

### **Features and Benefits**

- Built-in pre-drive IC
- IGBT power element
- CMOS compatible input (3.3 to 5 V)
- High-side gate driver using bootstrap circuit or floating power supply
- Integrated Fast Recovery Diode (FRD) as freewheeling diode for each IGBT
- Built-in protection circuit for controlling power supply voltage drop on VCC and VB (UVLO)
- Overcurrent protection circuit (OCP)
- OCP holding time configurable with RCIN pin
- Output of fault signal during operation of protection circuit
- Output current 3 A and 5 A
- Small SIP (SLA 23-pin)

### Packages: Power SIP



## Description

The SLA6806M and SLA6816MZ inverter power module (IPM) devices provide a robust, highly-integrated solution for optimally controlling 3-phase motor power inverter systems and variable speed control systems used in energy-conserving designs to drive motors of residential and commercial appliances. These ICs take 230 VAC input voltage, and 3 A or 5 A (continuous) output current. They can withstand voltages of up to 600 V (IGBT breakdown voltage).

The SLA6806M and SLA6816MZ power package includes an IC with all of the necessary power elements (six IGBTs), pre-driver ICs (two), and freewheeling diodes (six), needed to configure the main circuit of an inverter. This enables the main circuit of the inverter to be configured with fewer external components than traditional designs.

Applications include residential white goods (home applications) and commercial appliance motor control:

- Air conditioner fan
- Refrigerator compressor
- Dishwasher pump

## Functional Block Diagram



Figure 1. Driver block diagrams.

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#### **Selection Guide**

	IGBT Breakdown		Output Current			
Part Number	Voltage, V <sub>CES</sub> (min) (V)	Voltage, V <sub>CE(sat)</sub> (typ) (V)	Continuous, I <sub>O</sub> (max) (A)	Pulsed, I <sub>OP</sub> (max) (A)		
SLA6806M	600	1.75	5	7.5		
SLA6816MZ	600	1.75	3	4.5		

### Absolute Maximum Ratings, valid at T<sub>A</sub> = 25°C

Characteristic	Symbol		Remarks	Rating	Unit
IGBT Breakdown Voltage	V <sub>CES</sub>	V <sub>CC</sub> = 15 V, I	V <sub>CC</sub> = 15 V, I <sub>C</sub> = 1 mA, V <sub>IN</sub> = 0 V		V
Main Supply Voltage	V <sub>BB</sub>	Between VB	B and GND	400	V
Main Supply Voltage (Surge)	V <sub>BB(surge)</sub>	Between VB	B and GND	450	V
Logic Supply Voltage	V <sub>CC</sub>	Between VC	C and COM	20	V
Bootstrap Voltage	V <sub>BS</sub>	Between VB a	and HS (U,V, and W phases)	20	V
Output Current, Continuous		SLA6806M	T = 25°C	5	A
Output Current, Continuous	I <sub>O</sub>	SLA6816MZ	$T_{\rm C} = 25^{\circ}{\rm C}$	3	А
Output Current, Pulsed	I <sub>OP</sub>	SLA6806M		7.5	A
		SLA6816MZ	PW ≤ 100 µs	4.5	A
Input Voltage	V <sub>IN</sub>			-0.5 to 7	V
RC Pin Input Voltage	V <sub>RC</sub>	Between RC	and COM	7	V
	P <sub>D</sub>	SLA6806M		32.8	W
Allowable Power Dissipation		SLA6816MZ	$T_{\rm C}$ = 25°C, all elements operating	32.8	W
		All elements	operating (IGBT)	3.8	°C/W
Thermal Resistance (Junction to Case)	R <sub>θJC</sub>	All elements	operating (FRD)	5.4	°C/W
Case Operating Temperature	T <sub>COP</sub>			-20 to 100	°C
Junction Temperature (MOSFET)	TJ			150	°C
Storage Temperature	T <sub>stg</sub>			-40 to 150	°C

All performance characteristics given are typical values for circuit or system baseline design only and are at the nominal operating voltage and an ambient temperature,  $T_A$ , of 25°C, unless otherwise stated.

### **Recommended Operating Conditions**

Characteristic	Symbol	Remarks	Min.	Тур.	Max.	Units
Main Supply Voltage	V <sub>BB</sub>	Between VBB and LS, $I_{BB} \le 2 A$	-	_	400	V
Logic Supply Voltage	V <sub>CC</sub>	Between VCC and COM	13.5	_	16.5	V
Pull-Up Resistor (RC Input)	R <sub>c</sub>		33	_	390	kΩ
Capacitor (RC Input)	C <sub>C</sub>		1000	-	2200	pF
Bootstrap Resistor	R <sub>B</sub>		22	-	220	Ω
Minimum Input Dulas Width	t <sub>inmin(on)</sub>	On pulse	0.5	-	-	μs
Minimum Input Pulse Width	t <sub>inmin(off)</sub>	Off pulse	0.5	_	_	μs
PWM Carrier Frequency	f <sub>C</sub>		-	_	20	kHz
Dead Time	t <sub>dead</sub>		1.5	-	-	μs

### **Typical Application Diagram**



NOTE:

- All of the input pins are connected to GND with internal pull-down resistors rated at 100 k $\Omega$ , however, an external pull-down resistor may be required to secure stable condition of the inputs if high impedance conditions are applied to them.
- To use the OCP circuit, an external shunt resistor, RS, is needed. The RS value can be obtained from the formula:
- $R_{\rm S}(\Omega) = 0.5 \, {\rm V} / {\rm Overcurrent Detection Set Current (A)}$
- A blanking timer is built-in to mask the noise generated on RS at turn-on.
- The external electrolytic capacitors should be placed as close to the IC as possible, in order to avoid malfunctions from external noise interference. Put a ceramic capacitor in parallel with the electrolytic capacitor if further reduction of noise susceptibility is necessary.

Characteristics	Symbol		Conditions	Min	Тур	Max	Units
Logic Supply Voltage	V <sub>CC</sub>	Between VCC	and COM	13.5	15	16.5	V
Logic Supply Current	I <sub>CC</sub>	V <sub>CC</sub> = 15 V	V <sub>CC</sub> = 15 V		4	6	mA
Boot Supply Current	I <sub>BOOT</sub>	VB1 to U, VB2	to V, and VB3 to W = 15 V, HIN = 5 V per circuit	-	140	400	μA
Input Voltage	VIH	V <sub>CC</sub> = 15 V, ou	tput on	-	2.0	2.5	V
Input Voltage	V <sub>IL</sub>	V <sub>CC</sub> = 15 V, ou	tput off	1.0	1.5	-	V
Input Voltage Hysteresis	V <sub>lhys</sub>	V <sub>CC</sub> = 15 V		-	0.5	-	V
Input Current	IIH	High side, $V_{CC}$	= 15 V, V <sub>IN</sub> = 5 V	-	50	100	μA
Input Current	IIL	Low side, $V_{CC}$	= 15 V, V <sub>IN</sub> = 0 V	-	_	2	μA
	V <sub>UVHL</sub>			9.0	10.0	11.0	V
	V <sub>UVHH</sub>	High side, betw	High side, between VB and U, V, or W		10.5	11.5	V
Linder reltage Leek Out	V <sub>UVHhys</sub>	High side, hysteresis		-	0.5	-	V
Undervoltage Lock Out	V <sub>UVLL</sub>	Low side, between VB and U, V, or W		10.0	11.0	12.0	V
	V <sub>UVLH</sub>			10.5	11.5	12.5	V
	V <sub>UVLhys</sub>	Low side, hyste	eresis	-	0.5	-	V
	V <sub>FOL</sub>	- V <sub>CC</sub> = 15 V		0	-	1.0	V
FO Terminal Output Voltage	V <sub>FOH</sub>			4.0	-	5.5	V
Overcurrent Protection Trip Voltage	V <sub>TRIP</sub>	V <sub>CC</sub> = 15 V		0.45	0.50	0.55	V
RC Threshold Voltage	V <sub>RCH</sub>	V <sub>CC</sub> = 15 V		3.1	_	3.85	V
Overcurrent Protection Hold Time	tp	V <sub>RC</sub> = 5 V, R <sub>C</sub> :	= 330 kΩ, C <sub>C</sub> = 2200 pF	-	870	-	μs
Blanking Time	t <sub>blank</sub>	V <sub>CC</sub> = 15 V		_	2	-	μs
IGBT Leakage Current	I <sub>CES</sub>	V <sub>CC</sub> = 15 V, V <sub>C</sub>	<sub>E</sub> = 600 V, V <sub>IN</sub> = 0 V	-	_	1	mA
		SLA6806M	V <sub>CC</sub> = 15 V, I <sub>C</sub> = 5 A, V <sub>IN</sub> = 5 V	-	1.75	2.2	V
IGBT Saturation Voltage	V <sub>CE(sat)</sub>	SLA6816MZ	V <sub>CC</sub> = 15 V, I <sub>C</sub> = 3 A, V <sub>IN</sub> = 5 V	-	1.75	2.1	V
Diada Farward Valtara		SLA6806M	V <sub>CC</sub> = 15 V, I <sub>C</sub> = 5 A, V <sub>IN</sub> = 0 V	-	2.0	2.4	V
Diode Forward Voltage	VF	SLA6816MZ	V <sub>CC</sub> = 15 V, I <sub>F</sub> = 3 A, V <sub>IN</sub> = 0 V	-	1.65	2.0	V

#### ELECTRICAL CHARACTERISTICS, valid at T<sub>A</sub> = 25°C, unless otherwise noted

#### SLA6806M ELECTRICAL CHARACTERISTICS, valid at $T_A = 25^{\circ}C$ , unless otherwise noted

Characteristics	Symbol	Conditions	Min	Тур	Max	Units
	t <sub>dH(on)</sub>		-	510	_	ns
	t <sub>rH</sub>		-	80	_	ns
Switching Time, High Side	t <sub>rrH</sub>		-	90	_	ns
	t <sub>dH(off)</sub>	$V_{BB}$ = 300 V, $V_{CC}$ = 15 V, $I_{C}$ = 5 A, 0 V $\leq$ $V_{IN}$ $\leq$ 5 V, inductive load	-	330	_	ns
	t <sub>fH</sub>		-	115	-	ns
	t <sub>dL(on)</sub>		-	500	-	ns
	t <sub>rL</sub>		-	110	-	ns
Switching Time, Low Side	t <sub>rrL</sub>		-	100	_	ns
	t <sub>dL(off)</sub>		-	330	_	ns
	t <sub>fL</sub>		-	115	-	ns

#### SLA6016MZ ELECTRICAL CHARACTERISTICS, valid at T<sub>A</sub> = 25°C, unless otherwise noted

Characteristics	Symbol	Conditions	Min	Тур	Max	Units
	t <sub>dH(on)</sub>		-	315	-	ns
	t <sub>rH</sub>		-	50	_	ns
Switching Time, High Side	t <sub>rrH</sub>		-	80	_	ns
	t <sub>dH(off)</sub>	$V_{BB}$ = 300 V, $V_{CC}$ = 15 V, $I_{C}$ = 3 A, 0 V $\leq$ $V_{IN}$ $\leq$ 5 V, inductive load	-	375	_	ns
	t <sub>fH</sub>		-	165	_	ns
	t <sub>dL(on)</sub>		-	395	_	ns
	t <sub>rL</sub>		-	60	_	ns
Switching Time, Low Side	t <sub>rrL</sub>		-	75	-	ns
	t <sub>dL(off)</sub>		-	395	_	ns
	t <sub>fL</sub>		-	170	-	ns



Switching Characteristics Definitions

Mode	HINx	LINx	High-side MOSFET	Low-side MOSFET
	L	L	Off	Off
Nie wee eld	Н	L	On	Off
Normal <sup>1</sup>	L	Н	Off	On
	Н	Н	On	On
	L	L	Off	Off
OCP	Н	L	On	Off
UCP	L	н	Off	Off
	Н	Н	On	Off
	L	L	Off	Off
VCCx UVLO <sup>2</sup>	Н	L	Off	Off
VCCX UVLO <sup>2</sup>	L	Н	Off	Off
	Н	Н	Off	Off
	L	L	Off	Off
VBx UVLO <sup>3</sup>	Н	L	Off	Off
	L	н	Off	On
	Н	Н	Off	On

#### Input-Output Truth Table

<sup>1</sup>In the case where a pair of HINx and LINx signals are asserted at the same time, a shoot-through condition will occur. Software and hardware must be carefully designed to prevent this failure by setting both the high-side and the low-side IGBTs off.

<sup>2</sup>After the VCCx power rail recovers from a UVLO condition, a rising edge of HINx starts driving the highside IGBT (edge trigger). On the other hand, after the UVLO condition is released, the input level of the LINx pins reflects the state of the low-side IGBTs (level trigger).

<sup>3</sup>After the VBx power rail recovers from a UVLO condition, a rising edge of HINx starts driving the highside IGBT (edge trigger).

### High Side Driver Input/Output Timing Diagrams



After UVLO is released, IC operation is started by the first rising edge of input

Low Side Driver Input/Output Timing Diagrams



After RC charging and releasing, the OCP operation is started by the first rising edge of input

### **Pin-out Diagrams**



### **Terminal List Table**

Number	Name	Function
1	U	Output of U phase
2	VB1	High side bootstrap terminal (U phase)
3	VB2	High side bootstrap terminal (V phase)
4	VB3	High side bootstrap terminal (W phase)
5	VCC1	High side logic supply voltage
6	COM1	High side logic GND terminal
7	HIN3	High side input terminal (W phase)
8	HIN2	High side input terminal (V phase)
9	HIN1	High side input terminal (U phase)
10	VBB1	Main supply voltage 1 (connect to VBB2 externally)
11	VBB2	Main supply voltage 2 (connect to VBB1 externally)
12	W1	Output of W phase (connect to W2 externally)
13	V	Output of V phase
14	W2	Output of W phase (connect to W1 externally)
15	LS2	Low side emitter terminal (connect to LS1 externally)
16	RC	Overcurrent protection hold time adjustment terminal
17	LS1	Low side emitter terminal (connect to LS1 externally)
18	LIN3	Low side input terminal (W phase)
19	LIN2	Low side input terminal (V phase)
20	LIN1	Low side input terminal (U phase)
21	COM2	Low side GND terminal
22	FO	Overcurrent protection fault-signal output terminal
23	VCC2	Low side logic supply voltage

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# High Voltage 3-Phase Motor Drivers

## Package Outline Drawing

Leadform 2151

Dual rows, 23 alternating pins; vertical case mounting; pin #1 opposite pad side





Leadframe plating Pb-free. Device composition complies with the RoHS directive.

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# High Voltage 3-Phase Motor Drivers





Leadframe plating Pb-free. Device composition complies with the RoHS directive.

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# High Voltage 3-Phase Motor Drivers

## Package Outline Drawing Leadform 2153

Dual rows, 23 alternating pins; pins bent  $90^\circ$  for horizontal case mounting; pin #1 in outer row



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SANKEN ELECTRIC CO., LTD.

Preliminary Data Subject to Change Without Notice January 16, 2013 Because reliability can be affected adversely by improper storage environments and handling methods, please observe the following cautions.

#### **Cautions for Storage**

- Ensure that storage conditions comply with the standard temperature (5°C to 35°C) and the standard relative humidity (around 40 to 75%); avoid storage locations that experience extreme changes in temperature or humidity.
- Avoid locations where dust or harmful gases are present and avoid direct sunlight.
- Reinspect for rust on leads and solderability of products that have been stored for a long time.

#### **Cautions for Testing and Handling**

When tests are carried out during inspection testing and other standard test periods, protect the products from power surges from the testing device, shorts between adjacent products, and shorts to the heatsink.

#### **Remarks About Using Silicone Grease with a Heatsink**

- When silicone grease is used in mounting this product on a heatsink, it shall be applied evenly and thinly. If more silicone grease than required is applied, it may produce stress.
- Volatile-type silicone greases may permeate the product and produce cracks after long periods of time, resulting in reduced heat radiation effect, and possibly shortening the lifetime of the product.
- Our recommended silicone greases for heat radiation purposes, which will not cause any adverse effect on the product life, are indicated below:

Туре	Suppliers
G746	Shin-Etsu Chemical Co., Ltd.
YG6260	Momentive Performance Materials
SC102	Dow Corning Toray Silicone Co., Ltd.

#### **Heatsink Mounting Method**

Torque When Tightening Mounting Screws. The recommended tightening torque for this product package type is: 58.8 to 78.4 N•cm (6.0 to 8.0 kgf•cm).

#### Soldering

• When soldering the products, please be sure to minimize the working time, within the following limits:

260±5°C 10 s

380±5°C 5 s

 Soldering iron should be at a distance of at least 1.5 mm from the body of the products

#### **Electrostatic Discharge**

- When handling the products, operator must be grounded. Grounded wrist straps worn should have at least 1 MΩ of resistance to ground to prevent shock hazard.
- Workbenches where the products are handled should be grounded and be provided with conductive table and floor mats.
- When using measuring equipment such as a curve tracer, the equipment should be grounded.
- When soldering the products, the head of soldering irons or the solder bath must be grounded in other to prevent leak voltages generated by them from being applied to the products.
- The products should always be stored and transported in our shipping containers or conductive containers, or be wrapped in aluminum foil.

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