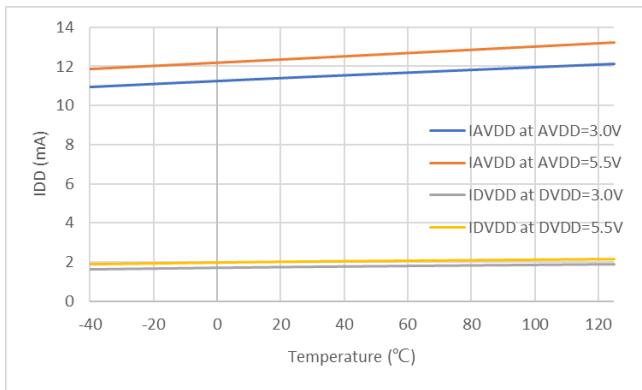


Figure 5.26 Supply Current vs Temperature

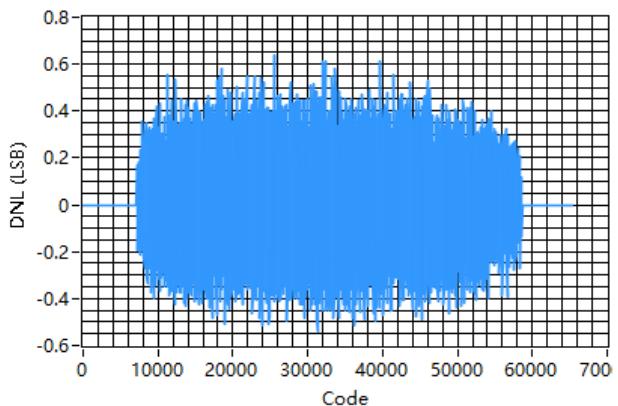


Figure 5.27 Typical Differential Nonlinearity

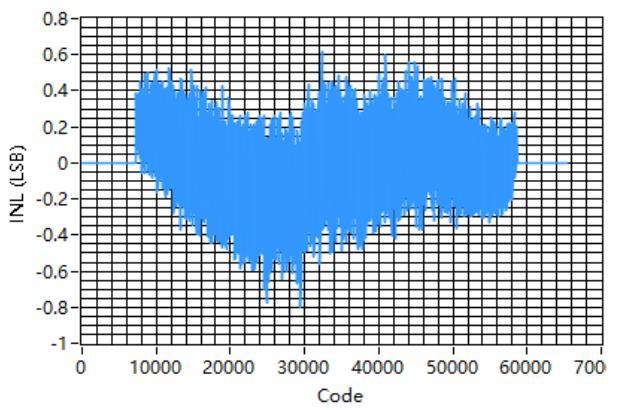


Figure 5.28 Typical Integral Nonlinearity

## 6. High Voltage Feature Description

### 6.1. Insulation and Safety Related Specifications

Parameters	Symbol	Value	Unit	Comments
Minimum External Air Gap (Clearance)	CLR	8	mm	Shortest terminal-to-terminal distance through air
Minimum External Tracking (Creepage)	CPG	8	mm	Shortest terminal-to-terminal distance across the package surface
Minimum internal gap	DTI	28	µm	Distance through insulation
Tracking Resistance (Comparative Tracking Index)	CTI	>600	V	DIN EN 60112 (VDE 0303-11); IEC 60112
Material Group		I		IEC 60664-1

### 6.2. Insulation Characteristics

Description	Test Condition	Symbol	Value	Unit
<b>DIN VDE 0110</b>			SOW8	SOW16
Installation Classification per DIN VDE 0110				
For Rated Mains Voltage $\leq$ 150Vrms			I to IV	I to IV
For Rated Mains Voltage $\leq$ 300Vrms			I to IV	I to IV
For Rated Mains Voltage $\leq$ 400Vrms			I to IV	I to IV
Climatic Classification			40/125/21	40/125/21
Pollution Degree per DIN VDE 0110, Table 1			2	2
Maximum repetitive isolation voltage		$V_{IORM}$	2121	$V_{PEAK}$
Maximum working isolation voltage	AC Voltage	$V_{IOWM}$	1500	$V_{RMS}$
	DC Voltage		2121	$V_{DC}$
Input to Output Test Voltage, Method B1	$V_{IORM} \times 1.875 = V_{pd(m)}$ , 100% production test, $t_{ini} = t_m = 1$ sec, partial discharge < 5 pC	$V_{pd(m)}$	3977	$V_{PEAK}$
Input to Output Test Voltage, Method A				
After Environmental Tests Subgroup 1	$V_{IORM} \times 1.6 = V_{pd(m)}$ , $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge < 5 pC	$V_{pd(m)}$	3394	$V_{PEAK}$
After Input and /or Safety Test Subgroup 2 and Subgroup 3	$V_{IORM} \times 1.2 = V_{pd(m)}$ , $t_{ini} = 60$ sec, $t_m = 10$ sec, partial discharge < 5 pC	$V_{pd(m)}$	2545	$V_{PEAK}$
Maximum transient isolation voltage	$t = 60$ sec	$V_{IOTM}$	7000	$V_{PEAK}$

Description	Test Condition	Symbol	Value	Unit	
Maximum Surge Isolation Voltage	Test method per IEC60065,1.2/50us waveform, $V_{TEST}=V_{IOSM} \times 1.6$	$V_{IOSM}$	6250	6250 $V_{PEAK}$	
Isolation resistance	$V_{IO} = 500V, T_{amb}=T_s$	$R_{IO}$	$>10^9$	$>10^9$ $\Omega$	
	$V_{IO} = 500V, 100^\circ C \leq T_{amb} \leq 125^\circ C$	$R_{IO}$	$>10^{11}$	$>10^{11}$ $\Omega$	
Isolation capacitance	$f = 1MHz$	$C_{IO}$	0.8	0.8 pF	
Safety input, output, or total power	$V_I = 5.5V, T_J = 150^\circ C, T_A = 25^\circ C$	$P_S$	1430	1524 mW	
Safety input, output, or supply current	$\theta_{JA} = 86^\circ C/W$ for SOP8, $V_I = 5.5V, T_J = 150^\circ C, T_A = 25^\circ C$	$I_S$	260	/ mA	
	$\theta_{JA} = 82^\circ C/W$ for SOP16, $V_I = 5.5V, T_J = 150^\circ C, T_A = 25^\circ C$		/	277 mA	
Maximum safety temperature		$T_s$	150	150 $^\circ C$	
<b>UL1577</b>					
Insulation voltage per UL	$V_{TEST} = V_{ISO}, t = 60 s$ (qualification), $V_{TEST} = 1.2 \times V_{ISO}, t = 1 s$ (100% production test)	$V_{ISO}$	5000	5000	$V_{RMS}$

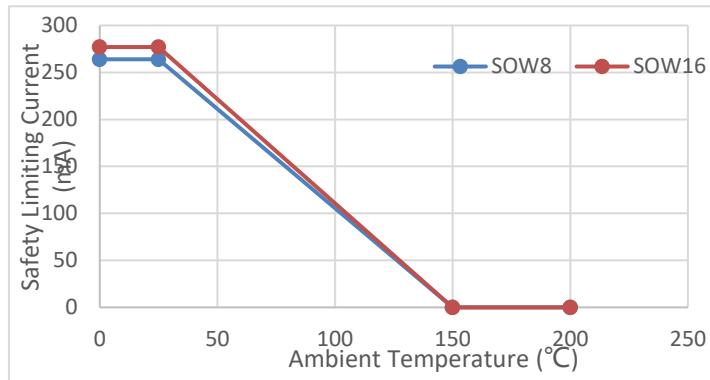


Figure 6.1 NSi1305 Thermal Derating Curve, Dependence of Safety Limiting Values with Case Temperature per DIN VDE V 0884-11

### 6.3. Regulatory Information

The NSi1305 are approved or pending approval by the organizations listed in table.

UL	VDE	CQC
UL 1577 Component Recognition Program	Approved under CSA Component Acceptance Notice 5A	DIN VDE V 0884-11(VDE V 0884-11):2017-01 Certified by CQC11-471543-2012 GB4943.1-2011
Single Protection, 5000V <sub>rms</sub> Isolation voltage	Single Protection, 5000V <sub>rms</sub> Isolation voltage	Reinforce Insulation 2121Vpeak, $V_{IOSM}=6250V_{peak}$ Reinforced insulation
Certificate No.E500602	Certificate No.E500602	File pending CQC20001264939

## 7. Function Description

### 7.1. Overview

The NSI1305 is a high performance isolated modulator that accept fully-differential input. The fully-differential input is ideally suited to shunt current monitoring in high voltage applications where isolation is required. The analog input is continuously sampled by a second-order Σ-Δ modulator in the device, which is driven by a pre-stage fully-differential amplifier in the device. With the internal voltage reference and clock generator, the modulator convert the analog input signal to a digital bitstream. The drivers (called TX in the Functional Block Diagram) transfer the output of the modulator across the isolation barrier that separate external clock, as shown in the Functional Block Diagram.

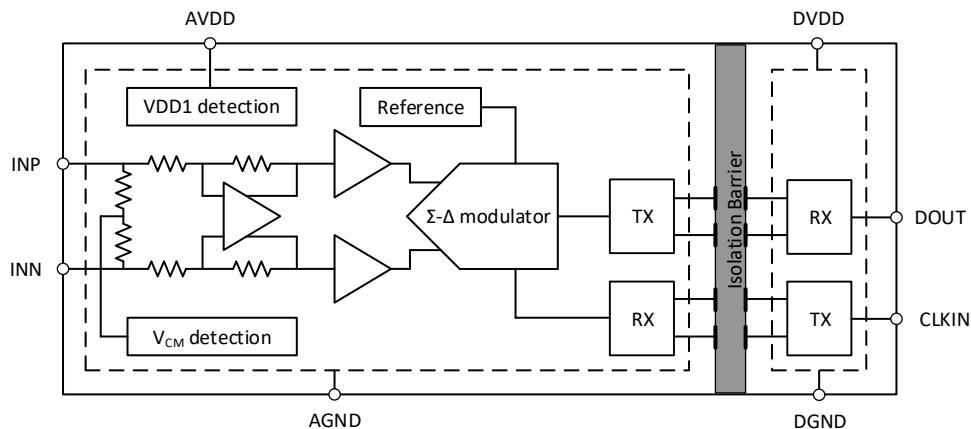


Figure 7.1 Function Block Diagram

### 7.2. Analog Input

There are two restrictions on the analog input signals (VINP and VINN).

- If the input voltage exceeds the range AGND – 6 V to AVDD + 0.5 V, the input current must be limited to 10 mA because the device input electrostatic discharge (ESD) diodes turn on.
- The linearity and noise performance of the device are ensured only when the analog input voltage remains within the specified linear full-scale range (FSR) and within the specified common-mode input voltage range.

### 7.3. Digital Input

The digital input refers to clock signal which provides the clock for modulator conversion and output data frame clock. The clock signal should be supplied by an external source with a frequency range from 5MHz to 21MHz.

### 7.4. Digital Output

The digital output provides a stream of ones and zeros that can accurately represents the analog input voltage. Within the linear input range, the density of ones in the bitstream is proportional to the input voltage.

Ideally for a 0V input signal, the modulator outputs a bitstream with 50% high time. For a 250mV input signal (for the NSI1305M25), the modulator outputs a bitstream with 89.06% high time. For a -250mV input signal (for the NSI1305M25), the modulator outputs a bitstream with 10.94% high time.

If the input signal is greater than or equal to 320mV (64 mV for the NSI1305x05), the modulator clips with a steam of all ones. If the input signal is less than or equal to -320mV (-64 mV for the NSI1305x05), the output of the modulator clips with a stream of all zeros. In this case, however, the NSI1305 generates a single 0 (if the input is at positive full-scale) or 1 every 128 clock cycles to indicate proper device function (see section 7.5 for more details).

Due to quantization noise of internal Σ-Δ modulator, the digital output frequency is not fixed, but the density of high time is fixed based on input voltage. Typical modulator digital output characteristics are as below.

Table 7.1 Input voltage with ideal corresponding density of high time at modulator data output, and ADC code

<i>Analog input</i>	<i>Input voltage</i>	<i>Density of high time</i>	<i>Digital output code(16-bit unsigned decimation)</i>
+Differential input voltage before	+320mV	99.22%	65024

clipping output	(+64mV for NSI1305x05)		
+Linear differential input full scale voltage	+250mV (+50mV for NSI1305x05)	89.06%	58366
Zero	0mV	50%	32768

Ideal density of 1s in the output bitstream for any input voltage(except for full-scale input, as described in section 7.5) at modulator data output can be calculated with:

$$\frac{V_{IN} + V_{CLIPPING}}{2 \times V_{CLIPPING}}$$

Similarly, the ADC code can be calculated, assuming a 16-bit unsigned decimation filter.

## 7.5. Fail-safe Output

NSI1305 integrates some diagnostic measures and offers a fail-safe output to simplify system-level design. The fail-safe function will be activated in following conditions:

- When the undervoltage of AVDD is detected ( $AVDD < AVDD_{UV}$ ), DOUT pin output a bitstream of all logic zeros, as shown in Figure 7.2.
- When the overvoltage of common-mode input voltage is detected ( $V_{CM} > V_{CMov}$ ), DOUT pin output a bitstream of all logic ones, as shown in Figure 7.2.

NOTE: If both of the faults above occur at the same time, DOUT pin output a bitstream of all logic zeros. (AVDD missing has a higher priority).

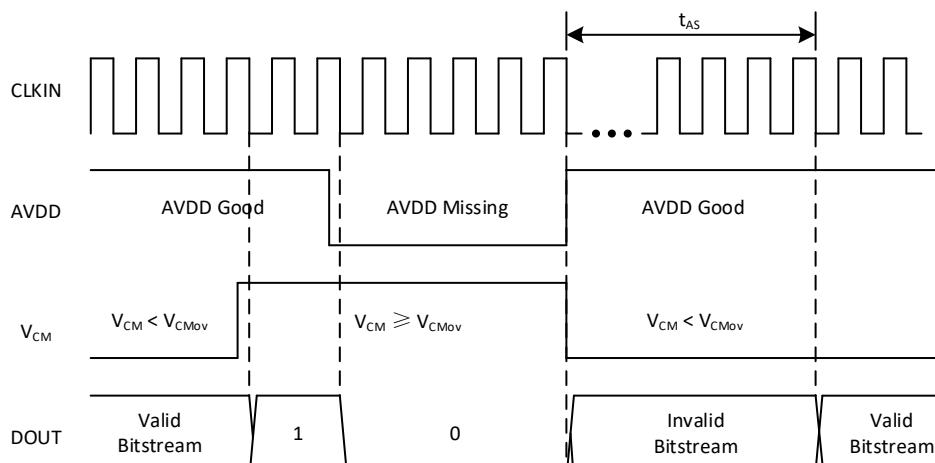


Figure 7.2 Fail-safe output

If an overrange input signal is applied to the NSI1305 ( $V_{IN} \geq V_{Clipping}$ ), the output generates a single 0 or 1 every 128 clock cycles, as shown in Figure 7.3.

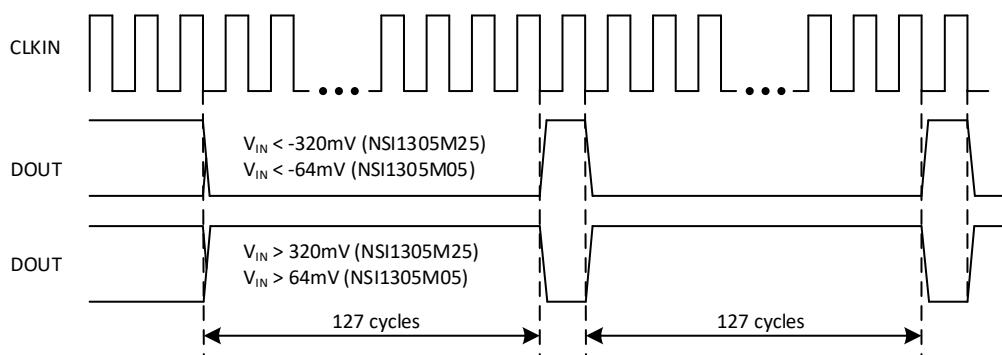


Figure 7.3 Overrange output

## 8. Application Note

### 8.1. Typical Application Circuit

NSI1305 is ideally suited to shunt resistor-based current sensing in high voltage applications such as frequency inverters. The typical application circuit is shown in Figure 8.1.

The voltage across the shunt resistor  $R_{sense}$  is applied to the differential input of NSI1305 through a RC filter. The internal second-order sigma-delta modulator converts the analog input to a single-bit output stream. The external digital system provides a clock source for the modulator and a digital filter for decimation and quantization noise filtering.

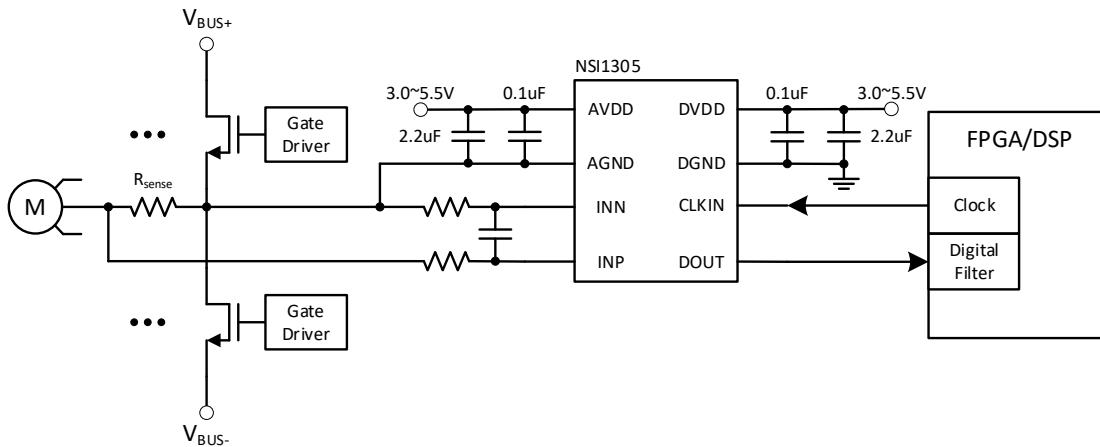


Figure 8.1 Typical application circuit in phase current sensing

### 8.2. Shunt Resistor Selection

Choosing a particular shunt resistor is usually a compromise between minimizing power dissipation and maximizing accuracy. Smaller sense resistor decreases power dissipation, while larger sense resistor can improve measure accuracy by utilizing the full input range of isolated amplifier.

There are two other factors should be considered when selecting the shunt resistor:

- The voltage-drop caused by the rated current range must not exceed the recommended linear input voltage range:  $V_{SHUNT} \leq FSR$ .
- The voltage-drop caused by the maximum allowed overcurrent must not exceed the input voltage that causes a clipping output:  $V_{SHUNT} \leq V_{Clipping}$ .

### 8.3. Digital Filter

The  $\Sigma\Delta$  modulator a characteristics of noise shaping. Most of the quantization noise is pushed from a low frequency to a higher frequency.

In order to reduce higher-frequency quantization noise, the modulator output is fed to the digital low-pass filter. Subsequently, the signal of interest passes through to the output of the digital filter, while much of the higher-frequency quantization noise is filtered out.

The digital filter serves another function – decimation. It creates a digital output code from the bitstream that the modulator outputs. The ratio of the modulator rate ( $f_{MOD}$ ) of the delta-sigma modulator to its output data rate ( $f_{DR}$ ) is the oversampling ratio (OSR). The relationship between  $f_{DR}$  and  $f_{MOD}$  is:

$$f_{DR} = f_{MOD} / OSR \quad \text{Equation 8.1}$$

A sinc3 filter is recommended since it's simple and requires less hardware resources. Equation 8.2 describes the transfer function of a sinc filter.

$$H(Z) = \left( \frac{1}{DR} \frac{(1-Z^{-DR})}{(1-Z^{-1})} \right)^N \quad \text{Equation 8.2}$$

where:

DR is the decimation rate;

N is the sinc filter order.

The filter can be implemented in an FPGA or DSP. The sinc filter creates a digital output code by taking a multi-order moving average of the modulator output over a certain number of modulator clock periods.

The higher the decimation rate, the higher the conversion accuracy, and the lower the output data rate. So, there is a trade-off between accuracy and data rate. All the characterization in this datasheet is tested with a sinc3 filter with an oversampling ratio (OSR) of 256.

The output data size is expressed in Equation 8.3. The 16 most significant bits are used to return a 16-bit result.

$$\text{Data Size} = N \times \log_2 DR \quad (\text{Equation 8.3})$$

The filter characteristics for a third-order sinc filter are summarized in Table 8.2.

Table 8.2 Sinc3 Filter Characteristics for 20 MHz CLKIN

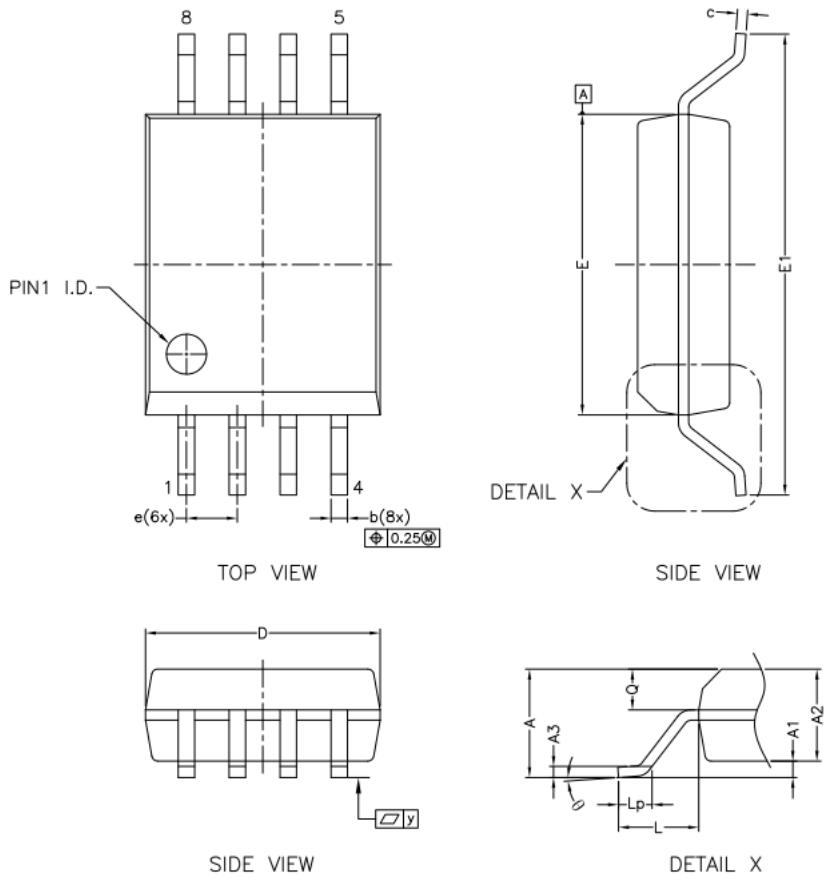
<b>Decimation Rate (DR)</b>	<b>Data Output Rate (kHz)</b>	<b>Data Size (bits)</b>	<b>Filter Response (kHz)</b>
32	625	15	163.7
64	312.5	18	81.8
128	156.2	21	40.9
256	78.1	24	20.4
512	39.1	27	10.2

#### 8.4. PCB Layout

There are some key guidelines or considerations for optimizing performance in PCB layout:

- NSI1305 requires a 0.1 $\mu$ F bypass capacitor between AVDD and AGND, DVDD and DGND. The capacitor should be placed as close as possible to the VDD pin. If better filtering is required, an additional 1~10 $\mu$ F capacitor may be used.
- Kelvin rules is recommended for the connection between shunt resistor to NSI1305. Because of the Kelvin connection, any voltage drops across the trace and leads should have no impact on the measured voltage.
- Place the shunt resistor close to the INP and INN inputs and keep the layout of both connections symmetrical and run very close to each other to the input of the NSI1305. This minimizes the loop area of the connection and reduces the possibility of stray magnetic fields from interfering with the measured signal.

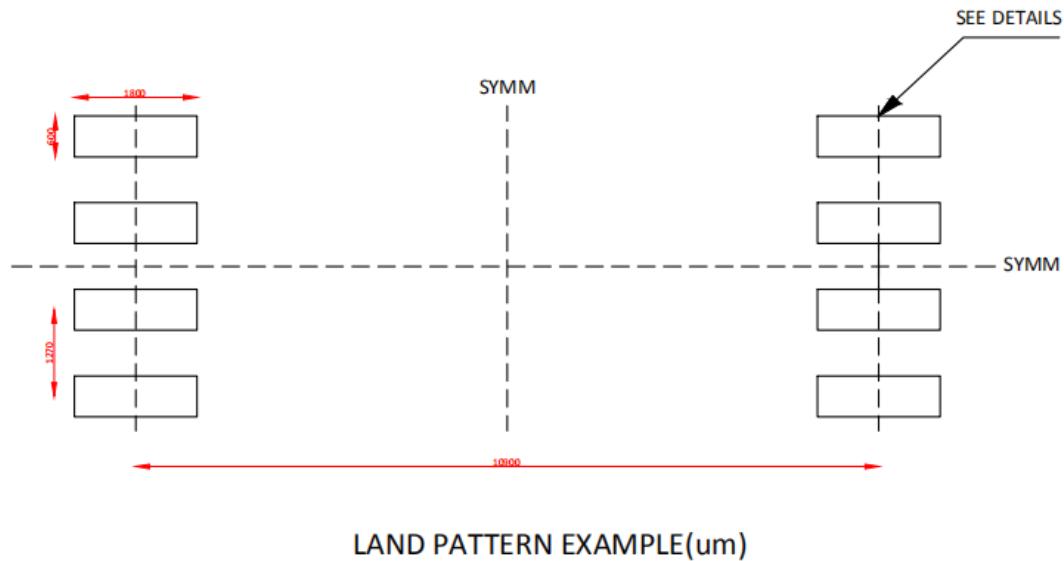
## 9. Package Information



\* CONTROLLING DIMENSION : MM

SYMBOL	MM		
	MIN.	NOM.	MAX.
A	--	--	2.80
A1	0.36	--	0.46
A2	2.20	2.30	2.40
A3	--	0.25	--
Q	0.97	1.02	1.07
b	0.31	0.41	0.51
c	0.13	--	0.33
D	5.75	5.85	5.95
E	7.40	7.50	7.60
E1	11.25	11.50	11.75
e	1.27 bsc		
L	2.00 bsc		
L <sub>p</sub>	0.50	--	1.00
y	--	0.10	--
$\theta$	0°	--	8°

Figure 9.1 SOW8 Package Shape and Dimension in millimeters



LAND PATTERN EXAMPLE(um)



Figure 9.2 SOW8 Package Board Layout Example

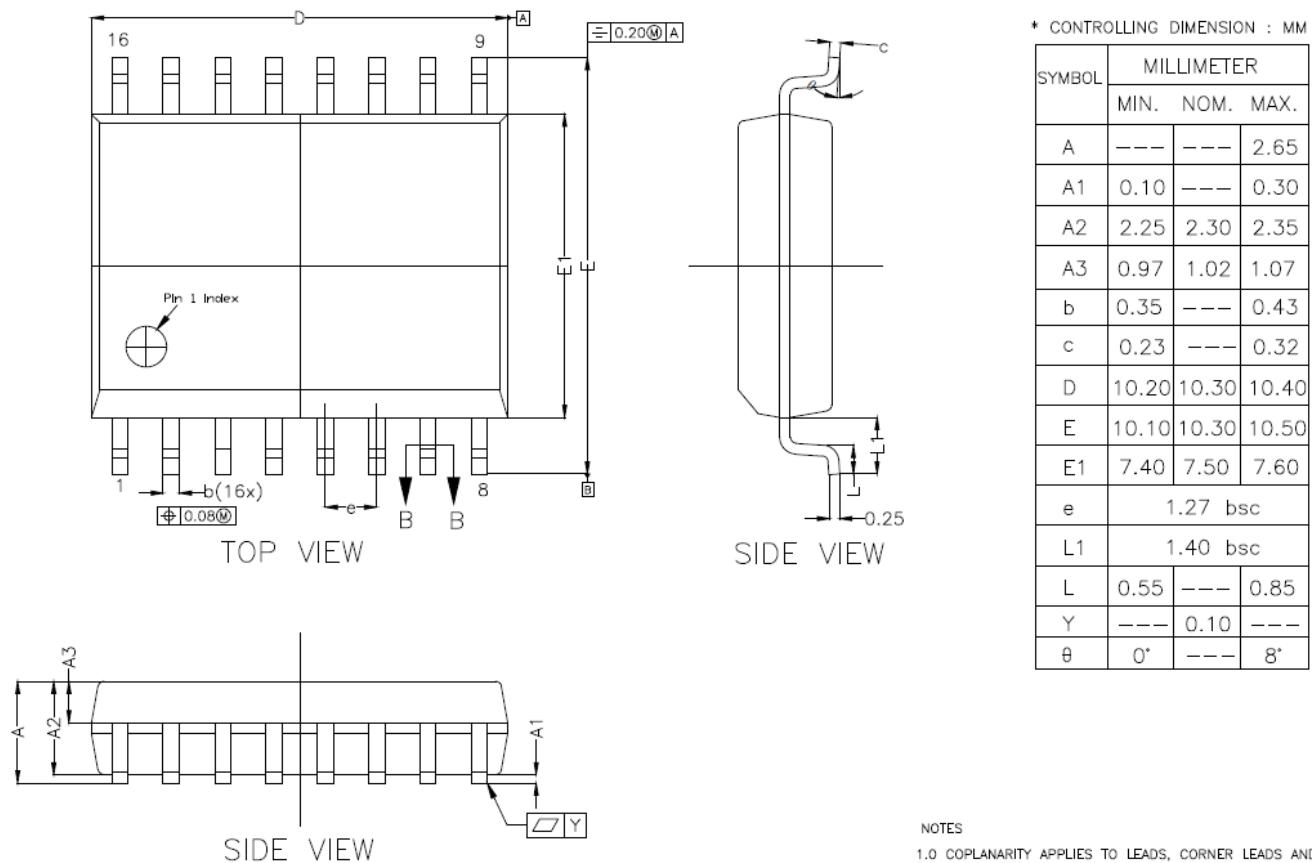


Figure 9.3 SOW16 package shape and dimension in millimeters

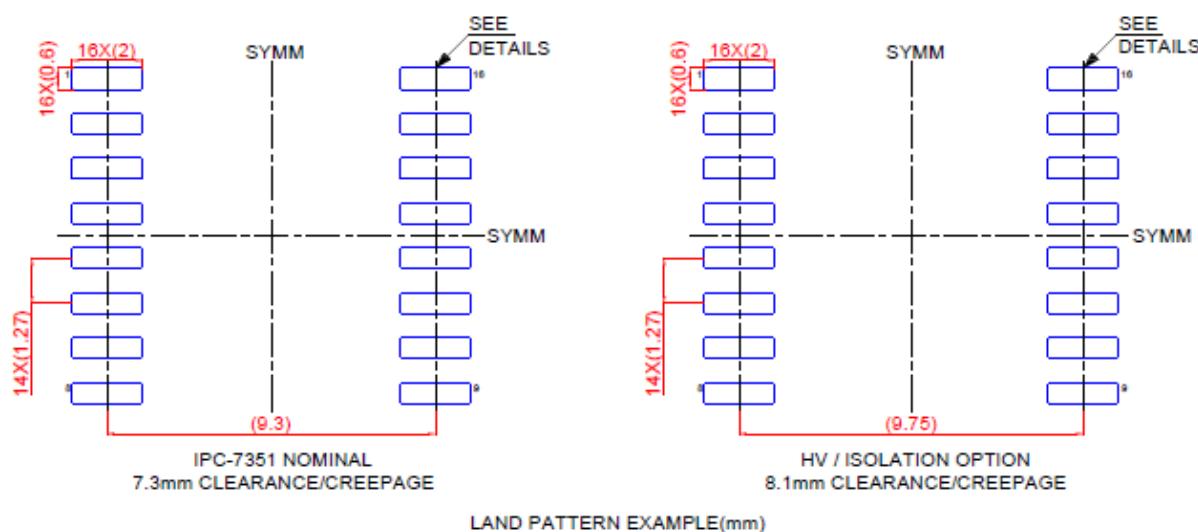


Figure 9.4 SOW16 Package Board Layout Example

## 10. Ordering Information

Part No.	Isolation Rating(kV)	Linear Input Range(mV)	Moisture Sensitivity Level	Temperature	Automotive	Package Type	Package Drawing	SPQ
NSI1305M05 -DSWVR	5	-50 ~ 50	Level-3	-40 to 125°C	NO	SOP8 (300mil)	SOW8	1000
NSI1305M25 -DSWVR	5	-250 ~ 250	Level-3	-40 to 125°C	NO	SOP8 (300mil)	SOW8	1000
NSI1305M05 -DSWR	5	-50 ~ 50	Level-2	-40 to 125°C	NO	SOP16 (300mil)	SOW16	1000
NSI1305M25 -DSWR	5	-250 ~ 250	Level-2	-40 to 125°C	NO	SOP16 (300mil)	SOW16	1000
NSI1305M05 -Q1SWVR	5	-50 ~ 50	Level-3	-40 to 125°C	Yes	SOP8 (300mil)	SOW8	1000
NSI1305M25 -Q1SWVR	5	-250 ~ 250	Level-3	-40 to 125°C	Yes	SOP8 (300mil)	SOW8	1000
NSI1305M05 -Q1SWR	5	-50 ~ 50	Level-2	-40 to 125°C	Yes	SOP16 (300mil)	SOW16	1000
NSI1305M25 -Q1SWR	5	-250 ~ 250	Level-2	-40 to 125°C	Yes	SOP16 (300mil)	SOW16	1000
Note: All packages are RoHS-compliant with peak reflow temperatures of 260°C according to the JEDEC industry standard classifications and peak solder temperatures.								

## 11. Documentation Support

Part Number	Product Folder	Datasheet	Technical Documents	Isolator selection guide
NSi1305	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>

## 12. Tape and Reel Information

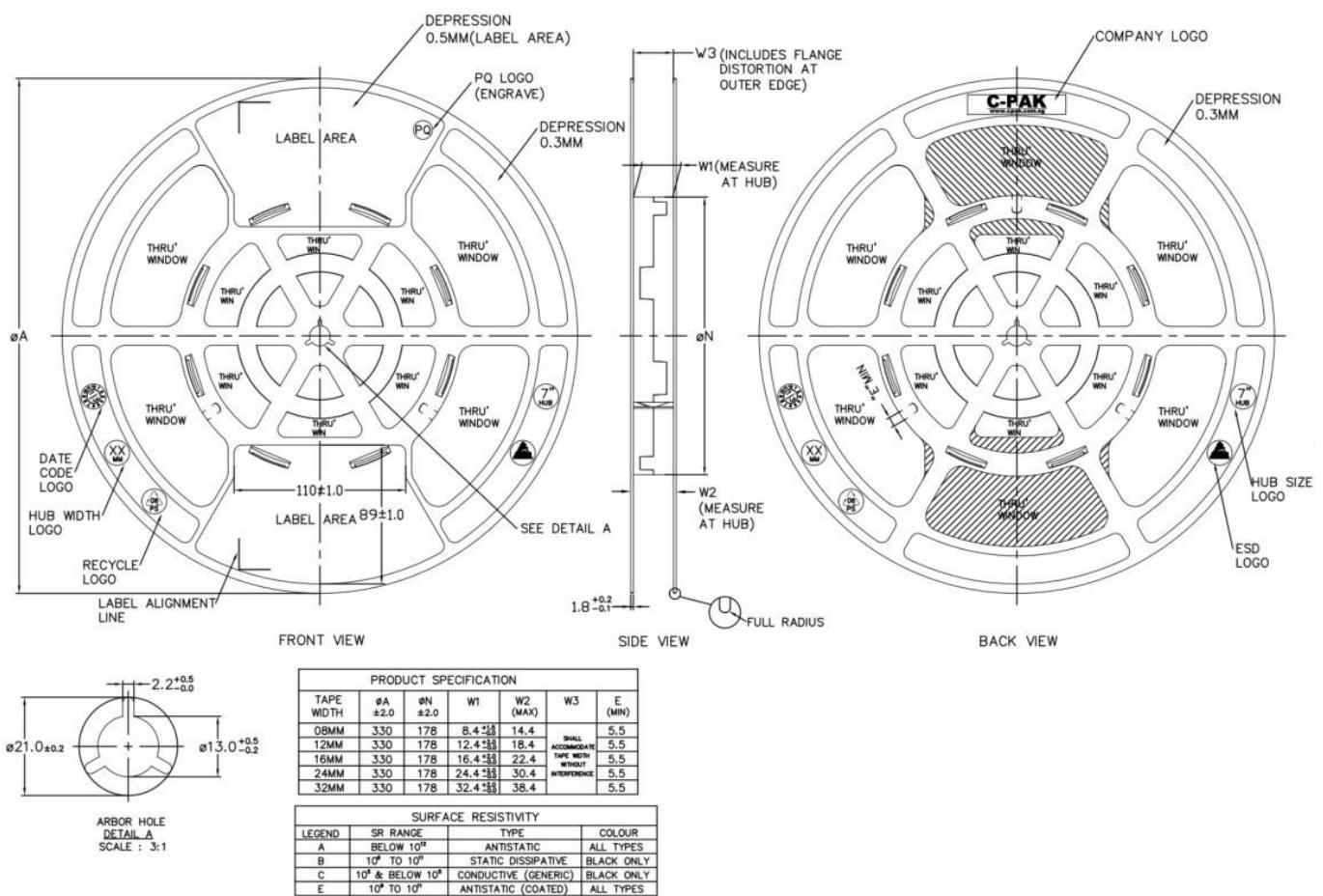


Figure 12.1 Tape Information

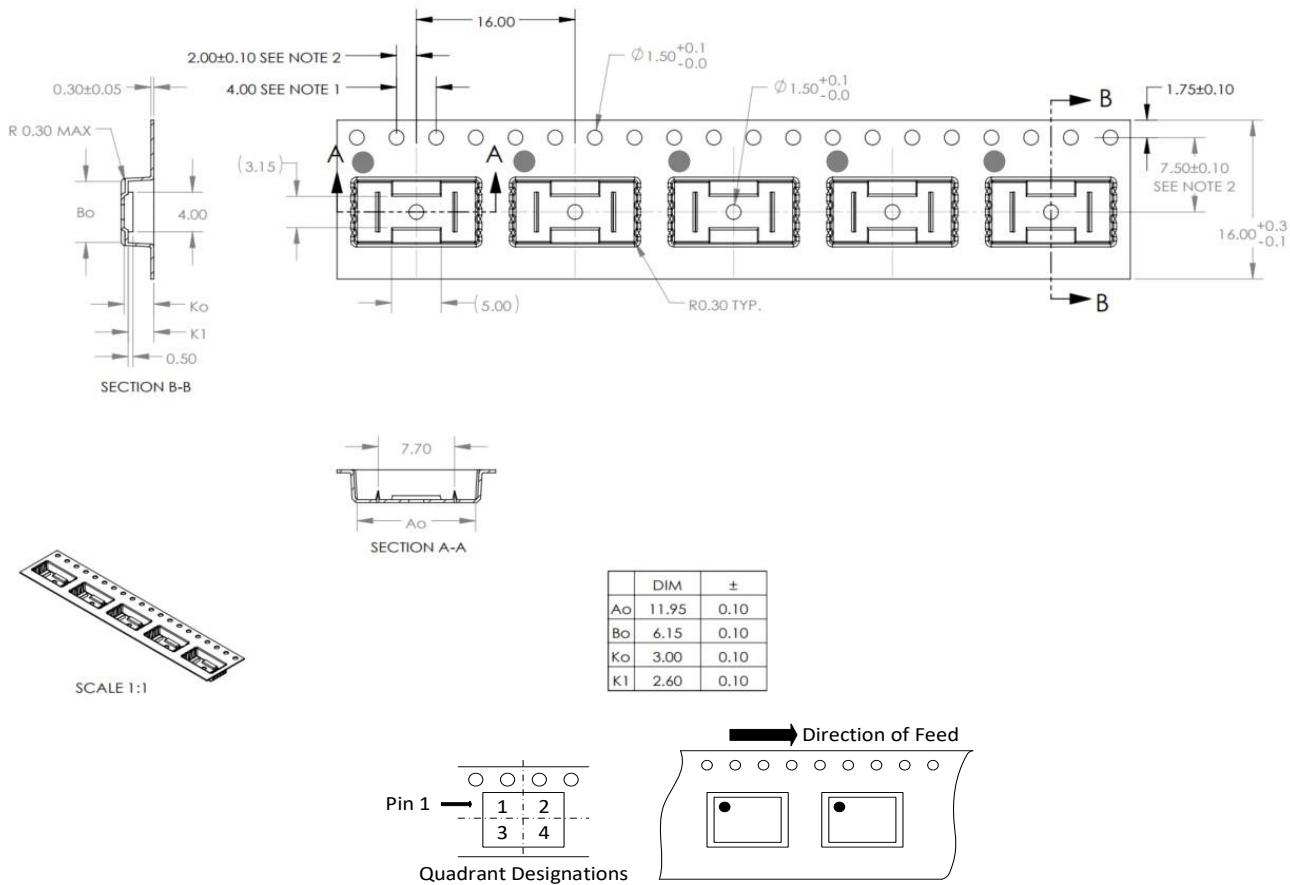
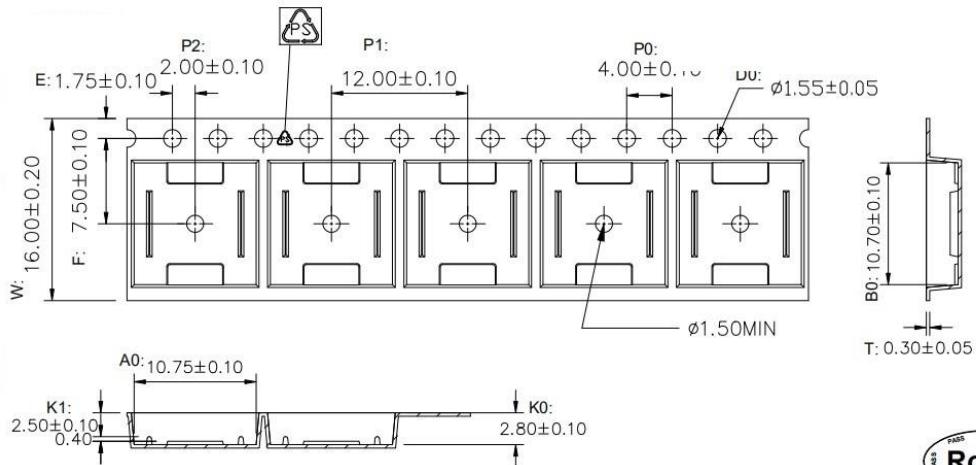


Figure 12.2 Reel Information of SOP8(300mil)



1. 10 sprocket hole pitch cumulative tolerance  $\pm 0.20$ .
2. Carrier camber is within 1 mm in 250 mm.
3. Material : Black Conductive Polystyrene Alloy .
4. All dimensions meet EIA-481 requirements.
5. Thickness :  $0.30 \pm 0.05$ mm.
6. Packing length per 22" reel : 378 Meters.(復卷 N=122)
7. Component load per 13" reel : 1000 pcs.
8. Surface resistivity :  $10^5 \sim 10^{10} \Omega/\square$



W	16.00 $\pm$ 0.20
A0	10.75 $\pm$ 0.10
B0	10.70 $\pm$ 0.10
K0	2.80 $\pm$ 0.10
K1	2.50 $\pm$ 0.10

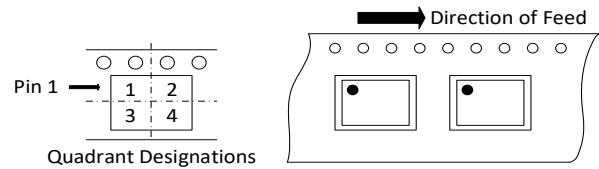


Figure 12.3 Reel Information of SOP16(300mil)

## 13. Revision History

Revision	Description	Date
1.0	Initial Release	2020/8/29

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