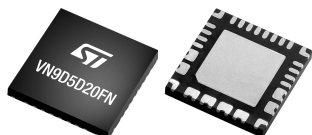


## 4-channel high-side driver with 24-bit SPI interface for automotive applications



QFN 6x6

### Product status link


[VN9D5D20FN](#)

### Product summary

<b>Order code</b>	VN9D5D20FNTR
<b>Package</b>	QFN 6X6
<b>Packing</b>	Tape and reel

## Features

Channel	V <sub>CC</sub>	R <sub>ON</sub> typ.	I <sub>LIMH</sub> typ.
0, 1	28 V	6.7 mΩ	72 A
2, 3		20 mΩ	34.5 A

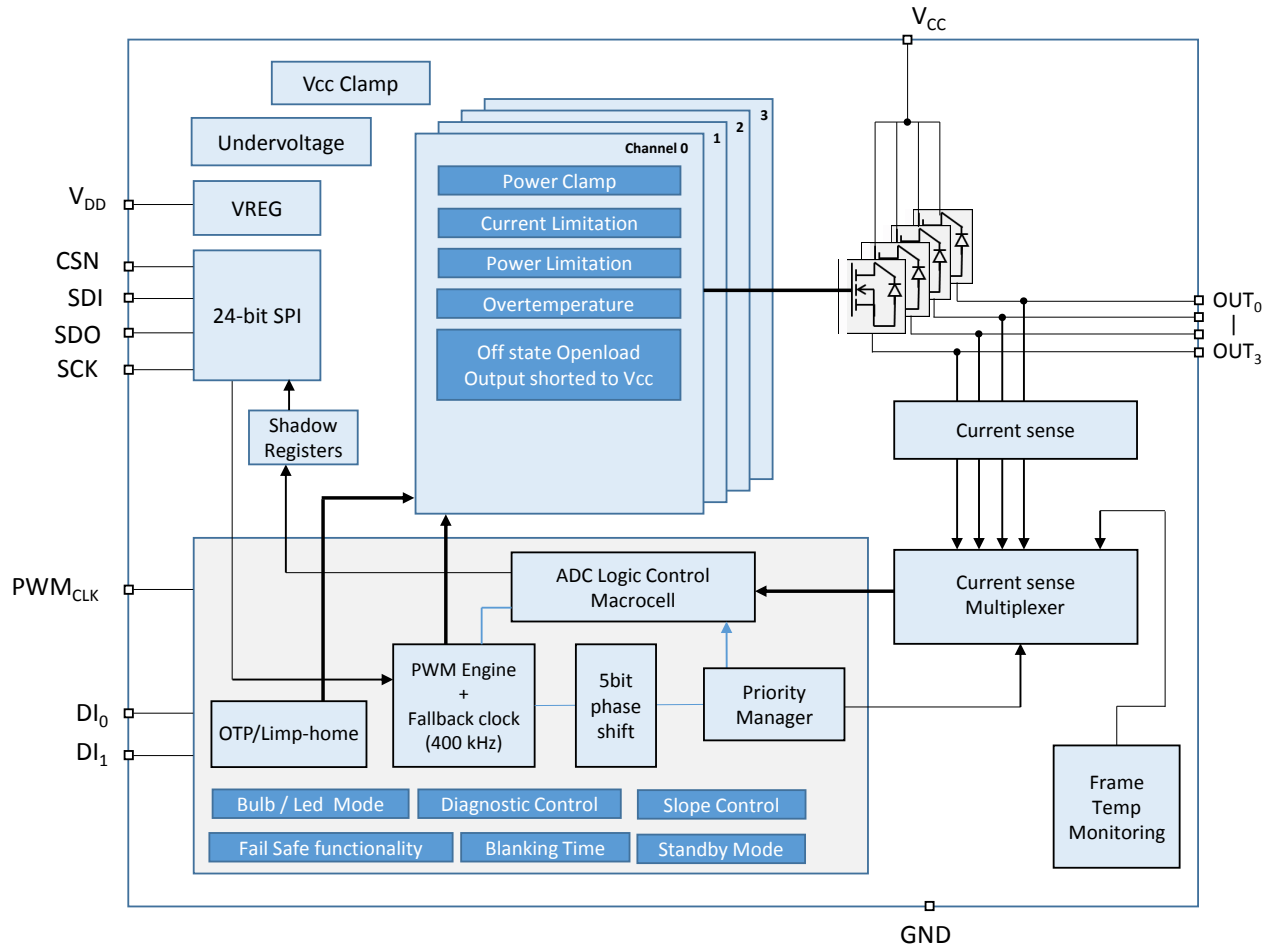
- AEC-Q100 qualified 
- General
  - Extreme low voltage operation for deep cold cranking applications (compliant with LV124, revision 2013)
  - 24-bit ST-SPI for full diagnostic and digital current sense feedback
  - Integrated 10-bit ADC for digital current sense
  - Integrated PWM engine with independent phase shift and frequency generation (for each channel)
  - Programmable Bulb/LED mode for all channels
  - Advanced limp-home functions for robust fail-safe system
  - Very low standby current
  - Optimized electromagnetic emissions
  - Very low electromagnetic susceptibility
  - Control through direct inputs and/or SPI
  - Compliant with European directive 2002/95/EC
- Diagnostic functions
  - Digital proportional load current sense
  - Synchronous diagnostic of over load and short to GND, output shorted to V<sub>CC</sub> and OFF-state open-load
  - Programmable case overtemperature warning
- Protection
  - Load current limitation
  - Self limiting of fast thermal transients
  - Undervoltage shutdown
  - Overvoltage clamp
  - Latch-off or programmable time limited auto restart (power limitation and overtemperature shutdown)
  - Load dump protected
  - Protection against loss of ground

## Description

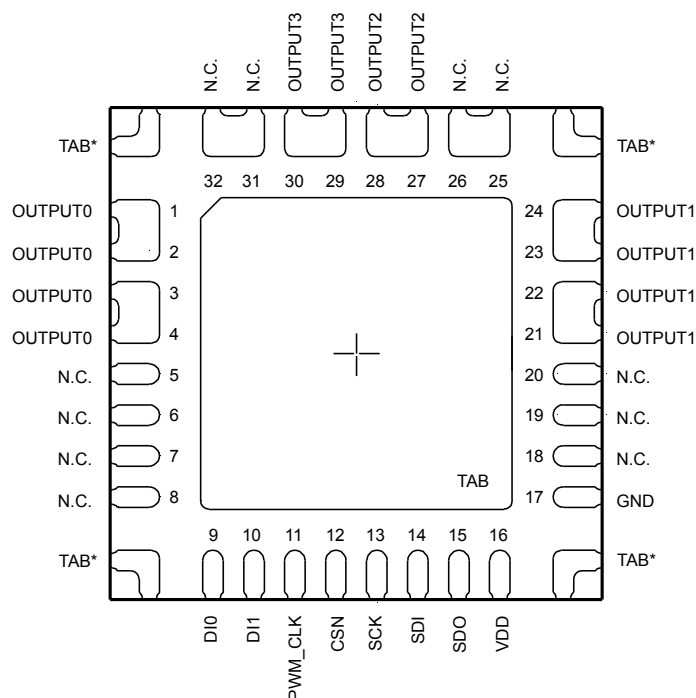
The VN9D5D20FN is a device made using STMicroelectronics VIPower technology. It is intended for driving resistive or inductive loads directly connected to the ground. The device is protected against voltage transient on  $V_{CC}$  pin. Programming, control, and diagnostics are implemented via the SPI bus. A digital current sense feedback for each channel is provided through an integrated 10-bit ADC with 0.1% of FSR. Dedicated trimming bits allow to adjust the ADC reference current. The device is equipped with four outputs controllable via SPI or with the 2-OTP assignable direct inputs. The device detects open-load in OFF-state conditions. Real-time diagnostic is available through the SPI bus (open-load, output short to  $V_{CC}$ , overtemperature, communication error, power limitation or latch off). Output current limitation protects the device in an over load condition. The device can limit the dissipated power to a safe level up to thermal shutdown intervention. Thermal shutdown can be configured as latched off or programmable time-limited auto restart. The device enters a limp-home mode in case of loss of digital supply ( $V_{DD}$ ), reset of digital memory or watchdog monitoring time-out event. In limp-home mode each output is set according to the programmed register: to be always OFF, or according to the 2x direct inputs pins.

## 1 Block diagram and pin description

Figure 1. Block diagram



GADG231120201557FSR

**Figure 2. Connection diagram (top view)**


\*: Electrically connected to TAB. Those pins are intended for thermo-mechanical purposes only. They have to be soldered, but must be electrically isolated at PCB level.

**Table 1. Pin functionality description**

Pin #	Name	Function
TAB	V <sub>CC</sub>	Battery connection: This is the backside TAB and is the direct connection to drain Power MOSFET switches.
17	GND	Ground connection: This pin serves as the ground connection for the logic part of the device.
9-10	DI0, DI1	Direct Input: Direct control for OUTx in limp-home mode through OTP programmed Direct Input assignment. Configurable as OR combination with the relevant SPI OUTx Control bit in Normal mode.
1-4	OUTPUT0	Power OUTPUT 0: It is the direct connection to the source Power MOSFET switch No. 0.
21-24	OUTPUT1	Power OUTPUT 1: It is the direct connection to the source Power MOSFET switch No. 1.
27-28	OUTPUT2	Power OUTPUT 2: It is the direct connection to the source Power MOSFET switch No. 2.
29-30	OUTPUT3	Power OUTPUT 3: It is the direct connection to the source Power MOSFET switch No. 3.
12	CSN	Chip select not (active low): It is the selection pin of the device. It is a CMOS compatible input.
13	SCK	Serial clock: It is a CMOS compatible input.
14	SDI	Serial data input: Transfers data to be written serially into the device on SCK rising edge.
15	SDO	Serial data output: Transfers data serially out of the device on SCK falling edge.
16	V <sub>DD</sub>	DC supply input for the digital control part and SPI interface. 3.3 V and 5 V compatible, this is the input of the internal Voltage Regulator.
11	PWM_CLK	PWM external clock. The frequency of the internal PWM signal is divided according to the programmed ratio. It is possible to select one of the 4xPWM divider ratios: from 1/512 to 1/4096.

## 2 Functional description

### 2.1 Device interfaces

- SPI: bi-directional interface, accessing RAM/ROM registers (CSN, SCK, SDI, SDO)
- DIx: input pins for outputs control while the device is in fail-safe mode, standby mode or reset mode (usable also in normal mode according to "Direct Input Enable Control Register" - DIENCR, setting)
- $V_{DD}$ : 5 V supply or 3.3 V supply. The internal regulator block which delivers internal logic supply voltage from  $V_{dd}$  input is able to handle both 3.3 V and 5 V.

### 2.2 Operating modes

The device can operate in seven different modes:

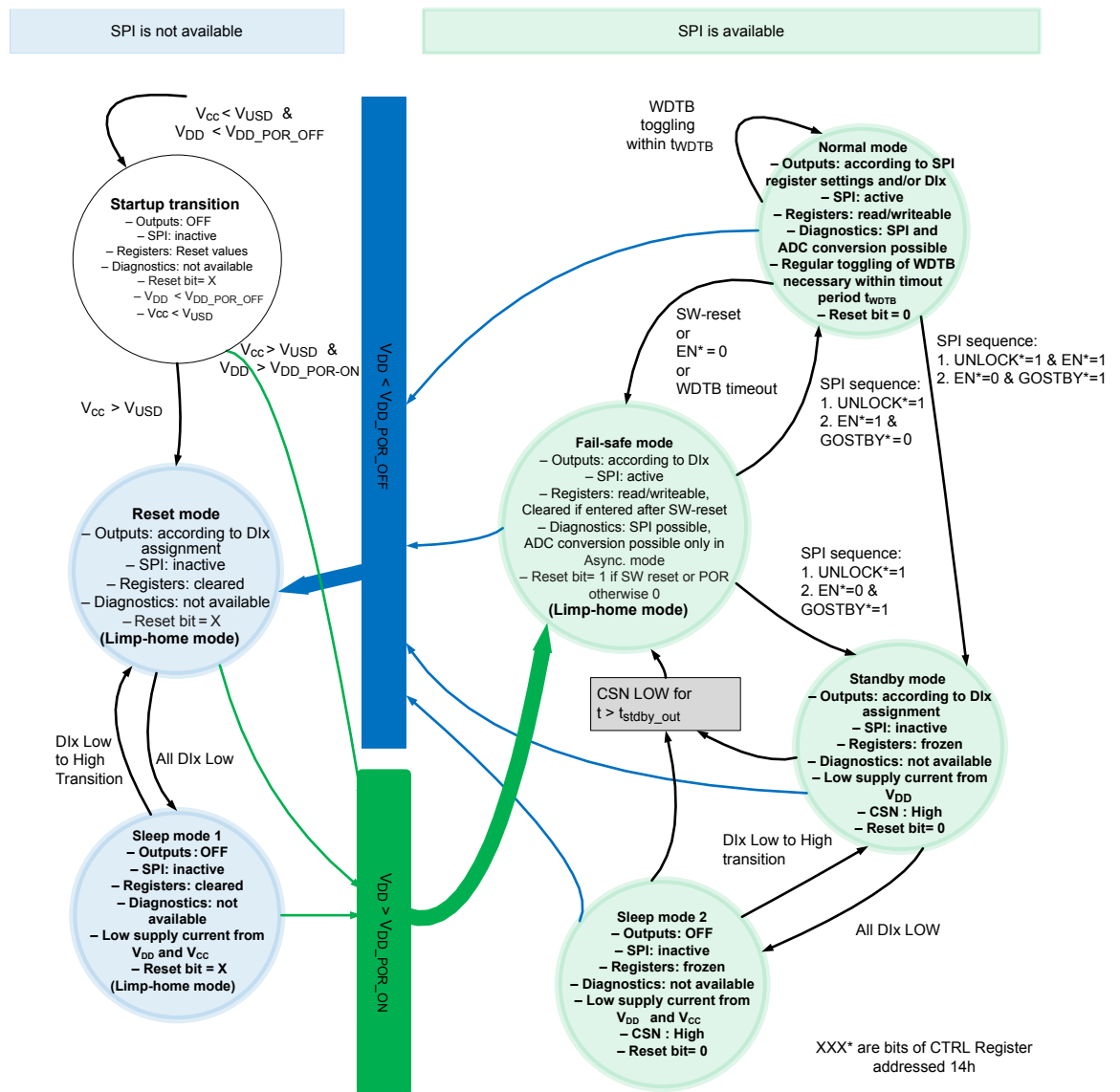
- Reset mode
- Fail-safe mode
- Normal mode
- Standby mode
- Sleep mode 1
- Sleep mode 2
- Battery undervoltage mode

The reset mode, the fail-safe mode, and the sleep mode 1 are combined into the limp-home mode. In this mode, the chip is able to operate without the connection to the SPI. All transitions between the states in limp-home mode are driven by  $V_{DD}$  and DIx. The outputs are controlled by the direct inputs DIx according to the internally programmed outputs assignment. By default, DI0 drives OUTPUT0,1, DI1 drives OUTPUT2,3.

**Table 2. Operating modes**

Operating mode	Entering conditions	Leaving conditions	Characteristics
Startup transition (this is not an operating mode)		<ul style="list-style-type: none"> <li><math>V_{CC} &gt; V_{USD}</math>: reset</li> <li><math>(V_{DD} &gt; V_{DD\_POR\_ON})</math> and <math>(V_{CC} &gt; V_{USD})</math>: fail-safe</li> </ul>	<ul style="list-style-type: none"> <li>Outputs: OFF</li> <li>SPI: inactive</li> <li>Registers: reset values</li> <li>Diagnostics: not available</li> <li>Reset bit = X</li> </ul>
Reset (limp-home mode)	<ul style="list-style-type: none"> <li>Startup mode: <math>V_{CC} &gt; V_{USD}</math></li> <li>Sleep 1: • Dlx low to high</li> <li>Any other mode: <math>V_{DD} &lt; V_{DD\_POR\_OFF}</math></li> </ul>	<ul style="list-style-type: none"> <li>All Dlx low: sleep 1</li> <li><math>V_{DD} &gt; V_{DD\_POR\_ON}</math>: fail-safe</li> </ul>	<ul style="list-style-type: none"> <li>Outputs: according to Dlx</li> <li>SPI: inactive</li> <li>Registers: reset values</li> <li>Diagnostics: not available</li> <li>Reset bit = X</li> </ul>
Sleep 1 (limp-home mode)	Reset: all Dlx = 0	<ul style="list-style-type: none"> <li><math>V_{DD} &gt; V_{DD\_POR\_ON}</math>: fail-safe</li> <li>Dlx low to high: reset</li> </ul>	<ul style="list-style-type: none"> <li>Outputs: OFF</li> <li>SPI: inactive</li> <li>Registers: reset values</li> <li>Diagnostics: not available</li> <li>Low supply current from <math>V_{CC}</math></li> <li>Reset bit = X</li> </ul>
Fail-safe (limp-home mode)	<ul style="list-style-type: none"> <li>Reset or sleep 1: <math>V_{DD} &gt; V_{DD\_POR\_ON}</math></li> <li>Standby or sleep 2: CSN low for <math>t &gt; t_{stdby\_out}</math></li> <li>Normal: EN = 0 or WDTB toggling timeout or SW-reset</li> </ul>	<ul style="list-style-type: none"> <li><math>V_{DD} &lt; V_{DD\_POR\_OFF}</math>: reset</li> <li>SPI sequence               <ol style="list-style-type: none"> <li>UNLOCK = 1</li> <li>GOSTBY = 0 and EN = 1: normal</li> </ol> </li> <li>SPI sequence               <ol style="list-style-type: none"> <li>UNLOCK = 1</li> <li>GOSTBY = 1 and EN = 0: standby</li> </ol> </li> </ul>	<ul style="list-style-type: none"> <li>Outputs: according to Dlx</li> <li>SPI: active</li> <li>Registers: read/write possible, cleared if entered after SW reset</li> <li>Diagnostics: SPI possible, ADC conversion possible only in asynchronous mode</li> <li>Reset bit = 1 if entered after SW reset or POR, else reset bit = 0</li> </ul>
Normal	<ul style="list-style-type: none"> <li>Fail-safe: SPI sequence               <ol style="list-style-type: none"> <li>UNLOCK = 1</li> <li>GOSTBY = 0 and EN = 1</li> </ol> </li> </ul>	<ul style="list-style-type: none"> <li><math>V_{DD} &lt; V_{DD\_POR\_OFF}</math>: reset</li> <li>SPI sequence               <ol style="list-style-type: none"> <li>UNLOCK = 1 &amp; EN = 1</li> <li>GOSTBY = 1 and EN = 0: standby</li> </ol> </li> <li>EN = 0 or WDTB time-out or SW reset: fail-safe</li> </ul>	<ul style="list-style-type: none"> <li>Outputs: according to SPI register settings and/or Dlx</li> <li>SPI: active</li> <li>Registers: read/write is possible</li> <li>Diagnostics: SPI and ADC conversion in all modes (sampled and asynchronous) are possible</li> <li>Regular toggling of WDTB is necessary within timeout period <math>t_{WDTB}</math></li> <li>Reset bit = 0</li> </ul>
Standby	<ul style="list-style-type: none"> <li>Normal: SPI sequence               <ol style="list-style-type: none"> <li>UNLOCK = 1 &amp; EN = 1</li> <li>GOSTBY = 1 and EN = 0</li> </ol> </li> <li>Fail-safe: SPI sequence               <ol style="list-style-type: none"> <li>UNLOCK = 1</li> <li>GOSTBY = 1 and EN = 0</li> </ol> </li> <li>Sleep 2: Dlx low to high</li> </ul>	<ul style="list-style-type: none"> <li><math>V_{DD} &lt; V_{DD\_POR\_OFF}</math>: reset</li> <li>CSN low for <math>t &gt; t_{stdby\_out}</math>: fail-safe</li> <li>All Dlx low: sleep 2</li> </ul>	<ul style="list-style-type: none"> <li>Outputs: according to Dlx</li> <li>SPI: inactive</li> <li>Registers: frozen</li> <li>Diagnostics: not available</li> <li>Low supply current from <math>V_{DD}</math></li> <li>CSN: High</li> <li>Reset bit = 0</li> </ul>
Sleep 2	Standby: all Dlx = 0	<ul style="list-style-type: none"> <li><math>V_{DD} &lt; V_{DD\_POR\_OFF}</math>: reset</li> <li>CSN low for <math>t &gt; t_{stdby\_out}</math>: fail-safe</li> <li>Dlx low to high: standby</li> </ul>	<ul style="list-style-type: none"> <li>Outputs: OFF</li> <li>SPI: inactive</li> <li>Registers: frozen</li> <li>Diagnostics: not available</li> </ul>

Operating mode	Entering conditions	Leaving conditions	Characteristics
			<ul style="list-style-type: none"> <li>Low supply current from <math>V_{DD}</math> and <math>V_{CC}</math></li> <li>CSN: high</li> <li>Reset bit = 0</li> </ul>
Battery undervoltage (this is not an operating mode)	Any mode: $V_{CC} < V_{USD}$	$V_{CC} > V_{USD} + V_{USDhyst}$ : back to last mode	<ul style="list-style-type: none"> <li>Outputs: OFF and independent from Dlx and SPI</li> <li>SPI: as the last mode</li> <li>Reset bit: as the last mode</li> </ul>

**Figure 3. Device state diagram**


GADG0404170859FSR

 For an overview over the operating modes and the triggering conditions, refer to [Section 6.4: Limp-home mode](#).

### 2.2.1 Startup transition phase

This is not an operation mode but a transition step to reset operation mode from the power-ON. In this phase, neither the digital supply voltage  $V_{DD}$  nor  $V_{CC}$  are available ( $V_{DD} < V_{DD\_POR\_ON}$  and  $V_{CC} < V_{USD}$ ).

This phase has not to be confused with undervoltage mode where also the power supply is not available ( $V_{CC} < V_{USD}$ ) after an operation mode. The device leaves this phase to reset mode as soon as  $V_{CC} > V_{USD}$ . In case ( $V_{CC} > V_{USD}$  and  $V_{DD} > V_{DD\_POR\_ON}$ ) then the device leaves this phase to fail-safe mode.

### 2.2.2 Reset mode

The device is in limp-home state.

Reset mode is entered after startup but also each time the digital supply voltage  $V_{DD}$  falls below  $V_{DD\_POR\_OFF}$  ( $V_{DD} < V_{DD\_POR\_OFF}$  and  $V_{CC} > V_{USD}$ ).

The outputs are controlled by the direct inputs  $DIx$  according to the internally programmed outputs assignment. At least one  $DIx$  is in logic high.

The SPI is inactive (no read/write possible) and the diagnostics are not available. The registers have the reset values.

The device leaves this mode only if  $V_{DD} > V_{DD\_POR\_ON}$ , or all  $DIx$  goes to low.

The reset bit inside the global status byte is unreadable since the SPI is inactive (for more information, refer to the [Section 4.3.1: Global status byte description](#)).

The diagnostics is not available, but the protections are fully functional. In case of overtemperature or power limitation, the outputs work in unlimited auto restart.

The device enters in reset mode under three conditions:

- Automatically during startup
- If it is in any other mode and if  $V_{DD}$  falls below  $V_{DD\_POR\_OFF}$
- If it is in sleep mode 1 and if one input  $DIx$  is set to 1

The device exits in reset mode under two conditions:

- If  $V_{DD}$  rises above  $V_{DD\_POR\_ON}$ , the device enters fail-safe mode
- If all inputs  $DIx$  are 0, the device enters sleep mode 1.

### 2.2.3 Fail-safe mode

The device is in limp-home state.

The digital supply voltage  $V_{DD}$  is available. ( $V_{DD} > V_{DD\_POR\_ON}$ ) and the SPI registers are active (SPI read/write).

In fail-safe mode, the digital current sense is available only in asynchronous mode and the digital fault diagnostic is available through the SPI bus.

The outputs are controlled by the direct inputs  $DIx$  regardless of the SPI commands.

The registers are cleared to their reset value if fail-safe is entered through a SW reset.

The reset bit is 1 if the last state was reset mode or the last command was a SW reset and it is reset to 0 after the first SPI access (for more information refer to [Section 4.3.1: Global status byte description](#)).

The protections are fully functional. In case of overtemperature or power limitation, the outputs work in unlimited auto restart.

The device enters fail-safe mode under the following conditions:

- If it is in reset mode or in sleep mode 1 and  $V_{DD}$  rises above  $V_{DD\_POR\_ON}$  ( $V_{DD} > V_{DD\_POR\_ON}$ )
- If it is in standby mode or in sleep mode 2 and CSN is low for  $t > t_{stdby\_out}$
- If it is in normal mode and bit EN is cleared
- If it is in normal mode and WDTB is not toggled within  $t_{WDTB}$  (watchdog timeout)
- If it is in normal mode and the SPI sends a SW reset

The device exits fail-safe mode under these conditions:

- If the SPI sends the goto normal mode sequence, the device enters normal mode:
  - In a first communication set bit UNLOCK = 1
  - In the consecutive communication set bit GOSTBY = 0 and bit EN = 1

This mechanism avoids entering the normal mode unintentionally.



- If the SPI sends the goto standby mode sequence, the device enters standby mode:
  - In a first communication set bit UNLOCK = 1
  - In the consecutive communication set bit GOSTBY = 1 and bit EN = 0

This mechanism avoids entering the standby mode unintentionally.

- If  $V_{DD}$  falls below  $V_{DD\_POR\_OFF}$ , the device enters in reset mode.

#### Transition to fail-safe mode from normal mode, using the SPI register

Only one frame is needed: Write "CTRL" 0x0001.

**Table 3. Frame 1 (write CTRL 0x0001)**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	0	0	0	0	0	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	1

#### Transition to fail-safe mode from normal mode by SW-reset

SPI reset occurs by using the "Read device information" command (applicable only on ROM area) at the reserved ROM address 0x3F. This is the equivalent of sending a 0xFF command.

Only one frame is needed: read "ROM" 0x3F.

**Table 4. Frame 1: read (ROM) 0x3F 0x--**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	1	1	1	1	1	1	1	1
DATA1	x <sup>(1)</sup>	x	x	x	x	x	x	x
	0	0	0	0	0	0	0	0
DATA2	x	x	x	x	x	x	x	x
	0	0	0	0	0	0	0	0

1. X: do not care. At least one of these bits must be zero, as 0xFFFF frame is not allowed.

The entry to the fail-safe mode can occur due to the CSN timeout.

In this specific case, the following procedure must be executed to leave the fail-safe mode:

- Removing the cause of the CSN stuck
- Toggling the CSN pin for a min  $t_{SHCH}$  (time to release the SDO line), see parameter in [Table 48. Dynamic characteristics](#)
- Sending the SPI frames

If the above procedure is not respected, the first SPI frame is rejected and the state transition fails.

### 2.2.4 Normal mode

In this mode, all device functions are available. The transition to this mode is only possible from a previous fail-safe mode.

Outputs can be driven by SPI commands or a combination of SPI commands and direct inputs DIx.

To maintain the device in normal mode, the watchdog toggle bit in the register CONFIG has to be toggled within the watchdog timeout period  $t_{WDTB}$  (see Table 48. Dynamic characteristics).

Diagnosis and current sense are available through the SPI bus (digital).

The protections are fully functional. The outputs can be set to latch-off or programmable time limited auto restart.

- In time limited, auto restart the outputs are switched on again automatically after an overtemperature or power limitation event within the limited programmed time frame (refer to Section 6.2: Blanking window values).
- In latch mode, the relevant status register has to be cleared to switch the outputs on again (refer to Section 6.2: Blanking window values).

The device enters normal mode under one condition:

- If it is in fail-safe mode and the SPI sends the goto normal mode sequence:
  - In a first communication set bit UNLOCK = 1–Write “CTRL” 0x4000;
  - In the consecutive communication set bit GOSTBY = 0 and bit EN = 1–Write “CTRL” 0x0800;

**Transition from fail-safe mode to normal mode is performed by two special SPI sequences**

- Frame 1: Write “CTRL” 0x4000
- Frame 2: Write “CTRL” 0x0800

**Table 5. Frame 1 (write CTRL 0x4000)**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	0	1	0	0	0	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	0

**Table 6. Frame 2 (write CTRL 0x0800)**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	0	0	0	0	1	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	0

The device exits normal mode under these conditions:

- If VDD falls below VDD\_POR\_OFF, the device enters reset mode.
- If the SPI sends the goto standby sequence, the device enters standby mode:
  - In a first communication set UNLOCK = 1 & EN = 1
  - In the consecutive communication set GOSTBY = 1 and EN = 0

This mechanism avoids entering standby mode unintentionally.

- If the SPI clears the EN bit (EN = 0), the device enters fail-safe mode.

- Watchdog time out: If WDTB is not toggled within the monitoring timeout period  $t_{WDTB}$ , the device enters fail-safe mode.
- If the SPI sends an SW reset command (command byte = 0xFFh), all registers are cleared and the device enters fail-safe mode.

## 2.2.5

### Standby mode

The device is in low consumption state of the digital part.

The outputs are controlled by the direct inputs Dlx only.

The current from  $V_{DD}$  is nearly 0.

The digital supply voltage  $V_{DD}$  is available. ( $V_{DD} > V_{DD\_POR\_ON}$ ) but SPI is inactive (no read/write is possible, the SPI registers are frozen to their last state before entering standby mode).

During standby mode, the above conditions are kept if at least one Dlx is in logic high.

CSN is inactive high state (independent of MCU).

The diagnostics is not available.

The protections are fully functional. The outputs are set to unlimited auto restart mode.

The device enters standby mode under three conditions:

- If it is in fail-safe mode and the SPI sends the goto standby sequence:
  - In a first communication set UNLOCK = 1
  - In the consecutive communication set GOSTBY = 1 and EN = 0
 This mechanism avoids entering standby mode unintentionally.
- If it is in normal mode and the SPI sends the goto standby sequence:
  - In a first communication set UNLOCK = 1 & EN = 1
  - In the consecutive communication set GOSTBY = 1 and EN = 0
 This mechanism avoids entering standby mode unintentionally.
- If it is in sleep mode 2 and one input Dlx is set to one.

The device exits standby mode under three conditions:

- If  $V_{DD}$  falls below  $V_{DD\_POR\_OFF}$ , the device enters in reset mode.
- If CSN is low for  $t > t_{stdby\_out}$ , the device wakes up. As the device is in fail-safe mode, the outputs are controlled through Dlx pins, the ADC conversion is possible only in asynchronous mode and the digital diagnostic is available through the SPI bus.
- If all direct inputs Dlx are 0, the device enters sleep mode 2 resulting in minimal supply current from  $V_{CC}$  and  $V_{DD}$ .

**Transition from fail-safe mode to standby mode using SPI: two frames needed.**

- Frame 1: write "CTRL" 0x4000
- Frame 2: write "CTRL" 0x8000

**Table 7. Frame 1 (write CTRL 0x4000)–fail-safe mode to standby mode**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	0	1	0	0	0	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	0

**Table 8. Frame 2 (write CTRL 0x8000)—fail-safe mode to standby mode**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	1	0	0	0	0	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	0

**Transition from normal mode to standby mode using SPI: two frames needed**

- Frame 1: write “CTRL” 0x4801
- Frame 2: write “CTRL” 0x8000

**Table 9. Frame 1 (write CTRL 0x4801)—normal mode to standby mode**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	0	1	0	0	1	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	1

**Table 10. Frame 2 (write CTRL 0x8000)—normal mode to standby mode**

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CMD	OC1	OC0	Address					
	0	0	0	1	0	1	0	0
DATA1	GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	Not used	Not used	Not used
	1	0	0	0	0	0	0	0
DATA2	Not used	Not used	Lockbit3	Lockbit2	Lockbit1	Lockbit0	PWMSYNC	Parity
	0	0	0	0	0	0	0	0

## 2.2.6

### Sleep mode 1

The device is in limp-home state.

The device has very low consumption for both digital and power parts. Current consumption from the digital part is nearly zero and the current consumption on  $V_{CC}$  is supply current in sleep mode 1.

The digital supply voltage  $V_{DD}$  is not available ( $V_{DD} < V_{DD\_POR\_OFF}$ ) and SPI is inactive (the read and write functions are not possible and all registers are cleared and have the reset values).

- The diagnostics is not available.
- The output stages are all off.
- Protections are inactive.

The device enters sleep mode 1 under one condition:

- If from reset mode, all direct inputs  $DIx$  are going low.

The device exits sleep mode 1 under two conditions:

- If  $V_{DD}$  rises above  $V_{DD\_POR\_ON}$ , the device enters fail-safe mode.
- If one of the inputs  $DIx$  is set to 1, the device enters in reset mode.

### 2.2.7 Sleep mode 2

The device is in very low consumption state for both digital and power parts. Current consumption from the digital part is  $I_{DDstd}$  and the current consumption on  $V_{CC}$  is supply current in sleep mode 2.

The digital supply voltage  $V_{DD}$  is available ( $V_{DD} > V_{DD\_POR\_ON}$ ) but SPI is not active (the read and write functions are not possible and all registers are frozen).

CSN is in inactive high state (independent of MCU).

In sleep mode 2 the following limitations must be considered:

- The diagnostics is not available
- The output stages are all off
- Protections are inactive

The device enters in sleep mode 2 under one condition:

- If from standby mode, all direct inputs  $DIx$  are going low

Sleep mode 2 can be left with three conditions:

- If  $V_{DD}$  falls below  $V_{DD\_POR\_OFF}$ , the device enters in reset mode
- If CSN is low for  $t > t_{stdby\_out}$ , the device enters fail-safe mode
- If one of the inputs  $DIx$  is set to 1, the device enters standby mode

### 2.2.8 Battery undervoltage mode

This is not an operation mode but a transition step, where the power supply voltage is ( $V_{CC} < V_{USD}$ ).

If the battery supply voltage  $V_{CC}$  falls below the undervoltage shutdown threshold ( $V_{CC} < V_{USD}$ ) the device enters battery undervoltage mode.

The current sense diagnostic is not available. The output stages are off regardless of SPI status or  $DIx$ .

Three different cases occur, depending on the operating mode:

#### 1. From normal mode and from fail-safe mode:

In this mode, the digital supply voltage  $V_{DD}$  is available ( $V_{DD} > V_{DD\_POR\_ON}$ ). The SPI is active and read/write functions are possible. The SPI diagnostics is available. After entering to the undervoltage mode, the information about the undervoltage is saved in a flag (VCCUV) in the OUTSRx register, the SPI register contents are retained. The SPI-register reading is always possible.

If  $V_{CC}$  rises above the threshold ( $V_{USD} + V_{USDhyst}$ ) the device returns to the last mode and the flag is cleared (VCCUV).

If during this state  $V_{DD}$  decreases to  $V_{DD} < V_{DD\_POR\_OFF}$ , the device is reset completely. The last operation mode information is lost, the device logic part is unpowered, therefore after increasing the supply voltage to ( $V_{CC} > V_{USD} + V_{USDhyst}$ ) the operation mode will be in reset mode.

If during this state, the  $DIx$  is changed, the operation mode is not changed and the output state is changed accordingly after  $V_{CC}$  recovering.

#### 2. From standby and sleep mode 2 modes:

In this mode, the digital supply voltage  $V_{DD}$  is available ( $V_{DD} > V_{DD\_POR\_ON}$ ). The SPI is not active and the registers are frozen. The SPI diagnostics is not available. After entering to the undervoltage mode, the information about the undervoltage is not saved in a flag (VCCUV).

If  $V_{CC}$  rises above the threshold ( $V_{USD} + V_{USDhyst}$ ) the device returns to the last mode. If during this state (undervoltage mode)  $V_{DD}$  decreases to  $V_{DD} < V_{DD\_POR\_OFF}$ , the device is reset completely. The last operation mode information is lost, the device logic part is unpowered, therefore after increasing the supply voltage to ( $V_{CC} > V_{USD} + V_{USDhyst}$ ) the operation mode will be in reset mode.

If during this state (under voltage mode) the  $DIx$  is changed, the operation mode is also changed. After  $V_{CC}$  recovering, this new operation mode is taken into account.

#### 3. From reset mode or sleep mode 1:

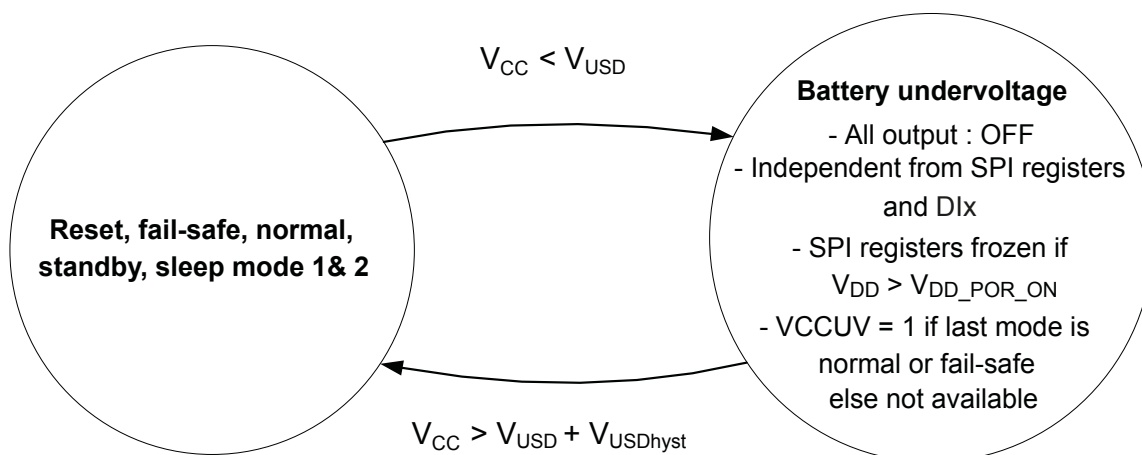
In this mode, the digital supply voltage  $V_{DD}$  is not available ( $V_{DD} < V_{DD\_POR\_OFF}$ ) and SPI is not active. It is not possible to read/write via SPI. All SPI registers have the reset values. After entering to the undervoltage mode, the information about the undervoltage is not saved in a flag (VCCUV).

If  $V_{CC}$  rises above the threshold  $V_{USD} + V_{USDhyst}$ , the device returns to the last mode. If during this state  $V_{DD}$  increases to  $V_{DD} > V_{DD\_POR\_ON}$ , the device is completely reset. After  $V_{CC}$  recovering ( $V_{CC} > V_{USD} + V_{USDhyst}$ ), there will be a startup transition.

The undervoltage flag (VCCUV) is not saved in the following operation modes:

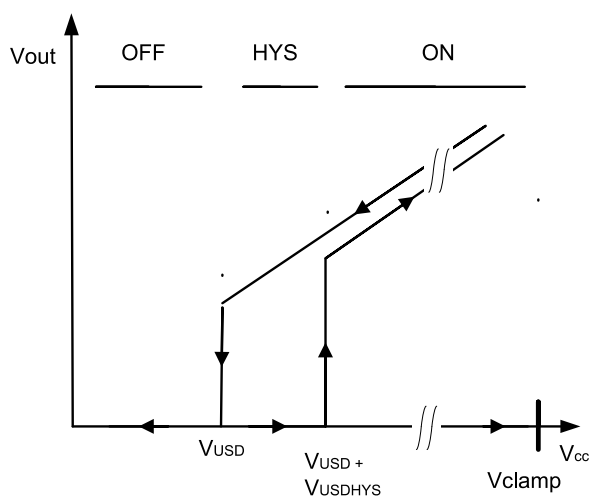
- Reset mode
- Sleep mode 1
- Sleep mode 2
- Standby mode

**Figure 4. Battery undervoltage shutdown diagram**



GADG0404171125PS

**Figure 5. Undervoltage shutdown**



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### 2.2.9

#### Limp-home mode

The reset mode, the fail-safe mode, and the sleep mode 1 are combined into the limp-home mode. In this mode, the chip is able to operate without the connection to the SPI. All transitions among the states in limp-home mode are driven by VDD and Dlx. The outputs are controlled by the direct inputs Dlx.

The Dlx inputs can be driven by either a  $\mu$ C I/O port or directly by KL15 (12 V) through series resistance. Each output has an OTP programmed direct input assignment for limp-home operation. Any output can be programmed to be always OFF in the limp-home, or according to DI0 pin state or according to DI1 pin state.

Default configurations are:

- DI0 drivers OUT0,1
- DI1 drivers OUT2,3

For a direct entry to the limp-home mode during normal operating mode, the MCU uses the watchdog toggle bit (WDTB) or a dedicated SPI command. Changing the polarity of the WDTB within watchdog timeout ( $t_{WDTB}$ ) keeps the device in normal mode.

## 3 Protections

### 3.1 Pre-warning

If the case-temperature rises above the case-thermal detection pre-warning threshold  $T_{CSD}$ , the bit  $T_{CASE}$  in the Global Status Byte is set.  $T_{CASE}$  is cleared automatically when the case-temperature drops below the case-temperature reset threshold  $T_{CR}$ .

### 3.2 Junction overtemperature (OT)

If the junction temperature of one channel rises above the shutdown temperature  $T_{TSD}$ , an overtemperature event (OT) is detected.

The channel is switched OFF and the corresponding bit in the Address OUTSRx register - Channel Feedback Status Register (CHFBSRx) is set. Consequently, the thermal shutdown bit (bit 4) in the Global Status Byte is set. Each output channel can be either set as latch-off or programmable time limited autorestart operations in case of junction overtemperature event.

- In Latched OFF operation, the output remains switched OFF and the corresponding bit "CHLOFFSRx" in the OUTSRx register is set, until the junction temperature falls below  $T_{RS}$  and a write command to the addressed latched OFF channel is sent (CHLOFFTCRx). The action will clear the corresponding bit "CHLOFFSRx" in the OUTSRx register and bit 4 in the Global Status Byte. Bit 4 only remains stuck at logic high if another fault condition is present at the same time.
- In time limited auto-restart, during the programmed time the output is switched off as described and switches on again automatically when the junction temperature falls below the reset temperature  $T_{RS}$ . The status bit "CHFBSRx" in the OUTSRx register is latched during OFF-state of the channel in order to allow asynchronous diagnostic and it is automatically cleared when the junction temperature falls below the thermal reset temperature of OT detection  $T_{RS}$ . After the programmed time expiration, the output remains switched OFF and acts as the above described latch-off mode.

### 3.3 Power limitation (PL)

If the difference between junction temperature and case temperature ( $\Delta T = T_J - T_C$ ) rises above the power limitation threshold  $\Delta T_{PLIM}$ , a power limitation event is detected.

The corresponding bit in the OUTSRx register - channel feedback status bit (CHFBSR) is set. The channel is switched OFF and therefore the bit 4 in the global status byte is set.

Each output channel can be either set as latch-off or programmable time limited autorestart operations in case of a power limitation event.

- In latched OFF operation, the output remains switched OFF and the corresponding bit "CHLOFFSRx" in the OUTSRx register is set, until the junction temperature falls below  $T_R$  and a write command to the addressed latched OFF channel is sent (CHLOFFTCRx). The action clears the corresponding bit "CHLOFFSRx" in the OUTSRx register and bit 4 in the global status byte. Bit 4 only remains stuck at logic high if another fault condition is present at the same time.
- In time limited autorestart, during the programmed time, the output is switched off as described and switches on again automatically when the difference between junction temperature and case temperature ( $\Delta T = T_J - T_C$ ) decreases below  $\Delta T_R$ . The status bit "CHLOFFSR" is latched during OFF-state of the channel in order to allow asynchronous diagnostic and it is automatically cleared when the difference between junction temperature and case temperature ( $\Delta T = T_J - T_C$ ) decreases below  $\Delta T_{RS}$ . After the programmed time expiration, the output remains switched OFF and acts as in the above described latch-off mode.



## 4 SPI functional description

### 4.1 SPI communication

The SPI communication is based on a standard ST-SPI 24-bit interface, using CSN, SDI, SDO and SCK signal lines.

Input data are shifted into SDI, MSB first while output data are shifted out on SDO, MSB first.

#### 4.1.1 Signal description

During all operations,  $V_{DD}$  must be held stable and within the specified valid range:  $V_{DD}$  min. to  $V_{DD}$  max.

**Table 11. SPI signal description**

Name	Function
Serial clock SCK	This input signal provides the timing of the serial interface. Data present at serial data input (SDI) are latched on the rising edge of serial clock (SCK). Data on serial data output (SDO) change after the falling edge of serial clock (SCK).
Serial data input SDI	This input signal is used to transfer data serially into the device. It receives data to be written. Values are sampled on the rising edge of serial clock (SCK).
Serial data output SDO	This output signal is used to transfer data serially out of the device. Data are shifted out on the falling edge of serial clock (SCK).
Chip select CSN	<p>When this input signal is high, the device is deselected and serial data output (SDO) is high impedance. Driving this input low enables the communication. The communication must start on a low level of serial clock (SCK). Data are accepted only if exactly 24 bits have been shifted in.</p> <p><b>Note:</b> as per the ST_SPI standard, in case of failing communication:</p> <ul style="list-style-type: none"> <li>CSN stuck at high: <ul style="list-style-type: none"> <li>If the device is in normal mode, a WDTB timeout will force the device into fail-safe mode. The serial data output (SDO) will stay in high impedance (High Z). Any valid communication arrived after this event will be accepted by the device.</li> </ul> </li> <li>CSN stuck at low: <ul style="list-style-type: none"> <li>in this case and whatever the mode of the device, a CSN timeout protection will be activated and force the device to release the SPI bus. Then the serial data output (SDO) will go into high impedance (High Z)</li> </ul> </li> </ul> <p>A reset of the CSN timeout (see <math>T_{SHCH}</math> in <a href="#">Table 48. Dynamic characteristics</a>) is activated with a transition low to high on CSN pin (or with a Power-on Reset or Software reset). With this reset, the serial data output (SDO) will be released and any valid communication will be accepted by the device. Without this reset, next communication will not be taken into account by the device.</p>

#### 4.1.2 Connecting to the SPI bus

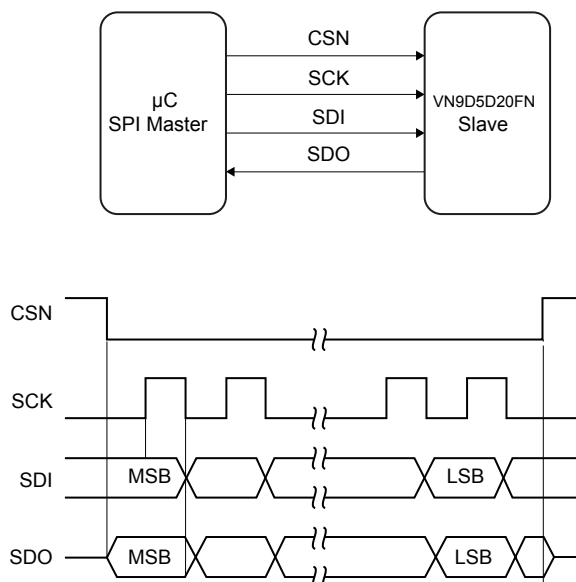
A schematic view of the architecture between the bus and devices can be seen in [Figure 7. Bus master and two devices in a normal configuration](#).

All input data bytes are shifted into the device, MSB first. The serial data input (SDI) is sampled on the first rising edge of the serial clock (SCK) after chip select (CSN) goes low. All output data bytes are shifted out of the device on the falling edge of SCK, MSB first on the first falling edge of the chip select (CSN).

### 4.1.3 SPI mode

Supported SPI mode during a communication phase can be seen in the following figure:

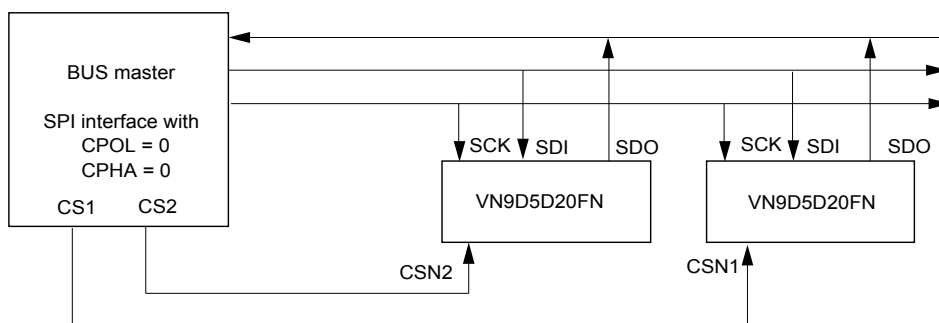
**Figure 6. Supported SPI mode**



This device can be driven by a microcontroller with its SPI peripheral running in the following mode:

- CPOL = 0, CPHA = 0

**Figure 7. Bus master and two devices in a normal configuration**



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## 4.2 SPI protocol

### 4.2.1 SDI, SDO format

SDI format during each communication frame starts with a command byte. It begins with two bits of operating code (OC0, OC1) which specify the type of operation (read, write, read and clear status, read device information) and it is followed by a 6-bit address (A0:A5). The command byte is followed by two input data bytes (D15:D8) and (D7:D0).

**Table 12. Command byte**

MSB	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB
OC1	OC0	A5	A4	A3	A2	A1	A0

**Table 13. Input data byte 1**

MSB	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB
D15	D14	D13	D12	D11	D10	D9	D8

**Table 14. Input data byte 2**

MSB	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB
D7	D6	D5	D4	D3	D2	D1	D0 <sup>(1)</sup>

1. D0 is the parity bit.

SDO format during each communication frame starts with a specific byte called global status byte (see [Table 15](#) for more details of bit0-bit7). This byte is followed by two output data bytes (D15:D8) and (D7:D0).

**Table 15. Global status byte**

MSB	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB
bit7	bit6	bit5	bit4	bit3	bit2	bit1	bit0

**Table 16. Output data byte 1**

MSB	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB
D15	D14	D13	D12	D11	D10	D9	D8

**Table 17. Output data byte 2**

MSB	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	LSB
D7	D6	D5	D4	D3	D2	D1	D0

### 4.2.2 Operating code definition

The SPI interface features four different addressing modes which are listed in [Table 18. Operating codes](#).

**Table 18. Operating codes**

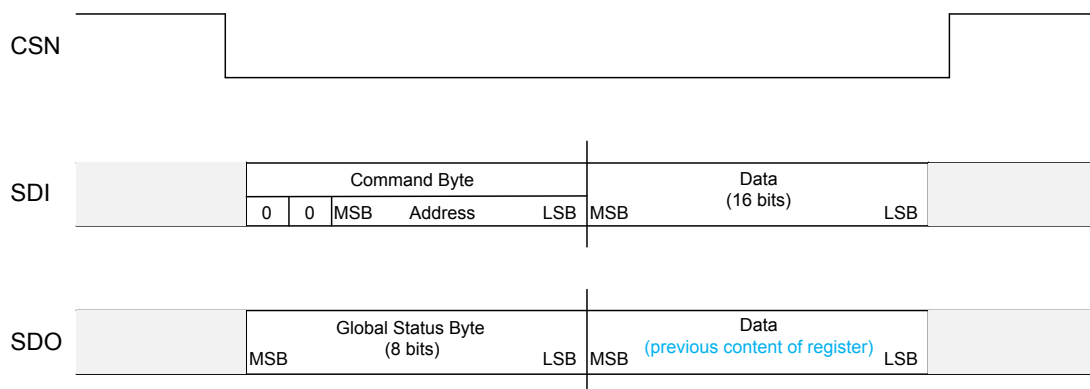
OC1	OC0	Meaning
0	0	Write operation
0	1	Read operation
1	0	Read and clear status operation
1	1	Read device information

#### Write mode

The write mode of the device allows to write the content of the input data byte into the addressed register (see list of registers in [Table 23. RAM memory map](#)). Incoming data are sampled on the rising edge of the serial clock (SCK), MSB first.

During the same sequence the outgoing data are shifted out MSB first on the falling edge of the CSN pin and the subsequent bits on the falling edge of the serial clock (SCK). The first byte corresponds to the Global Status Byte and the second to the previous content of the addressed register.

**Figure 8. SPI write operation**



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#### Read mode

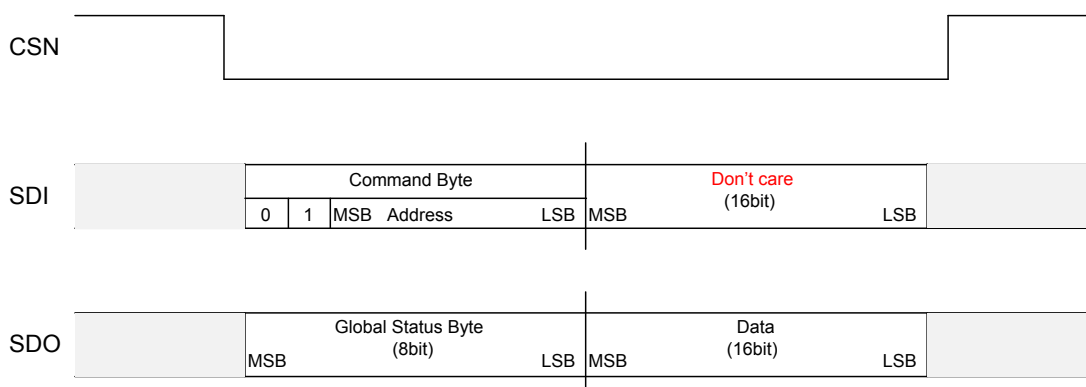
The read mode of the device allows to read and to check the state of any register.

Incoming data are sampled on the rising edge of the serial clock (SCK), MSB first.

The command byte allows to determine which register content is read, whilst the other two data bytes are "don't care".

In case of a read mode on an unused address, the global status/error byte on the SDO pin is followed by 0x0000 word.

In order to avoid inconsistency between the Global Status byte and the Status register, the Status register contents are frozen during the SPI communication.

**Figure 9. SPI read operation**


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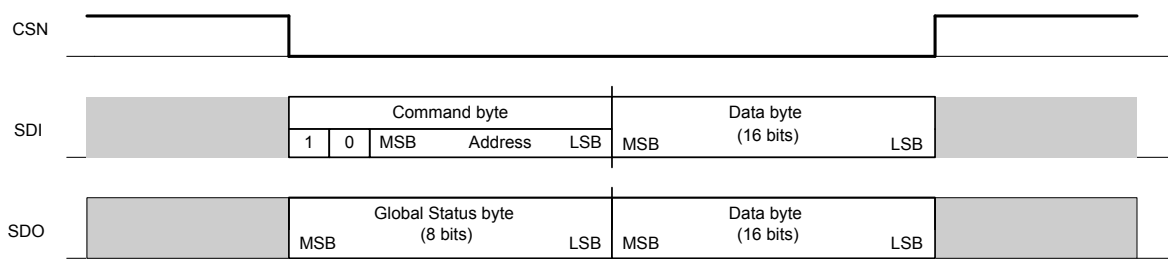
### Read and clear status command

The read and clear status operation is used to clear the content of the addressed status register (see [Table 23. RAM memory map](#)). A read and clear status operation with address 0x3Fh clears all Status registers simultaneously.

Incoming data are sampled on the rising edge of the serial clock (SCK), MSB first. The command byte allows to determine which register content is read and the payload bits set to 1 into the data byte determine the bits into the register which have to be cleared.

Outgoing data are shifted out MSB first on the falling edge of the CSN pin and others on the falling edge of the serial clock (SCK). The first byte corresponds to the Global Status byte and the second to the content of the addressed register.

In order to avoid inconsistency between the Global Status byte and the Status register, the Status register contents are frozen during SPI communication.

**Figure 10. SPI read and clear operation**


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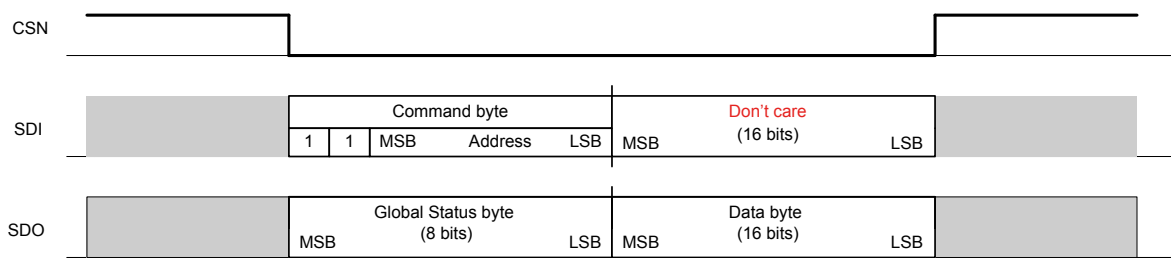
### Read device information

Specific information can be read but not modified during this mode. Accessible data can be seen in [Table 24. ROM memory map](#).

Incoming data are sampled on the rising edge of the serial clock (SCK), MSB first. The command byte allows to determine which information is read whilst the other two data bytes are "don't care".

Outgoing data are shifted out MSB first on the falling edge of the CSN pin and others on the falling edge of the serial clock (SCK). The first byte corresponds to the Global Status byte and the second to the content of the addressed register and the third byte is 0x00.

**Note:** ROM is based on the 8-bit registers, then even if 16 bits are returned, only the second byte contains the addressed ROM register.

**Figure 11. SPI read device information**


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### 4.2.3

#### Special commands

##### 0xFF – SW-Reset: set all control registers to default

An OpCode '11' (read device information) addressed at '111111' forces a software reset of the device, second and third bytes are "don't care" provided that at least one bit is zero.

*Note:*

*An OpCode '11' at address '111111' with data field equal to '1111111111111111' on the SPI frame is recognized as a frame error and the SPIE bit of GSB is set.*

**Table 19. 0xFF: SW\_Reset**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Command							
OC1	OC0	Address					
1	1	1	1	1	1	1	1
DATA1	X <sup>(1)</sup>	X	X	X	X	X	X
	0	0	0	0	0	0	0
DATA2	X	X	X	X	X	X	X
	0	0	0	0	0	0	0

1. X: do not care.

##### 0xBF - clear all status registers (RAM access)

When an OpCode '10' (read and clear operation) at address b'111111 is performed.

**Table 20. Clear all status registers (RAM access)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Command							
OC1	OC0	Address					
1	0	1	1	1	1	1	1
DATA1	X <sup>(1)</sup>	X	X	X	X	X	X
	0	0	0	0	0	0	0
DATA2	X	X	X	X	X	X	X
	0	0	0	0	0	0	0

1. X: do not care.

*Note:*

*Reset value = the value of the register after a power on.*

*Default value = the default value of the register. Currently this is equivalent to the reset value.*

*Cleared register = explicitly read and clear of the register, if it is not write protected.*

## 4.3 Register map

The device contains a set of RAM and ROM registers. The RAM registers are used for device configuration and device status while ROM registers are used for device identification. Since ST-SPI is used, the global status byte defines the device status, containing fault information.

### 4.3.1 Global status byte description

The data shifted out on SDO during each communication starts with a specific byte called global status byte. This one is used to inform the microcontroller about global faults, which can happen at channel-side level (that is, like thermal shutdown, OLOFF...) or on the SPI interface (like watchdog monitoring timeout event, communication error,...). This specific register has the following format:

**Table 21. Global status byte (GSB)**

MSB							LSB
GSBN	RSTB	SPIE	TSD/OTOVL	T <sub>CASE</sub>	L <sub>OFF</sub>	OL <sub>OFF</sub>	FS

**Table 22. Global status byte**

Bit	Name	Reset	Content
7	Global status bit not	0	The GSBN is a logically NOR combination of bit 0 to bit 6. This bit can also be used as the global status flag without starting a complete communication frame as it is present directly after pulling CSN low.
6	Reset bit	1	The RSTB indicates a device reset. In case this bit is set, all internal control registers are set to default and kept in that state until the bit is cleared. The reset bit is automatically cleared by any valid SPI communication
5	SPI error	0	The SPIE is a logical OR combination of errors related to a wrong SPI communication (SCK count and SDI stuck at errors). The SPIE bit is automatically set when SDI is stuck at High or Low. The SPIE is automatically cleared by a valid SPI communication.
4	Thermal shutdown (OT) or power limitation (PL) or VDS	0	This bit is set in case of thermal shutdown, power limitation or in case of high VDS (VDS) at turn-off detected on any channel. The contribution of high VDS failure is maskable.
3	T <sub>CASE</sub>	0	This bit is set if the frame temperature is greater than the threshold and can be used as a temperature prewarning. The bit is cleared automatically when the frame temperature drops below the case-temperature reset threshold (TCR).
2	Latch OFF (LOFF)	0	The device error bit is set in case when one or more channels are latched OFF
1	Open-load at off-state or output shorted to V <sub>CC</sub> (OLOFF)	0	The open-load at off state bit is set when an open-load off state or an output shorted to V <sub>CC</sub> condition is detected on any channel
0	Fail-safe	1	The bit is set in case the device operates in fail-safe mode. A detailed description of these root-causes and the fail-safe state itself is specified in the paragraph "Fail-safe state"

**Note:** The FFh or 00h combinations for the global status byte are not possible, due to the active low of the global status bit (bit 7), exclusive combination exists between bit 7 and bit 0 - bit 6. Consequently, a FFh or 00h combination for the global status byte must be detected by the microcontroller as a failure (SDO stuck to GND or to V<sub>DD</sub> or loss of SCK).

### 4.3.2 RAM

RAM registers can be separated according to the frequency of usage

- init - register is read / written during initialization phase (single shot action)
- continuous - read / write / read and clear registers often accessed, applying outputs control and diagnostic
- rare - read/ read and clear status of device registers accessed on demand (in case of failure)

**Table 23. RAM memory map**

Address	Name	Access	Content	Access type	Reset value
<b>CONTROL REGISTERS</b>					
00h	OUTCTRCR0	Read/Write	Output Control Configuration Register Channel 0	init	0x0000
01h	OUTCTRCR1	Read/Write	Output Control Configuration Register Channel 1	init	0x0000
02h	OUTCTRCR2	Read/Write	Output Control Configuration Register Channel 2	init	0x0000
03h	OUTCTRCR3	Read/Write	Output Control Configuration Register Channel 3	init	0x0000
Not available on this device					
Not available on this device					
08h	OUTCFGR0	Read/Write	Output Configuration Register 0	init	0x0000
09h	OUTCFGR1	Read/Write	Output Configuration Register 1	init	0x0000
0Ah	OUTCFGR2	Read/Write	Output Configuration Register 2	init	0x0000
0Bh	OUTCFGR3	Read/Write	Output Configuration Register 3	init	0x0000
Not available on this device					
Not available on this device					
10h	CHLOFFTCR0	Read/Write	Channel Latch-off Timing Control Register 1 (Channels 2, 1, 0)	init	0x0000
11h	CHLOFFTCR1	Read/Write	Channel Latch-off Timing Control Register 0 (Channel 3)	init	0x0000
13h	SOCR	Read/Write	Channel Control Register	init	0x0000
14h	CTRL	Read/Write	Control Register	init	0x0000
... not used area					
<b>STATUS REGISTERS</b>					
20h	OUTSR0	Read/Clear	Output Status Register Channel0	rare	0x0000
21h	OUTSR1	Read/Clear	Output Status Register Channel1	rare	0x0000
22h	OUTSR2	Read/Clear	Output Status Register Channel2	rare	0x0000
23h	OUTSR3	Read/Clear	Output Status Register Channel3	rare	0x0000
Not available on this device					
Not available on this device					
28h	ADC0SR	Read	Digital Current Sense Channel 0	continuous	0x0000
29h	ADC1SR	Read	Digital Current Sense Channel 1	continuous	0x0000
2Ah	ADC2SR	Read	Digital Current Sense Channel 2	continuous	0x0000
2Bh	ADC3SR	Read	Digital Current Sense Channel 3	continuous	0x0000
Not available on this device					
Not available on this device					
31h	ADC9SR	Read	Digital Frame Temperature Sense	continuous	0x0000

**Note:** Any command (write, read or read and clear status) executed on a “not used” RAM register, i.e. a not assigned address, does not have any effect: there is no change in the Global Status byte (no communication error, no error flag). The data written to this address is ignored. The data read from this address contains 0000h, independently of what has been written previously to this address.

A write command on “don’t care” bits of an assigned RAM register address does not have any effect: There is no change on the Global Status byte. The data written to the “don’t care bits” is ignored. The content of the “don’t care bits” remains at “0” independently of the data written to these bits.



### 4.3.3

### ROM

This memory is used for device identification.

**Table 24. ROM memory map**

Address	Name	Description	Access	Content
00h	Company code	Indicates the code of STM company	Read only	00H
01h	Device Family	Indicates the product family	Read only	01H
02h	Product Code 1	Indicates the first code of the product	Read only	58H
03h	Product Code 2	Indicates the 2nd code of the product	Read only	56H
04h	Product Code 3	Indicates the third code of the product	Read only	61H
0Ah	Version	Silicon version	Read only	09H
... not used area				
10h	SPI Mode	Different Modes of the SPI (see SPI mode)	Read only	A1H
11h	WD Type 1	Indicates the type of Watchdog used in the product	Read only	46H
13h	WD bit position 1	Indicates the address of the register containing the WD toggle bit	Read only	40H
14h	WD bit position 2	Indicates the position of the WD toggle bit	Read only	C1H
... not used area				
20h	SPI CPHA	Indicates the polarity and phase of the SPI interface	Read only	55H
3Eh	GSB Options	Options of GSB byte (standard GSB definition)	Read only	00H
3Fh	Advanced OP. Code			

#### 4.3.4 SPI modes

By reading out the <SPI mode> register general information of SPI usage of the device application registers can be read.

**Table 25. SPI mode**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Content
BR	DL2	DL1	DL0	SPI8	0	S1	S0	A1H

#### SPI burst read

**Table 26. SPI burst read**

Bit 7	Description
0	BR disabled
1	BR enabled

The burst read is implemented in this product so this bit is enabled.

#### SPI data length

The SPI data length value indicates the length of the SCK count monitor which is running for all the accesses to the device application registers. In case a communication frame with an SCK count is not equal to the reported one, the device leads to a SPI error and the data is rejected.

The frame length is specified on 3 bits in the SPI mode register located in the ROM part.

The 24-bit SPI communication is implemented in this product so these bits are '010'.

**Table 27. SPI data length**

Bit 6	Bit 5	Bit 4	Description
DL2	DL1	DL0	
0	0	0	Invalid
0	0	1	16bit SPI
0	1	0	24bit SPI
	...		...
1	1	1	64bit SPI

#### Data consistency check (parity/CRC)

For some devices, a data consistency check is required. Therefore, either a parity-check or for very sensitive systems a CRC may be implemented.

It is defined on 2 bits, in the SPI mode register located in the ROM part. A check is then applied on the incoming frame (SDI) while a calculation elaborated on one/multiple bits is done and integrated on the outgoing frame (SDO).

**Table 28. SPI data consistency check**

Bit 1	Bit 0	Description
S1	S0	
0	0	Not used
0	1	Parity used
1	0	CRC used
1	1	Invalid

In case either the parity or the CRC check is implemented it is always located at the end of the communication. The device is equipped with the parity control check. In the Tx device, the parity bit is calculated based on first 23 bits: even number of "1" will set the parity bit to "1", while odd number of "1" will set the parity bit to "0". In the Rx device, parity bit is calculated in the same way and compared with the received one. In case of different parity bit, the received SPI frame is discharged.

## 4.4 Outputs control

Depending on the actual device mode, outputs can be controlled by the SPI register or the Direct Input Dlx.

### SPI register SOCR

In normal mode outputs can be turned ON/OFF, applying Bit[n] = 1/0

[n]: is the related channel, n = 0 for the channel 0, and n = 3 for channel 3

#### Example 1:

Turning ON channel 1 and 2 with turning OFF the others (without taking PWM or phase shifting into account)

**Table 29. Write SOCR 0x13**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Command							
OC1	OC0	Address					
0	0	0	0	1	1	0	1
D15	Data 1						
Not used	Not used	Not used	Not used	SOCR3	SOCR2	SOCR1	SOCR0
x	x	x	x	0	1	1	0
D7	Data 2						
Not used	Not used	Not used	Not used	Not used	Not used	WDTB	Parity
x	x	x	x	x	x	1/0	0

### 4.4.1 Procedure to turn-ON the outputs in PWM operations

#### PWM operation

The status of the output drivers is configured via the SPI output control register (SOCR), the direct input enable bit "DIENCRx" in the OUTCFGRx register and the PWM mode control bits, PWMFCY0 and PWMFCY1, in OUTCFGRx register. The DIENCR selects if the OUTPUTX outputs are controlled also by the direct inputs INx or only by the SOCR. The PWMFCY bit selects the channel frequency. Refer to the following Table 30 for details in normal mode.

**Table 30. Output control truth table**

DIENCR (OUTCFGRx)	INx	SOCRx	DUTYCR	OUTPUTx
0	X	0	X%	OFF
0	X	1	X%	PWM <sup>(1)</sup>
1	L	0	X%	OFF
1	L	1	X%	PWM <sup>(1)</sup>
1	H	X	X%	PWM <sup>(1)</sup>

1. In case of DUTYCR = 100%, PWM = DC ON .

**Note:** In normal mode, outputs can be driven by SPI commands or a combination of SPI commands and direct inputs INx.

**Note:** In fail-safe mode, the outputs are controlled by the direct inputs INx regardless of SPI commands. It is possible to apply the PWM through the Dlx inputs. The PWM unit is not active in fail-safe mode, it is still possible to access to the relevant registers and to configure them.

To turn on channels, information must enter into the following registers:

- Select the PWM frequency by using the two bits PWMFCYx
- Select the PHASE information by using the 5 bits CHPHAx
- Select the switching slope by using the two bits SLOPECRx
- Select the channels configuration Bulb/LED by using the bit CCR
- Select the DUTYCYCLE information by using the 10 bits of the OUTCTRCRx registers
- Select the channel through the dedicated register "SOCR" in the channel control register
- Select the PWM triggering mode by using the single bit PWM\_TRIG of the CTRL register

The PWMSYNC bit resets the internal 12 bits clock counter. This allows to have a known time base and to synchronize different devices among each other.

The signal on the PWM\_CLK is divided internally by a factor from 4096 to 512 depending on the PWMFCY register to generate the base frequency for the output.

- PWM signal is generated by properly selecting 10 of 12 bits on the clock counter. PWM engine has a virtual 10-bit granularity except when PWM divider is set to 512, in this case only a 9-bit granularity is possible (LSB of 10bit generated PWM is fixed to zero). Duty cycle step can be modified with the granularity related to the 9-bit register.

The duty cycle of the output signal is configured for each OUTPUTX with the OUTCTRCR register using 10 bits (MSB first).

- Programming an output duty cycle at 000h results in a 0% duty cycle, it means the channel always OFF depending on the SOCR/DIx bit setting
- Programming an output duty cycle at 3FFh results in a 100% duty cycle (4095/4096), it means the channel always ON when the SOCR/DIx bit is set.
  - In normal mode the outputs are driven according to the SPI register setting and INx pins (DIx in OR with SPI) if the related DIENCR bit is set.

Set PWMSYNC bit in control register "CTRL" (to synchronize the internal PWM counter to the selected channels). The internal PWM counter is 12 bits depth, it is active whatever the state of the channels, if  $V_{DD} > V_{DD\_POR\_ON}$ . The set of **PWMSYNC** bit allows to reset the PWM counter.

Setting the PWM\_TRIG bit in the control register "CTRL" forces the device to calculate the falling edge of the PWM window in advance compared with the end of the PWM period, while with a reset value for this bit, the rising edge of the PWM window will be calculated through a delay at the start of the PWM period.

- PWM\_TRIG = 0 means channel switch on = PWM counter + phase shift counter (see examples 1 to 3)
- PWM\_TRIG = 1 means channel switch off = PWM counter (max) - phase shift counter - duty cycle (see example 4).

The phase shift of the output signal is configured for each OUTPUTx by internally concatenating the CHPHAx 5 bits with '00000' in order to get 10 bits (5 bits + 0000). Granularity of the phase shift is 5 bits. CHPHA = 00000b means a phase shift of 0 (internal 10bit phase shift is 0x000=0000000000b), while CHPHA = 11111b results in a maximum phase shift of 31/32 =(internal 10bit phase shift is 0x3E0=0000000000b). The phase shift is relative to the base frequency of the selected channel. Thus, the exact point in time when the channel switches on also depends on the operating mode of the selected channel.

**Table 31. Phase shift configuration**

Phase shift (%)	5 bits register (H)	10 bits register (H)	Phase shift (ms) PWM = 400 kHz divider = 2048	Phase shift (ms) PWM = 400 kHz divider = 1024	Phase shift (ms) PWM = 400 kHz divider = 512
9.4	03	60	0.481	0.24	0.12
28.1	09	120	1.439	0.719	0.360
46.9	0F	1E0	2.40	1.2	0.6
75	17	2E0	3.84	1.92	0.96
90	1C	380	4.608	2.304	1.152

A change in phase/duty will be taken into account after the next zero-crossing of the PWM counter.

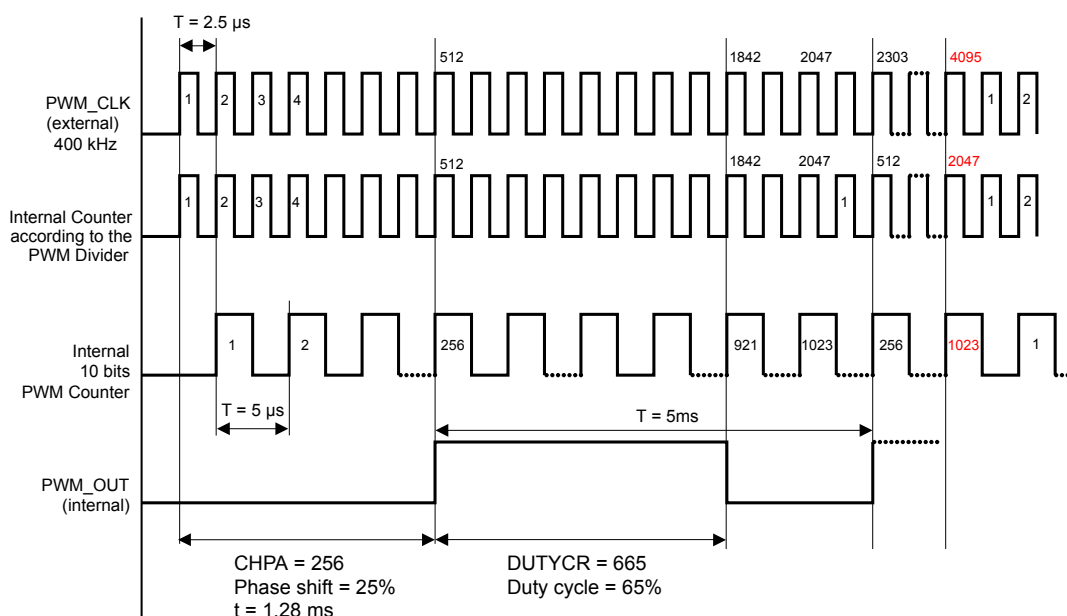
**Note:** *If the frequency on PWM\_CLK is too low ( $f < PWM\_CLK$ ), the device falls back to an internally generated PWM frequency of approximately 400 kHz. In this case, the PWMCLOCKLOW bit in the OUTSRx is set.*

### Example 1

Below, an example with a 65% duty cycle, PWM divider = 2048 and a 25% phase is given with a PWM sampling mode on the rising edge (PWM\_TRIG = 0):

- 65% duty cycle results in a DUTYCRx register content equal to 665 = Ch (65% x 1023 = 665 - 299)
- 25% phase results in a CHPA register content equal to 8 (25% x 31 = 8), equivalent to a content of 256 = 100h for a 10 bit register
- With an input frequency at PWM\_CLK pin of 400 kHz, the output frequency is 195 Hz

**Figure 12. Resulting waveform 1**



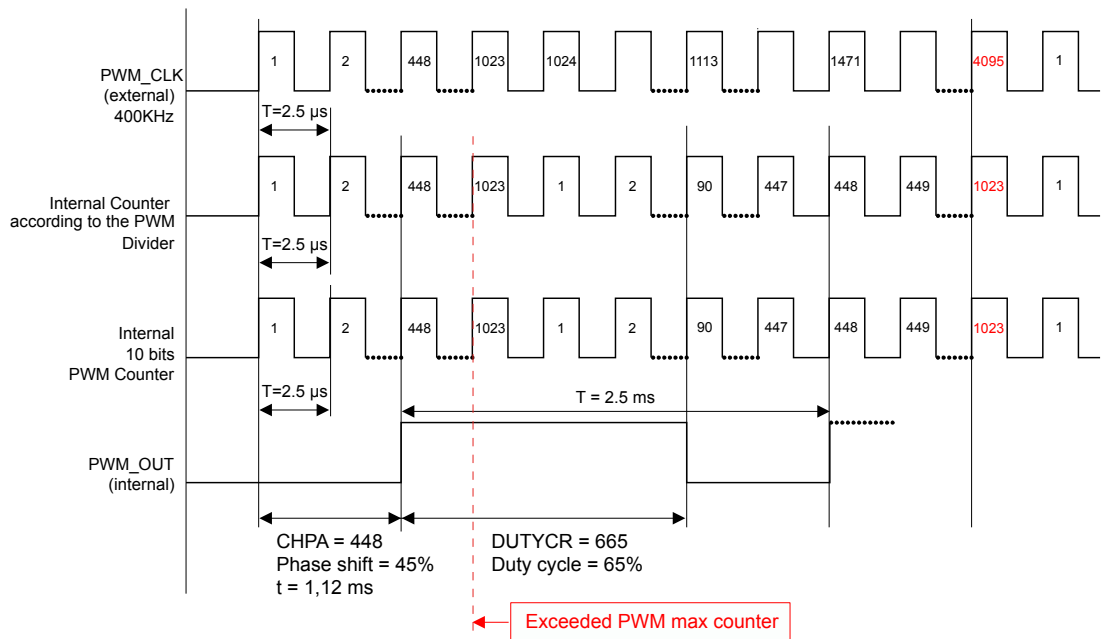
GADG0604171543PS

### Example 2

Below, an example with a 65% duty cycle, PWM divider = 1024 and a 45% phase is given with a PWM sampling mode on the rising edge (PWM\_TRIG = 0):

- 65% duty cycle results in a DUTYCRx register content equal to 665 = Ch (65% x 1023 = 665 - 299)
- 45% phase results in a CHPA register content equal to 14 (45% x 31 = 14), equivalent to a content of 448 = 1C0h for a 10 bit register
- With an input frequency at PWM\_CLK pin of 400 kHz, the output frequency is 390 Hz

**Figure 13. Resulting waveform 2**



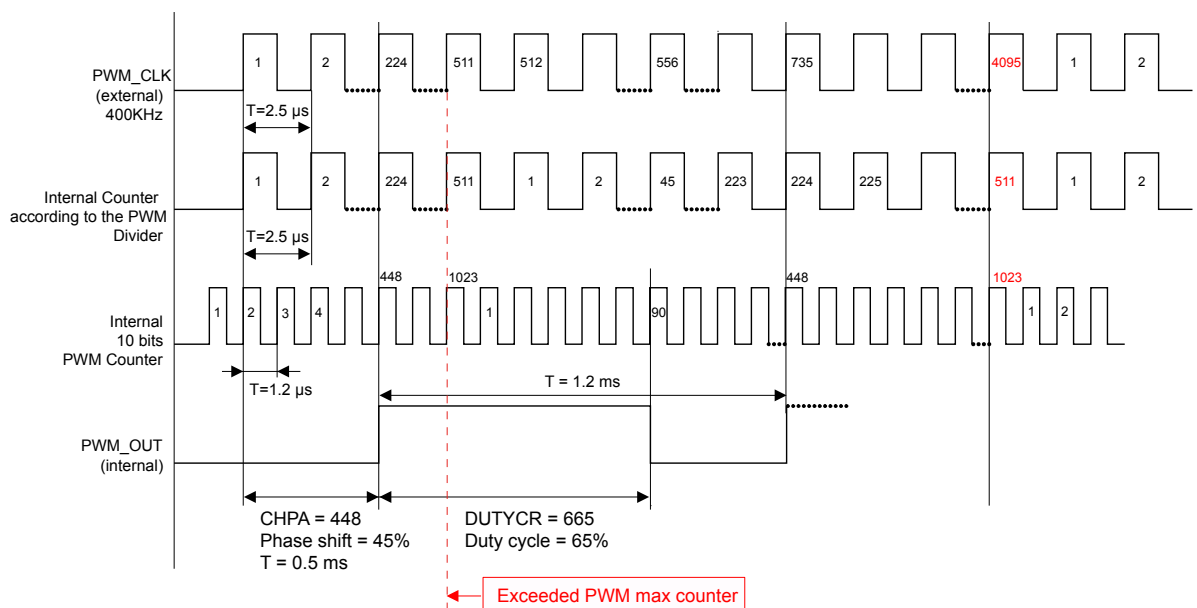
GADG311020171219MT

### Example 3

Below, an example with a 65% duty cycle, PWM divider = 512 and a 45% phase is given with a PWM sampling mode on the rising edge (PWM\_TRIG = 0):

- 65% duty cycle results in a DUTYCRx register content equal to 665 = Ch (65% x 1023 = 665 - 299)
- 45% phase results in a CHPA register content equal to 14 (45% x 31 = 14), equivalent to a content of 448 = 1C0h for a 10 bit register.
- With an input frequency at PWM\_CLK pin of 400 kHz, the output frequency is 781 Hz

**Figure 14. Resulting waveform 3**



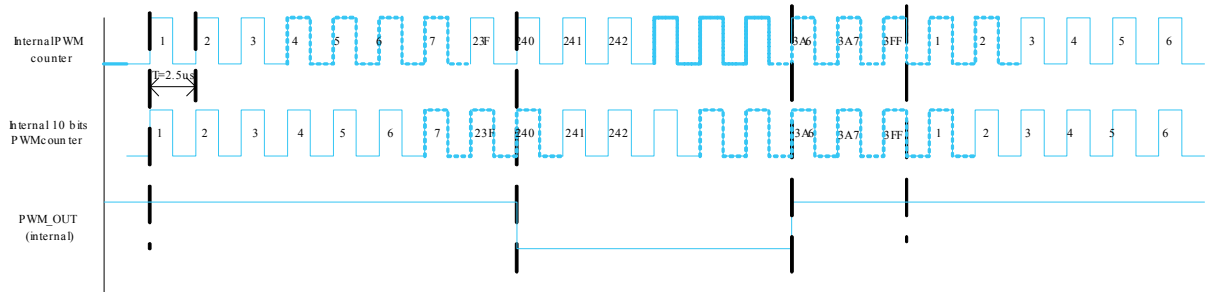
GADG311020171220MT

#### Example 4

Below, an example with a 65% duty cycle, PWM divider = 512 and a 45% phase is given with a PWM sampling mode on the falling edge (PWM\_TRIG = 1):

- 65% duty cycle results in a DUTYCRx register content equal to 665 ( $65\% \times 1023 = 665$ ) equivalent to a content of 299h.
- 45% phase results in a CHPA register content equal to 14 ( $45\% \times 31 = 14$ ), equivalent to a content of 448 = 1C0h for a 10 bit register.
- With an input frequency at PWM\_CLK pin of 400 kHz, the output frequency is  $400/1024 = 390$  Hz
- Due to PWM sampling mode on the falling edge, the PWM window has a start at 3A6h (not  $(299h + 1C0h)$ ) and a stop at 23Fh (not 1C0h).

Figure 15. Resulting waveform 4



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#### 4.4.2 OTP programming

The Direct Input assignment to Output through the OTP programming, is achievable by setting the two dedicated bits per channel in the OTP memory map according to the following table:

**Table 32. OTP memory map**

OTP memory map register (3Eh)	bit 1, bit 0	bit 1, bit 0	bit 1, bit 0	bit 1, bit 0
	00	01	10	11
CH3	DI1	DI0	DI1	OFF
CH2	DI1	DI0	DI1	OFF
CH1	DI0	DI0	DI1	OFF
CH0	DI0	DI0	DI1	OFF

*Note:* "00" represents the default configuration.

The device is provided with the default configuration corresponding to the first column in the previous table (00). A customized OTP configuration, for each channel, is possible by changing the two dedicated bits.

**Table 33. OTP programming**

Dlx assignment							
Ch3		Ch2		Ch1		Ch0	
bit 1	bit 0	bit 1	bit 0	bit 1	bit 0	bit 1	bit 0
n	n	n	n	n	n	n	n

Each Output Status Register (OUTSRx) per channel contains the bits DIOTP1, DIOTP0 which assigns the wanted direct input signal to the channel.

The OTP programming mode can be entered by applying a dedicated procedure in order to ensure a very high safety level for the stored configurations and to prevent from unwanted changing.

Further information about the OTP programming mode is provided in the dedicated user manual.

#### 4.4.3 Procedure to turn-ON the outputs with the direct input Dlx

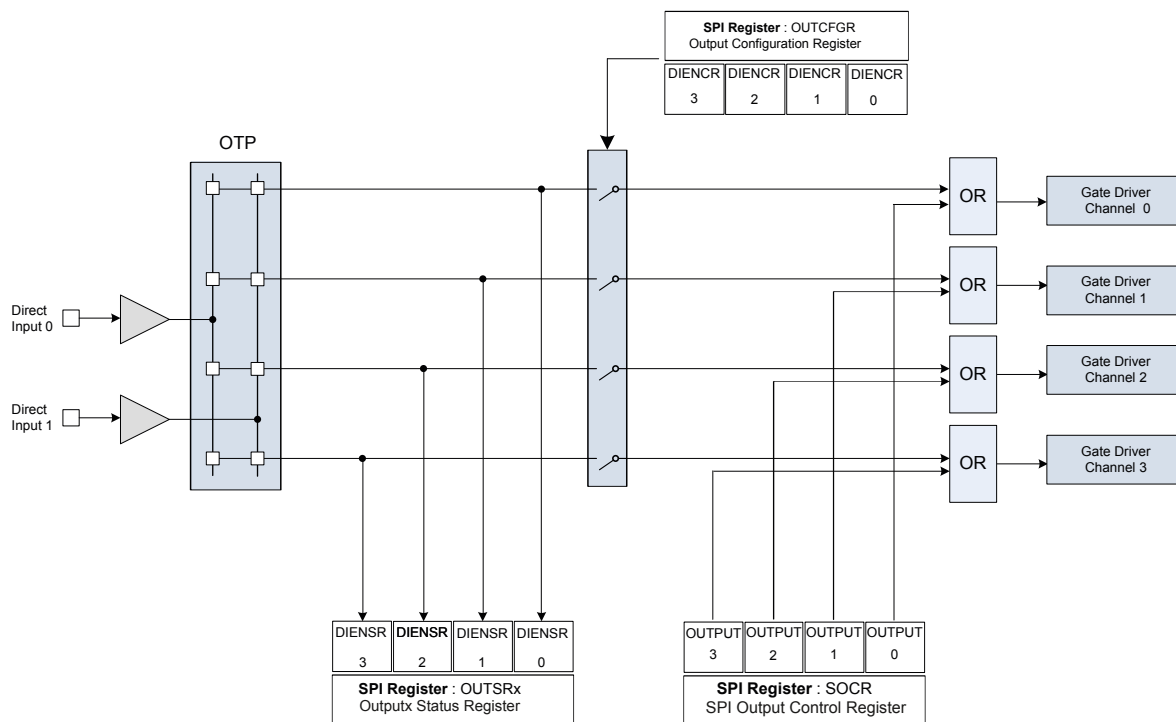
By applying logic level high/low to pin, Dlx turns ON/OFF the associated OTP selected outputs in fail-safe, standby and reset modes. In normal mode, Dlx effect is ORed with SPI configuration when DIENCR bit is set. Then this truth table specifies the output state:

**Table 34. Truth table**

DIECRx	SOCRx	Related Dlx logic status	OUTPUTx state
1	1	X	ON
1	0	L	OFF
1	0	H	ON
0	1	X	ON
0	0	X	OFF

The output channels can be configured to operate in BULB or LED mode using the channel control register (CCR). If the relevant bit in CCR is 0, the output is configured in BULB mode, if it is set to 1, the output is configured in LED mode (default value is 0).



**Figure 16. 4-channel direct input block diagram**


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## 4.5 Output switching slopes control

Output switching slopes are set by the two bits SLOPECR1, 2 in the OUTCFGCRx register (address from 0x08h to 0x0Dh depending on the channel). The switching slopes are shown in the following table:

**Table 35. Switching slopes**

SLOPECRx	Channel 0,1 (V/μs)	Channel 2,3 (V/μs)
00	Standard	Standard
01	Fast	Fast
10	Faster	Faster
11	Fastest	Fastest

## 4.6 Control registers

### OUTCTRCRx

### Outputs Control register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
RESERVED	RESERVED	DUTYCR9	DUTYCR8	DUTYCR7	DUTYCR6	DUTYCR5	DUTYCR4	DUTYCR3	DUTYCR2	DUTYCR1	DUTYCR0	RESERVED	OLOFFCR	WDTB	PARITY
R	R	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	R	RW	RW	R

**Address:** 0x00h to 0x03h  
**Type:** RW  
**Reset:** 0  
**Description:** Outputs Control register

[15:14]	RESERVED
[13:4]	DUTY_CR[9:0]: Set the duty cycle value. Bit 9 (MSB) - Bit 0 (LSB)
[3]	RESERVED
	OLOFFCR: enables an internal pull-up current generator to distinguish between the two faults: open-load OFF-state vs the output shorted to $V_{CC}$ fault.
[2]	1: Pull-up current generator enabled 0: Pull-up current generator disabled
[1]	WDTB: Watchdog toggle bit
[0]	PARITY: Parity bit

## OUTCFGRx

## Outputs configuration register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
SLOPECR1	SLOPECR0	RESERVED	CHPHA4	CHPHA3	CHPHA2	CHPHA1	CHPHA0	SPCR1	SPCR0	PWMFCY1	PWMFCY0	CCR	DIENCR	VDSMASK	PARITY
RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	R

**Address:** 0x08h to 0x0Bh

**Type:** RW

**Reset:** 0

**Description:** The “Output Configuration Register” allows to set the following important parameters for each channel:

- Switching related parameters
  - switching slopes;
  - phase of each channel;
  - PWM ratio;
- Channel configuration
  - BULB/LED mode;
  - Control through SPI / DIx pins;
  - Masking the VDS control at turn-off;
- Diagnostic configuration
  - Set the current sampling point for the digital conversion;

[15:14] SLOPECR[1:0]: Switching slope control bit 1 (MSB) and 0 (LSB)

[13] RESERVED

CHPHA[4:0]:

Set the Channel phase value[4:0]

00000: Resulting phase = 0/32

[12:8] 00001: Resulting phase = 1/32

....

11110: Resulting phase = 30/32

11111: Resulting phase = 31/32

SPCR[1:0]:

Current sampling Point[1:0]

SPCR1:0 SPCR0:0 stop mode: authorizes digital conversion to be launched just before the end of on phase of the selected channel.

[7:6] SPCR1:0 SPCR0:1 START mode: authorizes digital conversion to be launched at each beginning of on phase of the selected channel.

SPCR1:1 SPCR0:0 CONTINUOUS mode: authorizes digital conversion during all on phase of the selected channel.

SPCR1:1 SPCR0:1 FILTERED mode: authorizes digital conversion like CONTINUOUS mode with the use of low pass filter to filter data coming from the conversion. It is useful at low level output current.

PWMFCY[1:0]:

PWM frequency selection[1:0]

[5:4] Each output has a specific ratio for its PWM functionality. This mode is defined through two dedicated bits PWMFCY1 and PWMFCY0 of OUTCFGRx registers.

PWMFCY1:0 PWMFCY0:0 = PWM freq ratio:1024

PWMFCY1:0 PWMFCY0:1 = PWM freq ratio:2048

PWMFCY1:1 PWMFCY0:0 = PWM freq ratio:4096

PWMFCY1:1 PWMFCY0:1 = PWM freq ratio:512

When a combination is selected, the output frequency of the selected channel will be the PWM clock input frequency divided by the defined ratio.

---

CCR: Set the channel configuration (BULB-LED)

[3] 0: Bulb mode

1: LED mode

---

DIENCR: Direct input signal enable in normal mode (according to OTP allocation)

[2] Each output has an OTP programmed direct input assignment for limp-home operation. Any output can be programmed to be always OFF in the limp-home, or according to DI0 pin state or according to DI1 pin state. This programmed assignment can be read from DIOTP bits of OUTSRx status register. When DIENCR bit is set, DIx pin state assigned to the output is ORed with the SOCR/PHASE/DUTYCYCLE combination to control output state. In fail-safe, standby and reset modes applying log .1/0 to pin turns ON/OFF the associated OTP selected outputs.

---

[1] VDSMASK: VDS detection at turn-off masking bit

---

[0] PARITY: parity bit

---

## CHLOFFTCR0

### Channel Latch OFF Timer Control register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
CHLOFFTCR23	CHLOFFTCR22	CHLOFFTCR21	CHLOFFTCR20	CHLOFFTCR13	CHLOFFTCR12	CHLOFFTCR11	CHLOFFTCR10	CHLOFFTCR03	CHLOFFTCR02	CHLOFFTCR01	CHLOFFTCR00	RESERVED	RESERVED	RESERVED	PARITY
RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	R	R	R	R

**Address:** 0x10h

**Type:** RW

**Reset:** 0

**Description:** The output behavior in case of power limitation or thermal shut-down is programmable, as latch-off or Time limited auto-restart (t<sub>blanking</sub>). The default mode is the latch-off one which corresponds to have 0x0h in the register. In Time limited autorestart, when the channel is turned ON, after a transition from 0 to 1 of the corresponding SOCR bit or activation through associated DIx input when DIENCR bit is set, power limitation and thermal shutdown latches are inhibited for a programmed t<sub>blanking</sub> time.

See Programmable blanking window (PBW) for more details.

Two Registers are used for setting the t<sub>blanking</sub> values for each channel:

- CHLOFFTCR0 for channels 2,1,0;
- CHLOFFTCR1 for channel 3.

[15:12]	CHLOFFTCR[23:20]: It configures the output behavior in case of power limitation for the corresponding channel 2.
[11:8]	CHLOFFTCR[13:10]: It configures the output behavior in case of power limitation for the corresponding channel 1.
[7:4]	CHLOFFTCR[03:00]: It configures the output behavior in case of power limitation for the corresponding channel 0.
[3:1]	RESERVED
[0]	PARITY: parity bit

## CHLOFFTCR1

### Channel Latch OFF Timer Control register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	CHLOFFTCR33	CHLOFFTCR32	CHLOFFTCR31	CHLOFFTCR30	RESERVED	RESERVED	RESERVED	PARITY
R	R	R	R	R	R	R	R	RW	RW	RW	RW	R	R	R	R

**Address:** 0x11h

**Type:** RW

**Reset:** 0

**Description:** The output behavior in case of power limitation or thermal shut-down is programmable, as latch-off or Time limited auto-restart (t<sub>blanking</sub>). The default mode is the latch-off one which corresponds to have 0x0h in the register. In Time limited autorestart, when the channel is turned ON, after a transition from 0 to 1 of the corresponding SOCR bit or activation through associated DIx input when DIENCR bit is set, power limitation and thermal shutdown latches are inhibited for a programmed t<sub>blanking</sub> time.

See Programmable blanking window (PBW) for more details.

Two Registers are used for setting the t<sub>blanking</sub> values for each channel:

- CHLOFFTCR0 for channels 2,1,0;
- CHLOFFTCR1 for channel 3.

[15:12]	CHLOFFTCR[53:50]: RESERVED.
[11:8]	CHLOFFTCR[43:40]: RESERVED.
[7:4]	CHLOFFTCR[33:30]: It configures the output behavior in case of power limitation for the corresponding channel 3.
[3:1]	RESERVED
[0]	PARITY: parity bit

The blanking window duration in case of power limitation or thermal shutdown events can be set according to the following table:

**Table 36. Programmable t<sub>blanking</sub> values**

CHLOFFTCRx3	CHLOFFTCRx2	CHLOFFTCRx1	CHLOFFTCRx0		
0	0	0	0	0x0	latch OFF cfg (default)
0	0	0	1	0x1	16ms
0	0	1	0	0x2	32ms
.	.	.	.		
1	1	1	0	0xE	224ms
1	1	1	1	0xF	240ms

## SOCR

### Channel control register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
RESERVED	RESERVED	RESERVED	RESERVED	SOCR3	SOCR2	SOCR1	SOCR0	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED	WDTB	PARITY
R	R	R	R	RW	RW	RW	RW	R	R	R	R	R	R	RW	R

**Address:** 0x13h

**Type:** RW

**Reset:** 0

**Description:** The SOCR register is used to turn ON/OFF the related channel. The WDTB bit that must be toggled within  $t_{WDBT}$  (watchdog timeout) to avoid entering in fail-safe mode. This bit is already present in the output control register and it is duplicated in the SOCR register to simplify the SPI usage.

[15:12]	RESERVED
	SOCR3 bit controls output state of channel 3
[11]	1 - output enabled 0 - output disabled
	SOCR2 bit controls output state of channel 2
[10]	1 - output enabled 0 - output disabled
	SOCR1 bit controls output state of channel 1
[9]	1 - output enabled 0 - output disabled
	SOCR0 bit controls output state of channel 0
[8]	1 - output enabled 0 - output disabled
[7:2]	RESERVED
[1]	Watchdog toggle bit
[0]	PARITY: parity bit

**CTRL**
**Control register**

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
GOSTBY	UNLOCK	CTDTH1	CTDTH0	EN	PWM_TRIG	RESERVED	RESERVED	RESERVED	LOCKEN4	LOCKEN3	LOCKEN2	LOCKEN1	LOCKEN0	PWMSYNC	PARITY
RW	RW	RW	RW	RW	RW	R	R	R	RW	RW	RW	RW	RW	W	R

**Address:** 0x14h

**Type:** RW

**Reset:** 0

**Description:** Control register

GOSTBY: Go to standby.

- [15] It is necessary to do 2 write accesses to enter standby:
1. Write UNLOCK = 1
  2. Write GOSTBY = 1 and EN = 0

UNLOCK: unlock bit

- [14] UNLOCK bit allows protected SPI transactions. It means that the next SPI communication will automatically clear this bit and prevent any change of protected data (like slope control or BULB/LED mode for example). As a consequence, modifying a protected data requires to set UNLOCK bit in a first communication and write the protected data during the next communication.

CTDTH[1:0]: Case thermal detection threshold. These bits allow to configure the case thermal detection of the device. Three temperature thresholds are available by programming these two bits.

- [13:12]
1. CTDTH1:0 CTDTH0:0 = detection temperature:120 °C
  2. CTDTH1:0 CTDTH0:1 = detection temperature:130 °C
  3. CTDTH1:1 CTDTH0:X = detection temperature:140 °C

EN: enter normal mode

- [11]
- 1: normal mode
- 0: fail-safe mode
- It is necessary to do 2 write accesses to enter normal mode:
1. Write UNLOCK = 1
  2. Write EN = 1

PWM\_TRIG: PWM triggering mode

- [10]
- 0: PWM trigger according to the rising edge of PWM period and phase shift configuration
- 1: PWM trigger according to the falling edge of PWM period and phase shift configuration

- [9:7] RESERVED

LOCKEN[4:0]: Protected transaction mode

LOCKEN4: Lock enable for slope control SLOPECRx

LOCKEN3: Lock enable for BULB/LED mode CCRx

LOCKEN2: Lock enable for phase shift CHPHAx

- [6:2]
- LOCKEN1: Lock enable for configurable blanking time CHLOFFTCRx
- LOCKEN0: Lock enable for PWM clock synchronization

When the bit is set (LOCKENx = 1), it is used to have a protected transaction:

- setting UNLOCK bit
- modify the relevant configuration register

When LOCKENx=0 (reset value), the related configuration registers are altered with a simple write command.



- 
- [1] PWMSYNC: PWM clock synchronization.  
PWMSYNC =1 clears PWM internal counter. It automatically resets at next SPI communication
- 
- [0] PARITY: parity bit
-

## OUTSRx

### Output Status Channels 0 to 3 register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
DIENSR	DIOTP1	DIOTP0	CHFBSRx	VDSFSRx	STKFLTRx	OLPUSRx	CHLOFFSRx	RST	SPIE	PWMCLOCKLOW	VCCUV	RESERVED	RESERVED	RESERVED	PARITY
R	R	R	RC	RC	RC	R	R	RC	RC	RC	R	R	R	R	R

**Address:** 0x20h to 0x23h

**Type:** RC

**Reset:** 0

**Description:** The output status register reports the status of the selected channel based on the configuration register and in case of fault condition.

[15]	DIENSR: Direct input status, image of associated DI logic level according to OTP allocation.
[14]	DIOTP1: Associated DIx input description bit 1
[13]	DIOTP0: Associated DIx input description bit 0
	CHFBSRx: Channel feedback status.
[12]	Channel feedback status. Combination of Power limitation, OT, OVERLOAD detection (VDS at turn-off). The CHFBSRx provides a logical "OR" combination of VDS (overload), PL (power limitation), OT (overtemperature) failure flags related to OUTPUTx, and it's cleared by a read and clear command.
	VDSFSRx: VDS Feedback status.
[11]	This bit is '1' if VDS is high at turn-off, indicative of a potential overload condition
[10]	STKFLTRx: Output stuck to V <sub>CC</sub> /open-load off state status.
[9]	OLPUSRx: Output pull-up generator status.
[8]	CHLOFFSRx: Channel Latch-off status. This bit is set when overload blanking time has elapsed and channel is latched off.
[7]	RST: Chip reset
[6]	SPIE: SPI error
[5]	PWMCLOCKLOW: PWM clock frequency too low.
[4]	VCCUV: V <sub>CC</sub> undervoltage
[3:1]	RESERVED
[0]	PARITY: parity bit

## ADCxSR

### Digital Current Sense register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
RESERVED	RESERVED	ADCxSR9 (MSB)	ADCxSR8	ADCxSR7	ADCxSR6	ADCxSR5	ADCxSR4	ADCxSR3	ADCxSR2	ADCxSR1	ADCxSR0 (LSB)	RESERVED	SOCR <sub>x</sub>	UPDTSR	PARITY
R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R

**Address:** 0x28h to 0x2Bh

**Type:** R

**Reset:** 0

**Description:** The register contains the digital value of the current flowing on the selected channel. It reports the result of the digital current conversion. It is updated according to the modes set by the two bits (SPCR1 and SPCR0) of the OUTCFGR<sub>x</sub>.

[15:14]	RESERVED
[13:4]	ADCxSR[9:0]: The 10 bit register contains the digital value of OUTPUT <sub>x</sub> current.
	ADCxSR9 (MSB)
	ADCxSR0 (LSB)
[3]	RESERVED
	SOCR <sub>x</sub> : SOCR Bit controls output state of channel x.
[2]	1 - output enabled
	0 - output disabled
[1]	UPDTSR: updated status bit.
	This bit is set when a value is updated and cleared when register is read.
[0]	PARITY: parity bit

## ADC9SR

## Digital Case Thermal Sensor Voltage register

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
RESERVED	RESERVED	ADC9SR9 (MSB)	ADC9SR8	ADC9SR7	ADC9SR6	ADC9SR5	ADC9SR4	ADC9SR3	ADC9SR2	ADC9SR1	ADC9SR0 (LSB)	RESERVED	RESERVED	UPDTSR	PARITY
R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R

**Address:** 0x31h

**Type:** R

**Reset:** 0

**Description:** The register contains the result of the digital conversion of the case temperature.

[15:14] RESERVED

ADC9SR[9:0]: The 10 bit register contains the digital value of case temperature sensor voltage.

[13:4] ADC9SR9 (MSB)

ADC9SR0 (LSB)

$T_{CASE} (typ.) = 401.8 \text{ }^{\circ}\text{C} - 1.009 * \text{ADC9SR}[13:4]$

[3:2] RESERVED

UPDTSR: updated status bit.

[1] This bit is set when a value is updated and cleared when register is read.

[0] PARITY: parity bit

## OTP memory map (Reserved)

OTP is automatically read into registers at Reset.

OTP memory map contains Direct Inputs assignment to outputs,

Direct Inputs assignment data (2 bits per channel):

**Table 37. OTP memory map (reserved)**

Dlx assignment bit 1	Dlx assignment bit 0	CH3	CH2	CH1	CH0
0	0	DI1	DI1	DI0	DI0
0	1	DI0	DI0	DI0	DI0
1	0	DI1	DI1	DI1	DI1
1	1	OFF	OFF	OFF	OFF

## 5 Diagnostic

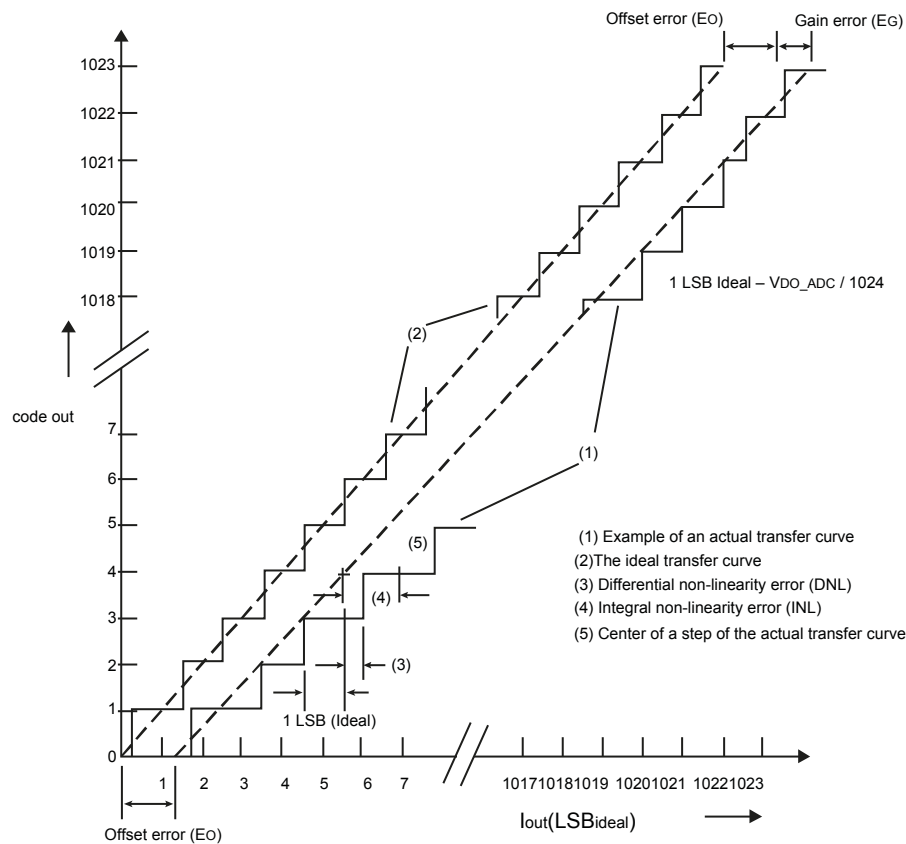
Device is capable of providing digital diagnostic information through the SPI interface.

### 5.1 Digital current sense diagnostic

#### 5.1.1 ADC characteristics

Here are the “Differential non linearity” and “Integral non Linearity” typical curves for the 10-bit ADC converter.

**Figure 17. ADC characteristics and error definition**



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#### 5.1.2 ADC operating principle

The device provides a 10-bit successive approximation register (SAR) analog-to-digital converter. It is used to provide a digital information about the current sense feedback proportional to the output current and the temperature read by the internal sensor. An integrated LP (progressive average) filter can be used to filter data coming from the ADC conversion reducing the effect of random noise coming from the analog current sense amplifier.

**Note:** *The internal ADC is able to work in both normal and fail-safe conditions.*

The integrated ADC control logic is designed to lead to a good 10-bit approximation of current sense/temperature feedback.

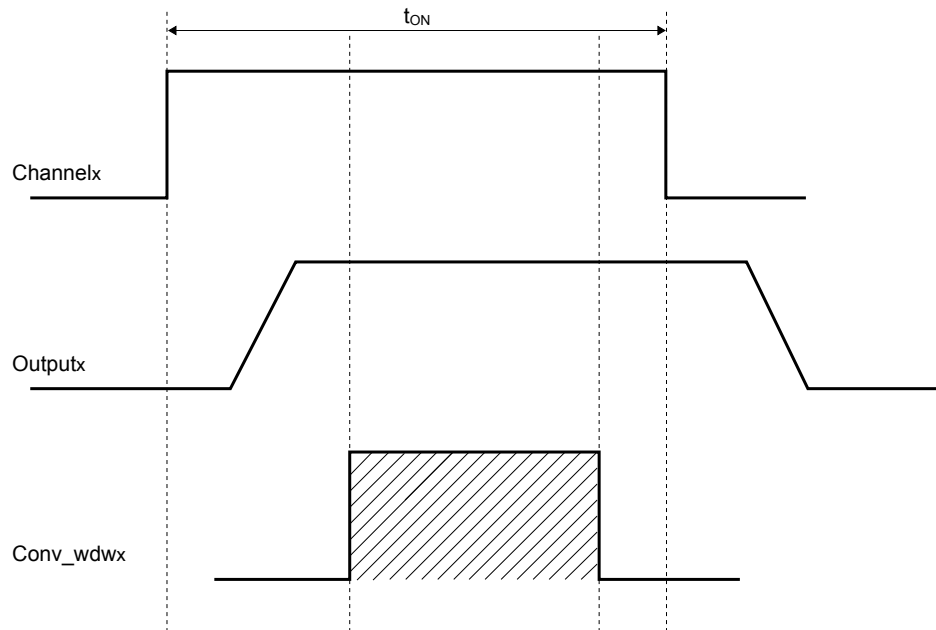
After each conversion, an updated bit “UPDTSR” is set to advise about new conversion data. This bit is reset after the Read process of the dedicated RAM register.

The data is maintained in the register until the next conversion results are available. The ADC register is refreshed at the end of each conversion and maintained during the conversion of the current sample. The data is converted on the 10-bit register, the formula is equal to:

$I_{out\_conv} = \text{data} (10\text{bit})/K;$

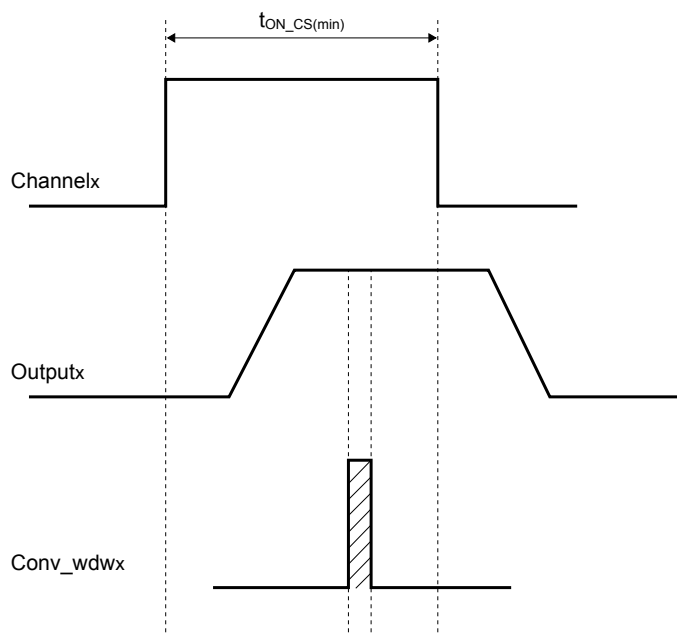
An analog multiplexer has been implemented to connect the different channels to the amplifier and the ADC block. Due to the current sense amplifier settling time when switching from the current sense mode of one channel, to the current sense mode of another channel, a priority management is implemented to control the time when the data conversion can be done in a safe/stable way and to arbitrate the concurrent ADC sampling requests (see next figures).

**Figure 18. Conversion window generation**



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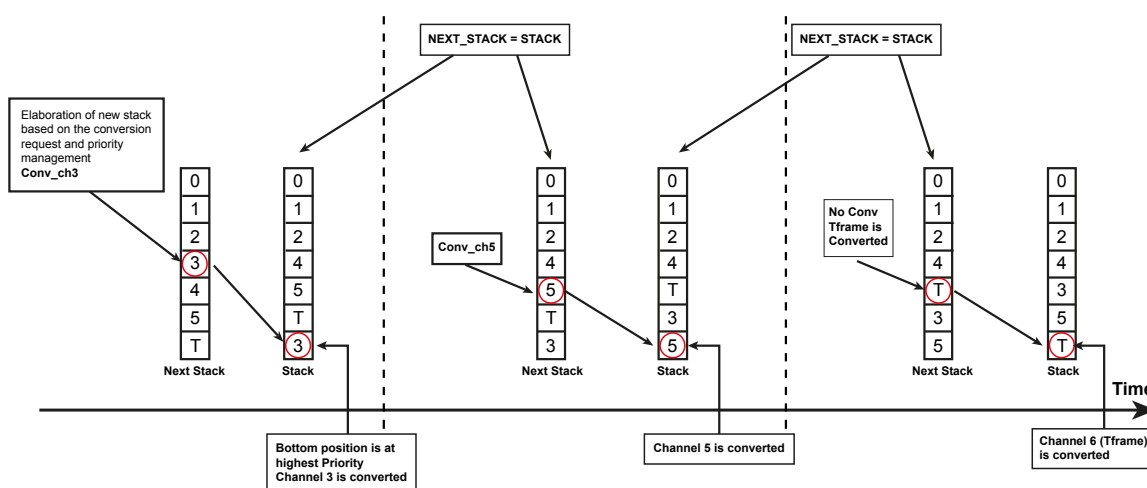
A minimum conversion time ( $t_{ON\_CS(min)}$ ) is defined to allow the signal stabilization at the input of the ADC converter and considering the sampling time. The user should manage the phase shift in a way that maximum two channels can be sampled in the same time window.

**Figure 19. Minimum ON time for digital current sense availability**


GADG311020171223MT

The sequence of channels to be converted is managed through an internal stack:

- Stack size is equal to the number of channels plus frame temperature sensor
- A conversion of the selected channel is done based on the information stored at the end of the stack (see [Figure 20. Channel's sequence internal stack](#))
- After the reset of the device or when no channels are active, the conversion of the frame temperature sensor is done continuously
- When the conversion of a channel\_x has to start, the channel\_x is moved to the end of the stack while the other remaining channels are moved up

**Figure 20. Channel's sequence internal stack**


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**Note:** The figure above shows an example of how the priority is managed through a stack in a six channels device.

### 5.1.3 Registers

The results of the digital conversion are stored in the “Digital Current Sense Registers” Two registers are used for the digital conversion of the output current and case sense temperature respectively:

- **ADCxSR** (address from 28h to 2Dh) - for digital Outputx current (one register x channel)
- **ADC9SR** (address 31h) - Digital case temperature sensor voltage sense register

**Table 38. Registers**

Register name	Bit 15,14	Bit 13..4	Bit 3	Bit 2	Bit 1	Bit 0
ADCxSR 28h to 2Dh	Not Used	Digital Value of OUTx current	Not Used	SOCRx Possibility to control the Outx state (Read only)	UPDTSR Updated status bit. It is set when value is updated and cleared when register is read	Parity
ADC9SR 31h	Not Used	Digital Value of case temperature sensor voltage	Not Used	Not Used	UPDTSR Updated status bit. It is set when value is updated and cleared when register is read	Parity

### 5.1.4 Synchronous, Asynchronous Modes

#### 5.1.4.1 Normal mode

The ADC conversion can work in 4 different sampling modes (start, stop, continuous or filtered) according to the table below. Two bits per channel “SPCR1” and “SPCR0” allocated in the Output Configuration Register “OUTCFGRx”, allow 4 different sampling modes:

**Table 39. Sampling mode configuration**

SPCR1	SPCR0	Sampling mode
0	0	STOP Mode
0	1	START Mode
1	0	CONTINUOUS Mode
1	1	FILTERED Mode

#### 5.1.4.2 Synchronous mode

##### Synchronous Mode in PWM condition

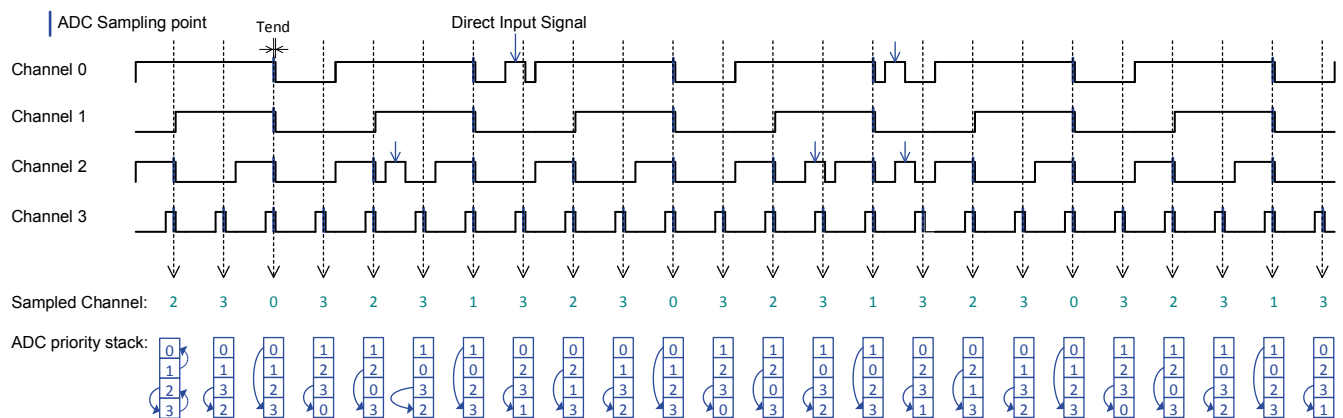
- Sampling is done according to the PWM rising and falling edge (see [Figure 21. Sequence of channels](#)). See [Table 39. Sampling mode configuration](#) for more details about the registers configuration.
- The sampling priority will be always allocated at higher priority

##### Registers configuration

- SPCR10=0h: Synchronous triggered by rising edge on internal PWM. Conversion is executed on the rising edge of the conversion window (see [Figure 21. Sequence of channels](#)).
- The ADC real sampling is managed to trigger the sampling point with margin versus falling edge.
- SPCR10=1h: Synchronous triggered by falling edge of the internal PWM signal. Conversion is executed on the falling edge of the conversion window (see [Figure 21. Sequence of channels](#)).

*Note:* The ADC real sampling is managed to trigger the sampling point with margin versus rising edge.



**Figure 21. Sequence of channels**


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### 5.1.4.3 Asynchronous mode

In asynchronous mode the ADC result register is continuously refreshed, provided that the channel is commanded either through the direct input signal or the SOCR register. Conversion is executed during the complete conversion window except for the priority arbitration.

Since the ADC register is continuously refreshed, its conversion priority is always lower than the sampled channels.

Once the PWM counter reaches a value for which synchronous diagnostic of another channel is requested, the internal MUX switches to this channel and serves the ADC sampling request (channels in synchronous mode have higher priority compared to those in asynchronous mode). Once this sampling is completed the MUX switches back to the asynchronous sampling channel, provided that no higher priority sampling requests from other channels occur. If two or more channels are configured in asynchronous mode, the MUX will sequentially switch through those channels, always interrupted when higher priority synchronous sampling requests occur.

The thermal case sampling has always low priority for the ADC conversion and so it can be interrupted by any channel in sample mode.

#### Registers configuration:

SPCR10 = 2h and SOCRx = 1: asynchronous with continuous sampling:

Asynchronous mode, the ADC result register is continuously refreshed, provided that the channel is commanded either through the direct input signal or the SOCR register. Conversion is executed during the complete conversion window except for the priority arbitration. Since the ADC register is continuously refreshed, its conversion priority is always lower than sampled channels.

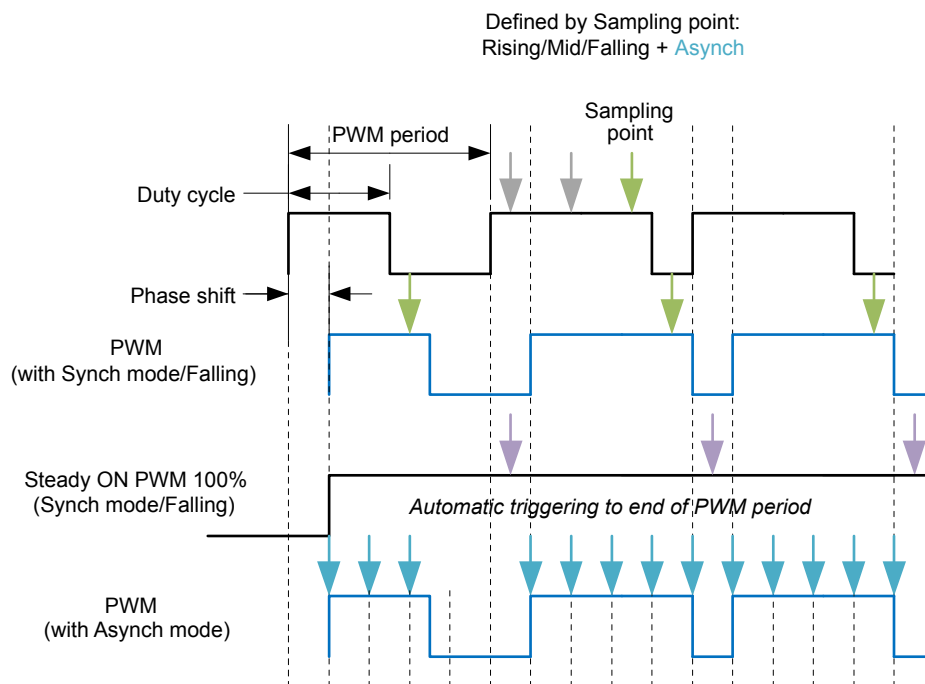
SPCR10 = 3h and SOCRx = 1: asynchronous with continuous sampling and digital LP filter:

- The integrated LP filter is activated
- This component filters data coming from the ADC conversion reducing the effect of random noise coming from the analog current sense amplifier

SPCR10 = 3h, SOCRx = x and Dlx = high: if a channel is commanded off through SOCR, but commanded on through the direct input, the asynchronous sampling mode is forced - asynchronous with continuous sampling.

The thermal case sampling has always low priority for the ADC conversion, and so it can be interrupted by any channel in simple mode. Thermal case conversion is always in asynchronous continuous mode. In fail-safe condition the ADC conversion is always in asynchronous/continuous mode.

- Conversion is executed during the complete conversion window.
- No priority management is applied, channels are converted according to their position in the stack. No interruption and no priority management are possible. In the case of multiple channels active at the same time, the conversion starts with the first one in the stack.

**Figure 22. Asynchronous with continuous sampling**


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#### 5.1.4.4

#### Sampling concept

- PWM mode (internal engine) → All the synchronous modes are available (start, stop, continuous or filtered)
- DC mode (internal engine) → ADC works in Continuous Mode. The conversion window follows the channel control input signal
  - DC mode by/without DI: No difference, since this condition is equivalent to PWM with 100% of duty (the sampling will be always in continuous mode).
- PWM mode by DI (external source) → the DIx information is combined (O-red) with the channel control signal. Sampling will be executed according to the PWM mode settings.
  - With SPCR<sub>x</sub>=2h,3h, sampling is possible (continuous/filtered mode).

#### 5.1.4.5

#### Synchronous mode in DC condition (PWM with 100% duty cycle) => equivalent to asynchronous mode

This mode is equivalent to the asynchronous mode.

**Table 40. ADC configuration registers**

SOCrx	DIx	DutyCrx	SPCR1,0	Conversion mode	Feedback type
1	X	X	00	Synchronous triggered by falling edge on the internal PWM signal	Output current
1	X	X	01	Synchronous triggered by rising edge of the internal PWM signal	Output current
1	X	X	10	Asynchronous with continuous sampling	Output current
1	X	X	11	Asynchronous with continuous sampling and digital LP filter	Output current
0	1	X	X	(Fail-safe mode) Asynchronous with continuous sampling	Output current
X	X	X	X	Tframe conversion (Always lower priority than current sampled modes)	Tframe sensor voltage

## 5.2 Integrated LP (progressive average) filter

In asynchronous mode, when the filtered mode is selected through the dedicated bits “SPCR1 = 1” and “SPCR0 = 1”, the integrated LP filter is activated. This component filters the data coming from the ADC conversion reducing the effect of random noise coming from the analog current sense amplifier.

Features of the integrated LP filter:

- First order LP filter on 16 samples
- First result after one sample with progressive averaging of 16 successive samples

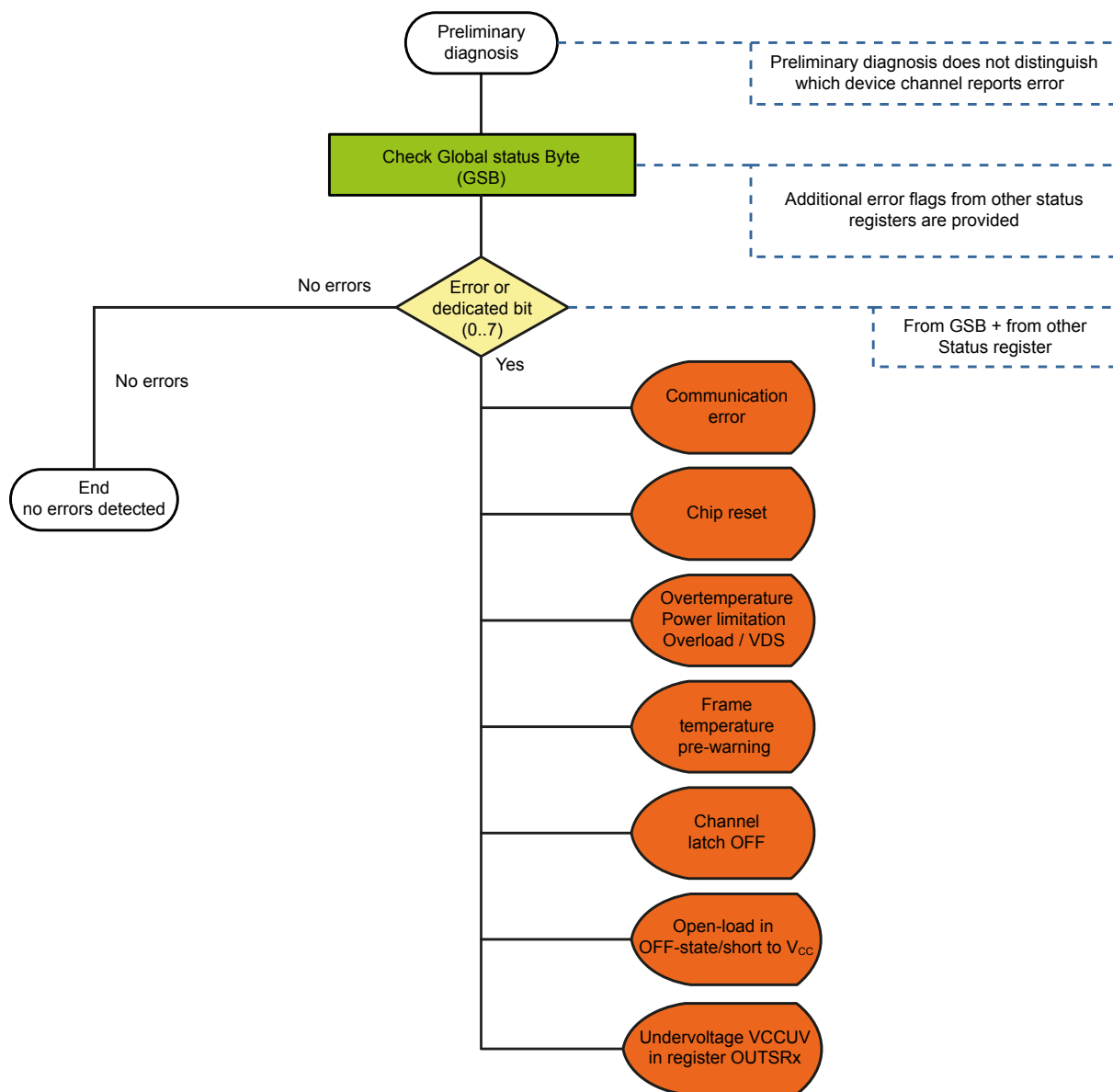
$$data(N) = \left( data(N-1) \cdot \frac{15}{16} \right) + data_i/16$$

- Continues to accumulate samples during the PWM operation
- Keeps digitalized value when the channel is turned off

## 5.3 Digital diagnostic

The global status byte (GSB) provides the preliminary status of the device during the SPI communication with the device. It informs about the device actual mode (normal/fail-safe).

By reading the additional status registers, more detailed information is provided. Status information is stored in the status registers.

**Figure 23. Status registers**


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### 5.3.1 Status registers

**Table 41. Status registers**

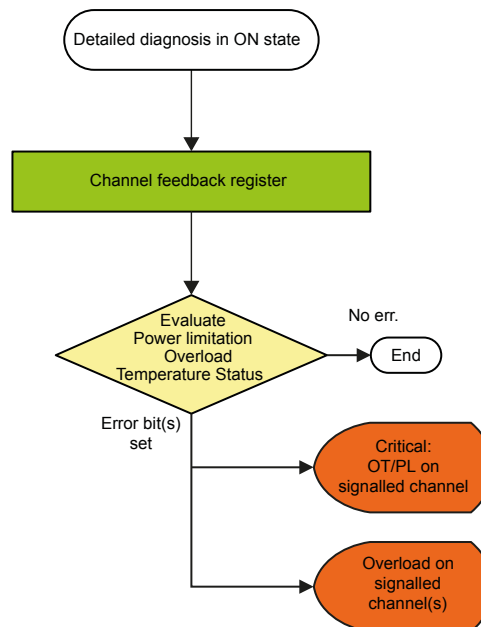
Address	Name	Access	Description
20h to 23h	OUTSR	Read/clear	Outputs status Register from 0x20h (channel 0) to 0x23h (channel 3) (see register map for detailed description)
28h to 2Bh	ADCxSR	Read	Digital current sense registers. from 0x28h (channel 0) to 0x2Bh (channel 3)
31h	ADC9SR	Read	Digital case temperature sensor current sense register.

## 5.4 Overload (VDS high voltage, overload (OVL))

During low duty cycle PWM operation on a shorted load, the ON time is normally too short to allow power limitation or overtemperature detection, and the ADC output does not report the current flowing on the channel. In this situation, the detection of the overload condition is quite difficult. To overcome this, the voltage drop on the PowerMOS ( $V_{DS}$ ) is measured every time the channel is turned OFF. If  $V_{DS}$  (voltage across PowerMOS output stage) exceeds the threshold defined by the parameter  $V_{DS\_OVL}$ , an overload condition is detected. The corresponding bit in the overload status register VDSFSR (address from 0x20h to 0x23h depending on the channel) of the OUTSRx register is set (see the [Section 5.11: VDS feedback status register “VDSFSR” in OUTSRx register](#)).

Consequently, the bit 4 in the Global Status Byte is set if it is not masked in the CONFIG register through the dedicated “VDSMASK” bit. The VDSFSR is a warning and the channel can be switched on again even if the VDSFSRx bit is set. The VDSFSRx bit remains unchanged until a read and clear command on VDSFSR is sent by the SPI or until the output is turned off the next time, when the VDS is evaluated again. In case of low duty cycle PWM operation (i.e. 3% typical at 200 Hz in Bulb mode), if the output channel is switched ON for a very short time,  $V_{DS}$  might be greater than a threshold defined by the parameter  $V_{DS\_OVL}$  even if the output is not in overload state so that a false warning is issued.

**Figure 24. Diagnostics flowchart for digital overload detection**



GADG1004171611PS

Please refer to [Section 4.3.1: Global status byte description](#), [Section 5.8: CHFBSR: channel feedback status bit in OUTSRx register](#) and [Section 5.11: VDS feedback status register “VDSFSR” in OUTSRx register](#).

## 5.5 Open-load ON-state detection

The open-load ON-state is performed by reading the digital current sense. In case the output is on and the reported digital current sense value is below the requested threshold, the open-load condition can be reported.

## 5.6 Open-load OFF-state detection

After the channel is completely OFF, if the output voltage  $V_{out}$  exceeds the open detection threshold  $V_{OL}$ , an open-load in OFF state / Stuck to  $V_{CC}$  event is reported.

As a consequence, the corresponding bit STKFLTR in the OUTSRx register (address 0x20h to 0x23h) is set, the  $OL_{OFF}$  bit in the global status Register and the global status bit NOT are set accordingly.

The STKFLTRx bit is set in OFF-state if  $V_{OUT} > V_{OL}$  and the  $t_{DOLOFF}$  (turn-off delay time) has elapsed. It gives information about open-load or a stuck to  $V_{CC}$ , which depends on the configuration of the OUTCTRCRx OLOFFCR bit. The bit is continuously refreshed in OFF-state and it is latched during ON-state. In order to clear the bit in ON-state, it is necessary to send a Read and Clear command.

STKFLTRx = 1: open-load in OFF-state or stuck to  $V_{CC}$  condition occurred for OUTPUTx

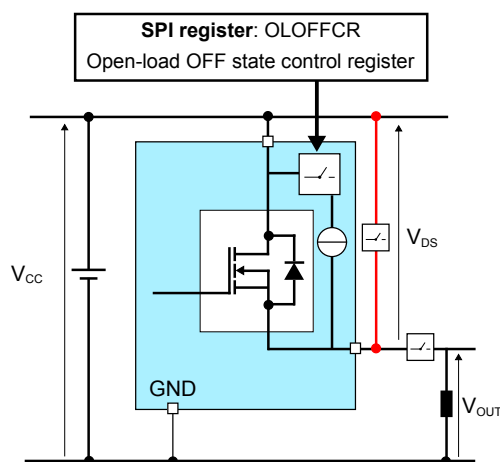
STKFLTRx = 0: no fault detected

To avoid false detection, the diagnosis starts after the turn-off of a channel with an additional delay  $t_{DOLOFF}$ . To distinguish between an open-load OFF-state event and a short to  $V_{CC}$  condition, an internal pull-up current generator can be enabled for each channel by setting the corresponding open-load off state bit  $OL_{OFFCR}$  (bit 2) in the outputs control registers "OUTCTRCRx" (address from 0x00h to 0x03h depending on the channel).

The activated pull-up current generators are active in normal mode, in fail-safe mode, and in standby mode.

Differently, in sleep mode 2 the current generators are switched off. The register contents, however, are saved also in sleep mode 2, consequently the current generators are reactivated after a return to standby or a wake-up to fail-safe mode. A hardware reset ( $V_{DD} < V_{DD\_POR\_OFF}$ ) or a software reset (command byte = FFh) clears all register contents and hence the current generators are switched off.

**Figure 25. Open-load OFF-state detection**

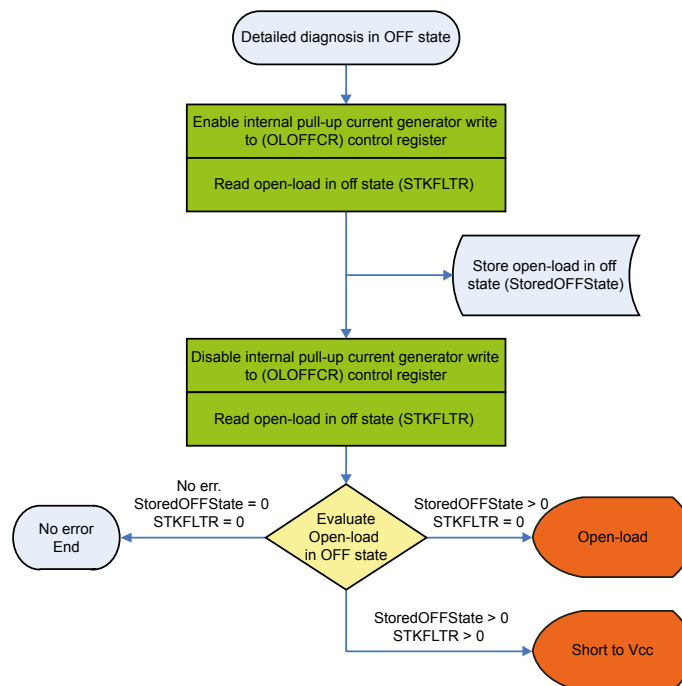


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**Table 42. STKFLTR state**

	With internal pull-up generator	Without internal pull-up generator
Case 1: load connected	"0" / no fault	"0" / no fault
Case 2: no load	"1" / fault	"0" / no fault
Case 3: output shorted to $V_{CC}$	"1" / fault	"1" / fault

**Figure 26. Diagnostic flowchart for open-load off-state respectively stuck to V<sub>CC</sub> failure**



GADG1004171628PS

## 5.7 DIENSR: direct input status bit in OUTSRx register (address from 0x20 to 0x23)

DIENSR bits read back the logic level of the Dlx input assigned through OTP to the specific channel. The Dlx pins are used to control the outputs channel in case of fail-safe condition or in normal operation if the related "DIENCR" bit set.

## 5.8 CHFBSR: channel feedback status bit in OUTSRx register

The CHFBSRx provides a logical "OR" combination of VDS (overload), PL (power limitation), OT (overtemperature) failure flags related to OUTPUTx. The contributions of the VDS failure flags to the channel feedback status register and global status byte can be maskable through the VDSmaskx bit in the OUTCFGRx registers.

CHFBSRx = 1: Channel OUTPUTx on failure

CHFBSRx = 0: Channel OUTPUTx no failure

The bits are refreshed continuously in ON-state and latched in OFF-state.

In order to clear the bit in OFF-state, it is necessary to send a read and clear command.

## 5.9 Open-load in OFF-state / Stuck to $V_{CC}$ status bit “STKFLTR” in OUTSRx register

This bit is set when the output voltage of the selected channel exceeds the detection threshold at turn-OFF.

The STKFLTRx bit is set in OFF-state if  $V_{OUT} > V_{OL}$  and the  $t_{DOFF}$  (turn-off delay time) has elapsed. It gives information about open-load or a stuck to  $V_{CC}$  which depends on the configuration of the OUTCTCRx OLOFFCR bit register. The bit is continuously refreshed in OFF-state and it is latched during ON-state. In order to clear the bit in ON-state it is necessary to send a Read and Clear command.

STKFLTRx=1: open-load in OFF-state or stuck to  $V_{CC}$  condition occurred for OUTPUTx

STKFLTRx=0: no fault detected

## 5.10 Channels latch-off status bit “CHLOFFSR” in OUTSRx register

The CHLOFFSR bit is set as soon as there is a fault condition identified as Power-limitation or over-temperature. Latch OFF flags register. There is one bit per channel.

In case a latch-off condition occurs, the faulty channel can be reactivated by performing a write operation to the corresponding bits of the latched channel in CHLOFFTCRx register. A SW reset event clears the content of the register.

## 5.11 VDS feedback status register “VDSFSR” in OUTSRx register

This bit represents the VDS feedback status. The device is equipped with one VDS bit per channel.

The bit is set in case an overload condition is detected on the related channel. The bit is set independently of the OT.PL. flag.

The VDSFSRx bit is set if, at the instant when the channel is commanded off or is latched off, the  $V_{CC} - V_{OUT}$  voltage drop exceeds the  $V_{DS\_OVL}$  threshold. The bit is latched until the next turn OFF. In order to clear the bit it is necessary to send a read and clear command.

The VDSFSRx bit is set to:

1: overload event occurred for OUTPUTx

0: no fault detected

**Note:** *As the status register is not updated while CSN is low, it is possible that the update of the VDSFSR is delayed until the next time it is commanded off, if the Power MOSFET is turned off during an SPI-frame.*

The contributions of the “VDSFSR” failure flags to the channel feedback status register and global status byte can be maskable through the VDSmaskx bit in the OUTCFGRx registers.



## 6 Programmable blanking window (PBW)

Dedicated registers (CHLOFFTCR1 and CHLOFFTCR0) provide a variable and programmable blanking window in case of power limitation or overtemperature event for each channel. During this period, the corresponding channel is in auto-restart mode and the channel is allowed to stay in power-limitation and/or overtemperature state before latching off, once blanking time has expired, if the cause of the power limitation or overtemperature event is still present. In this case the channel latches off and the related flag in the latch-off error register (CHLOFFSR) is set. Latch-off flag is also reported in the Global Status Byte (see [Section 4.3.1: Global status byte description](#)).

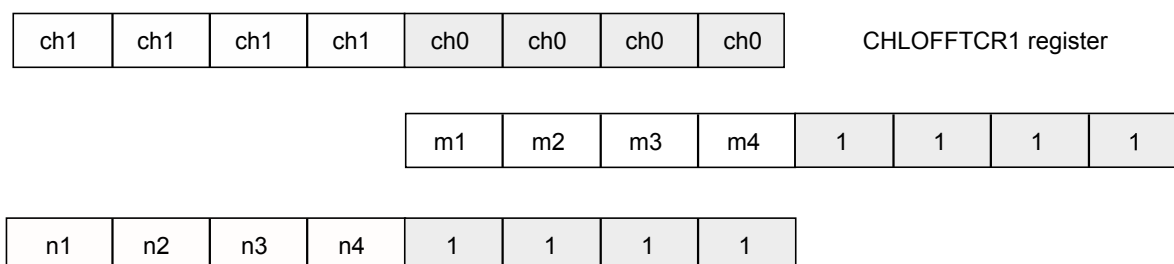
If during the blanking time the cause of power limitation and/or overtemperature event disappears, the timer stops then the rest of the blanking time will be available for another power limitation and/or overtemperature event. Therefore it is up to MCU to reset the timer by refreshing the programmed value in the dedicated register (CHLOFFTCR1 or CHLOFFTCR0).

MCU can keep the device in auto-restart forever artificially, by refreshing the programmed blanking time.

### 6.1 Timer

The 4-bit value per channel written in the registers CHLOFFTCR1 or CHLOFFTCR0 is translated internally into an 8-bit value. The four MSB of this 8-bit value correspond to the content of CHLOFFTCRx register, while the four LSB are filled with 0xF. The 8-bit value refers to an analogue timer value.

Figure 27. Internal timer process



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The granularity of the 8-bit counter is  $t_{STEP}$ . At each power limitation or overtemperature event, the 8-bit counter is decreased by the number of steps equal to the duration of power limitation or overtemperature event. If power limitation or overtemperature phase lasts for less than  $t_{STEP}$  the counter is decreased by one step.

After each downcount of the 8-bit register, the 4 MSB bits will be transferred to the 4 bits of the corresponding CHLOFFTCRx register in order to refresh this register to the new value of the timer. The microcontroller can read only the 4 MSB bits content of the register. In consequence, the microcontroller can detect a change of every 16 steps of downcounting.

The Timer down-counts, if the flag is set as the consequence of the event of power limitation or overtemperature. At the end of the timer's step, the flag is checked. It will be reset if the event is not present.

The Timer stops down-counting, each time the event has disappeared, or if the channel is turned into OFF state. This option doesn't include the one step down-counting if the flag is set for the first time.

If the event is not present, the Timer will stop down-counting and will reset the flag.

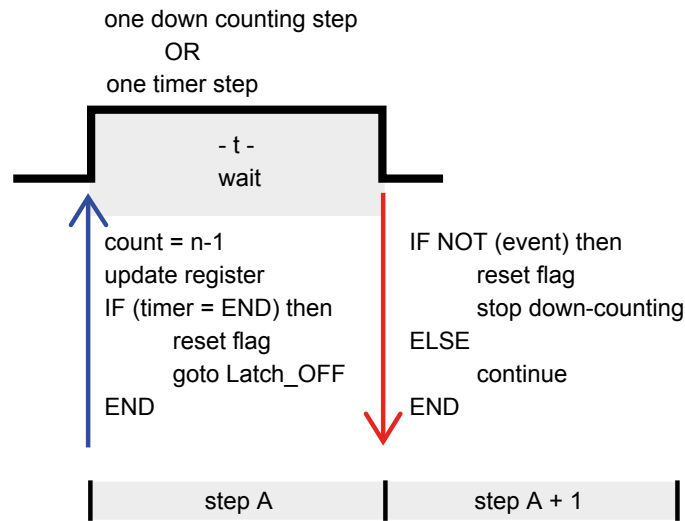
In case the Timer reaches the ZERO, the system goes to the latching off state and the related flag in the latch-off error register is set.

Downcounting is stopped and the content of the 8-bit counter is frozen, when the channel is commanded off through the Direct Input or the SOCR register. The Timer can stay with an already down-counted value for a long time. It is up to the MCU to reset it.

MCU can keep the device in auto-restart mode forever artificially, by refreshing the timer register value so not to reach Zero.

Here following is the figure related to the one timer step. The actions are performed after the rising and falling edges.

Figure 28. One timer step actions



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## 6.2

### Blanking window values

The range of the configurable blanking window is shown in Table 43. Time values written by MCU and their real value in timer register.

Blanking window reserved values:

- 0x0: It configures the channel in Latch-OFF mode without blanking time. Consequently, the channel will latch-off upon the first occurrence of power limitation or overtemperature event
- 0x1 to 0xF: This value represents the time duration it will be written by MCU in the register (latch-off timer register) "CHLOFFTCR<sub>x</sub>" (address 0x10h and 0x11h). During this time, the device is allowed to stay in a power-limitation and/or overtemperature state before latching off if the "event" is still active or present. The minimum value of the timer, known as zero, is 0x0F. When the timer reaches this value the latch-off action is triggered

The following table shows the time values written by the MCU and their real value in the timer Register.

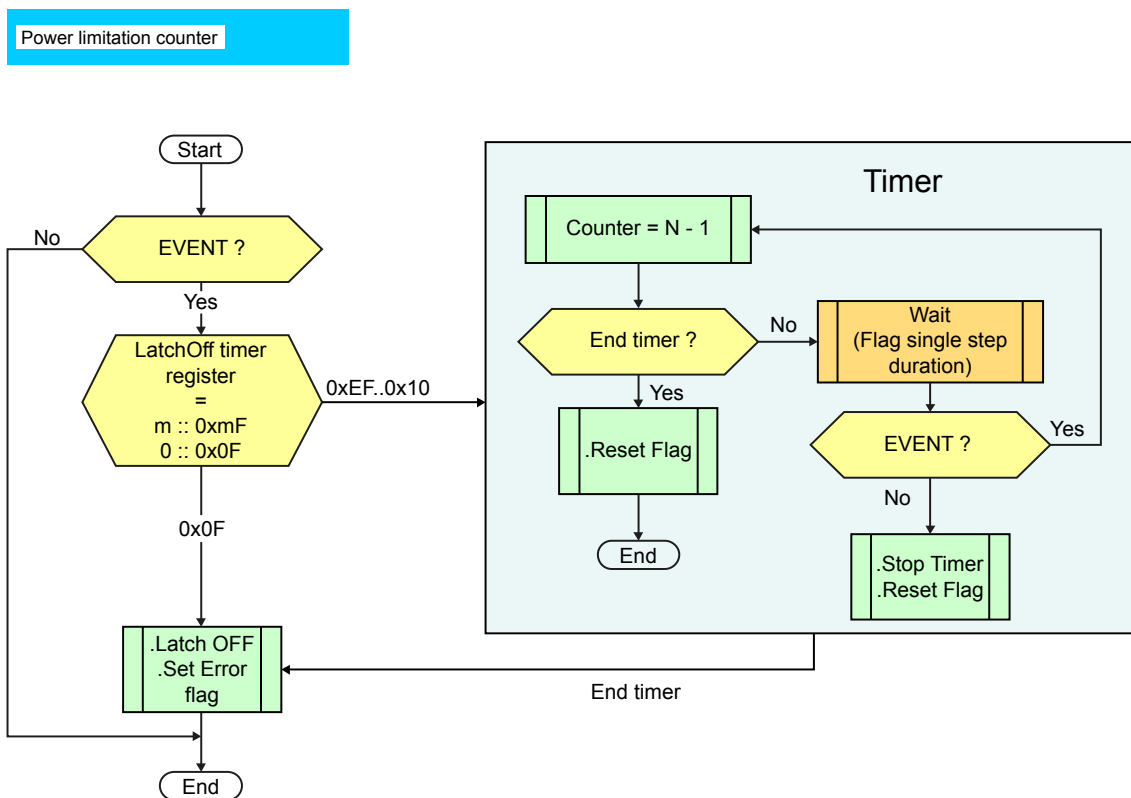
Table 43. Time values written by MCU and their real value in timer register

CHLOFFTCR <sub>x3</sub>	CHLOFFTCR <sub>x2</sub>	CHLOFFTCR <sub>x1</sub>	CHLOFFTCR <sub>x0</sub>	0xm	0xmF	Typical value of blanking time
0	0	0	0	0x0	0x0F	Latch-OFF (ZERO)
0	0	0	1	0x1	0x1F	16 ms
0	0	1	0	0x2	0x2F	32 ms
0	0	1	1	0x3	0x3F	48 ms
....				0x4	0x4F	64 ms
....				...	...	...
1	1	1	0	0xE	0xEF	224 ms
1	1	1	1	0xF	0xFF	240 ms

### 6.3 Power limitation counter

The flowchart below displays the flow of the events and states. It does not include the timer update by MCU.

Figure 29. Power limitation counter flowchart



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### 6.4 Limp-home mode

In limp-home mode, the device is in unlimited autorestart operation. The blanking time window has no effect on the duration of the autorestart. The timers in the limp-home mode are frozen and are inactive. This ensures full independence of the limp-home mode operation.

### 6.5 Registers

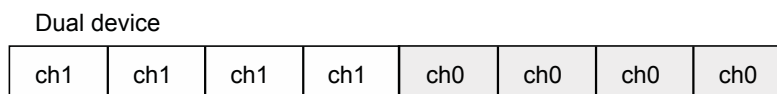
For more details refer to the SPI register and Diagnostics.

- Address 0x10h - Channel Latch OFF Timer Control Register (CHLOFFTCR0)
- Address 0x11h - Channel Latch OFF Timer Control Register (CHLOFFTCR1)

Two 16-bit registers (Latch-OFF timer: R/W) are used for the channel behavior configuration and the timer value setting.

For each channel 4 bits are used. The value is written by MCU from 0x0 to 0xF.

Figure 30. Example of behavior channel configuration



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**Latch-Off timer register access**

- **Write command** – stores new value, read-back (during write command) old value equal to the timer down-counting.
  - Any write command will clear the flag in the latch-OFF-flag register and reset the timer.
  - This function will be used by MCU to clear the flag in the latch-OFF-flag register, which is read only register.
- **Read command** – reads currently down-counted timer value. If the channel was latched because of timer expiration, the channel is kept latched after read command.
- **Channels latch-off status bit – CHLOFFSRx in OUTSRx** (address 0x20h to 0x23h depending on the channel) Each channel has one CHLOFFSR flag. In case of latch-OFF of a channel, this flag will be set and readable by MCU. This bit must be cleared to allow the channel to resume operation through a read/clear operation.

## 7 Electrical specifications

### 7.1 Absolute maximum ratings

Stressing the device above the rating listed in the [Table 44. Absolute maximum ratings](#) may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Table 44. Absolute maximum ratings**

Symbol	Parameter		Value	Unit
V <sub>CC SC</sub>	Maximum supply voltage for full short-circuit protection		18	V
V <sub>CCJS</sub>	Maximum jump start voltage for single pulse short circuit protection		28	V
V <sub>CC</sub>	DC supply voltage		36	V
-V <sub>CC</sub>	Reverse DC supply voltage (without external components)		0.3	V
I <sub>OUT0,1,2,3</sub>	Maximum DC output current		Internally limited	A
-I <sub>OUT0,1</sub>	Reverse DC output current		17.6	A
-I <sub>OUT2,3</sub>			8.8	A
I <sub>PWM_CLK</sub>	DC PWM_CLK input current		+3/-1	mA
V <sub>SDO</sub>	DC SPI pin voltage		V <sub>DD</sub> + 0.3	V
-V <sub>SDO</sub>	Reverse DC SPI pin voltage		0.3	V
I <sub>SDO</sub>	SDO pin current (as per LV124 / ISO 16750-2) t<=2min		54	mA
I <sub>SDI,CSN,SCK</sub>	DC SPI pin current		+10/-1	mA
I <sub>DD</sub>	DC Digital Control supply current		+10/-1	mA
	Digital Control supply current (as per LV124 / ISO 16750-2) t<=2min		80	mA
V <sub>DD</sub>	DC Digital Control supply		6	V
-V <sub>DD</sub>	Reverse DC Digital Control supply		0.3	V
I <sub>DIN0,1</sub>	DC direct input current		+10/-1	mA
E <sub>MAX</sub>	Maximum switching energy (single pulse); T <sub>Jstart</sub> = 150 °C, CH <sub>0, 1</sub> , Bulb mode		50	mJ
	Maximum switching energy (single pulse); T <sub>Jstart</sub> = 150 °C, CH <sub>0, 1</sub> , LED mode		14	
	Maximum switching energy (single pulse); T <sub>Jstart</sub> = 150 °C, CH <sub>2, 3</sub> , Bulb mode		14	
	Maximum switching energy (single pulse); T <sub>Jstart</sub> = 150 °C, CH <sub>2, 3</sub> , LED mode		3	
ESD	Electrostatic discharge (ANSI-ESDA-JEDEC-JS-001-2014)	DI <sub>0,1</sub>	1500	V
		V <sub>DD</sub>	1500	
		PWM_CLK	1500	
		CSN, SDI, SCK,SDO	1500	
		OUT <sub>0,1,2,3</sub>	4000	
		V <sub>CC</sub>	4000	
T <sub>J</sub>	Operating junction temperature range		-40 to 150	°C
T <sub>STG</sub>	Storage temperature range		-55 to 150	°C

Symbol	Parameter	Value	Unit
$I_{LAT}$	Latch-up current	$\pm 20$	mA

## 7.2 Thermal data

**Table 45. Thermal data**

Symbol	Parameter	Typ. value	Unit
$R_{thJB}$	Thermal resistance, junction-to-board (JEDEC JESD 51-8)	8.7	°C/W
$R_{thJA}$	Thermal resistance, junction-to-ambient	See Figure 42. $R_{thJA}$ vs PCB copper area in open box free air conditions.	

## 7.3 SPI electrical characteristics

$2.7\text{ V} < V_{DD} < 5.5\text{ V}$ ,  $-40\text{ °C} < T_J < 150\text{ °C}$ , unless otherwise specified

**Table 46. DC characteristics**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b><math>V_{DD}</math> pin</b>						
$V_{DD\_POR\_ON}$	Power-on reset threshold. Device leaves the Reset mode. Supply of digital part is reset.	$V_{DD}$ increasing; $V_{CC} > V_{USD}$	1.65	2.15	2.65	V
$V_{DD\_POR\_OFF}$	Power-on shutdown threshold. Device enters in reset mode. Supply of digital part in shutdown.	$V_{DD}$ decreasing; $V_{CC} > V_{USD}$	1.4	1.9	2.4	V
$V_{POR\_HYST}$	Power-on reset hysteresis			0.2		V
$I_{DD}$	Digital part supply current in normal mode	$V_{DD} = 5\text{ V}$ ; SPI active without frame communication		1	1.5	mA
$I_{DDstd}$	Digital part supply current in standby state	$V_{DD} = 5\text{ V}$ ; $T_J = 125\text{ °C}$ ; $I_{Nx} = 0\text{ V}$		5	35	μA
<b>SDI, SCK, PWM_CLK pins</b>						
$I_{IL}$	Low level Input current	$V_{SDI,SCK} = 0.3 V_{DD}$	1		10	μA
$I_{IH}$	High level Input current	$V_{SDI,SCK} = 0.7 V_{DD}$	1		10	μA
$V_{IL}$	Input low voltage				$0.3 V_{DD}$	V
$V_{IH}$	Input high voltage		$0.7 V_{DD}$			V
$V_{I\_HYST}$	Input hysteresis voltage			0.5		V
$V_{PWM\_CLK}$	PWM_CLK clamping voltage	$I_{IN} = 3\text{ mA}$	9		15	V
		$I_{IN} = -1\text{ mA}$		-0.7		V
$V_{SDI\_CL}$	SDI clamping voltage	$I_{IN} = 1\text{ mA}$	6		8.2	V
		$I_{IN} = -1\text{ mA}$		-0.7		V
$V_{SCK\_CL}$	SCK clamping voltage	$I_{IN} = 1\text{ mA}$	6		8.2	V
		$I_{IN} = -1\text{ mA}$		-0.7		V

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b>SDO pin</b>						
$V_{OL}$	Output low voltage	$I_{SDO} = -5 \text{ mA}$ ; CSN low; fault condition			$0.2 V_{DD}$	V
$V_{OH}$	Output high voltage	$I_{SDO} = 5 \text{ mA}$ ; CSN low; no fault condition	$0.8 V_{DD}$			V
$I_{LO}$	Output leakage current	$V_{SDO} = 0 \text{ V}$ or $V_{DD}$ , CSN high	-5		5	$\mu\text{A}$
<b>CSN pin</b>						
$I_{IL\_CSN}$	Low level Input current	$V_{CSN} = 0.3 V_{DD}$	-10		-1	$\mu\text{A}$
$I_{IH\_CSN}$	High level Input current	$V_{CSN} = 0.7 V_{DD}$	-10		-1	$\mu\text{A}$
$V_{IL\_CSN}$	Output low voltage				$0.3 V_{DD}$	V
$V_{IH\_CSN}$	Output high voltage		$0.7 V_{DD}$			V
$V_{HYST\_CSN}$	Input hysteresis voltage			0.5		V
$V_{CL\_CSN}$	CSN clamping voltage	$I_{IN} = 1 \text{ mA}$	6		8.2	V
		$I_{IN} = -1 \text{ mA}$		-0.7		V

**Table 47. AC characteristics (SDI, SCK, CSN, SDO, PWM\_CLK pins)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$C_{OUT}$	Output capacitance (SDO)	$V_{OUT} = 0 \text{ V}$ to $5 \text{ V}$	-	-	10	pF
$C_{IN}$	Input capacitance (SDI)	$V_{IN} = 0 \text{ V}$ to $5 \text{ V}$	-	-	10	pF
	Input capacitance (other pins)	$V_{IN} = 0 \text{ V}$ to $5 \text{ V}$	-	-	20	pF

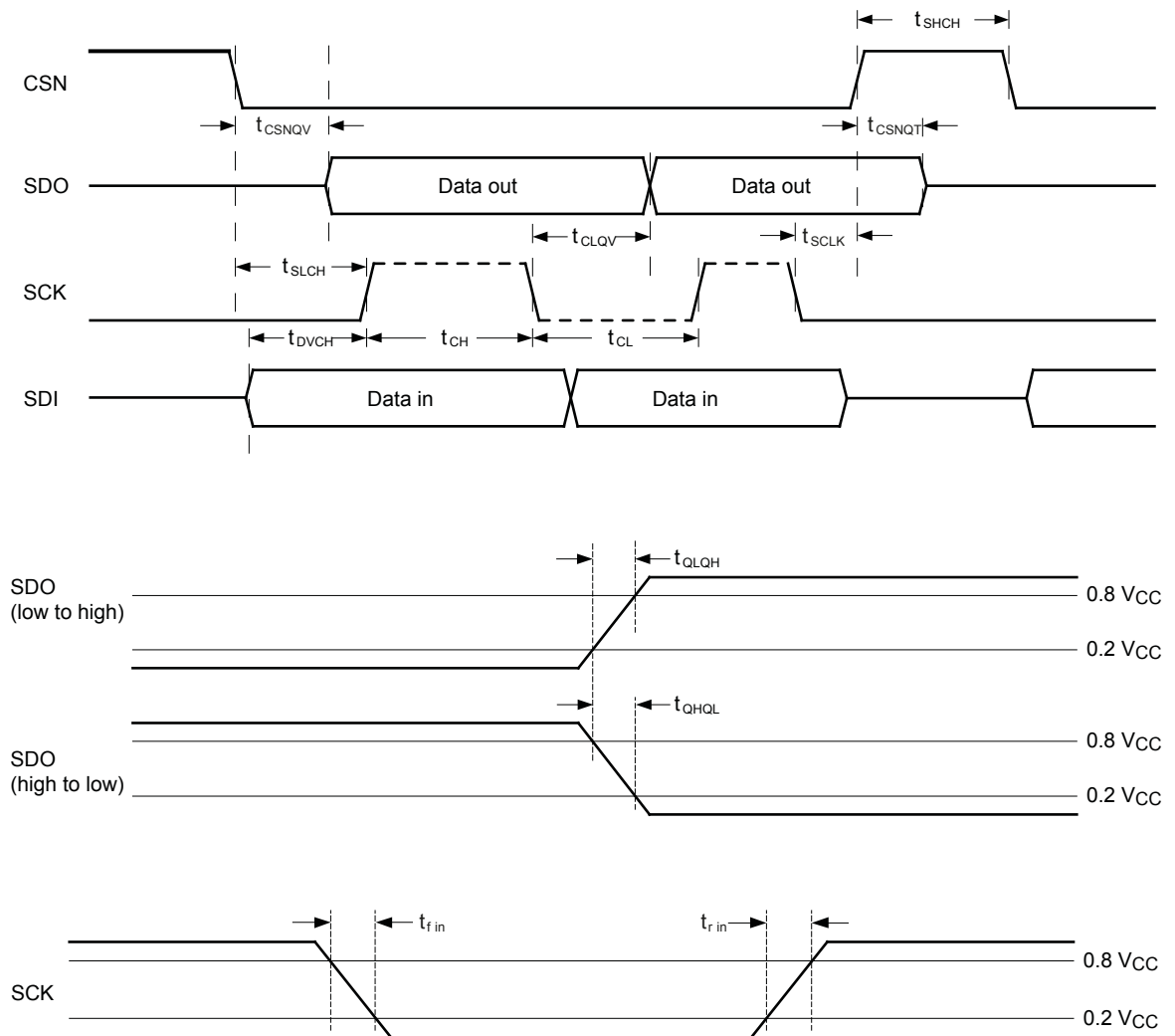
**Table 48. Dynamic characteristics**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$f_C$	Clock frequency	Duty cycle = 50%	0		8	MHz
$t_{WHCH}$	CSN timeout: time to release SDO bus		30		70	ms
$t_{WDTB}$	Watchdog toggle bit timeout		30		70	ms
$t_{SLCH}$	CSN low setup time		60			ns
$t_{SHCH}$	CSN high setup time		600			ns
$t_{DVCH}$	Data in setup time		10			ns
$t_{CHDX}$	Data in hold time		15			ns
$t_{CH}$	Clock high time		60			ns
$t_{CL}$	Clock low time		60			ns
$t_{CLQV}$	Clock low to output valid	$C_{OUT} = 1 \text{ nF}$		75		ns
$t_{QLQH}$	Output rise time	$C_{OUT} = 1 \text{ nF}$		55		ns
$t_{QHQL}$	Output fall time	$C_{OUT} = 1 \text{ nF}$		55		ns
$t_{WU}$	Rising edge of VDD to first allowed communication		3		23	$\mu\text{s}$
$t_{stdby\_out}$	Minimum time during which CSN must be toggled low to go out of STDBY mode		20	65	150	$\mu\text{s}$
$t_{SCLK}^{(1)}$	SCK setup time before CSN rising		20			ns
$t_{CSNQV}^{(1)}$	CSN low to output valid				200	ns

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{CSNQT}^{(1)}$	CSN high to output tristate				200	ns

1. Parameter specified by design and evaluated by characterization, not tested in production.

**Figure 31. SPI dynamic characteristics**



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## 7.4 Electrical characteristics

7 V < V<sub>CC</sub> < 28 V, -40 °C < T<sub>J</sub> < 150 °C, unless otherwise specified.

**Table 49. Power section**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V <sub>CC</sub>	Operating supply voltage		4	13	28	V
V <sub>USD</sub>	Undervoltage shutdown			2	2.7	V
V <sub>USDhyst</sub>	Undervoltage shutdown hysteresis			0.1		V
V <sub>clamp</sub>	V <sub>CC</sub> clamp voltage	I <sub>CC</sub> = 20 mA, I <sub>OUT0,1,2,3</sub> = 0 A, T <sub>J</sub> = -40 °C	35	-	-	V
		I <sub>CC</sub> = 20 mA, I <sub>OUT0,1,2,3</sub> = 0 A, 25 °C < T <sub>J</sub> < 150 °C	36	38	45	V
I <sub>S</sub>	Supply current	Sleep mode1, V <sub>CC</sub> = 13 V, T <sub>J</sub> = 25 °C, V <sub>DD</sub> = 0 V		0.1	0.8	μA
		Sleep mode2, V <sub>CC</sub> = 13 V, T <sub>J</sub> = 25 °C, V <sub>DD</sub> = 5 V		0.1	0.8	μA
		ON-state (all channels OFF), V <sub>CC</sub> = 13 V, V <sub>DD</sub> = 5 V, I <sub>OUT</sub> = 0 A		2.0	3.3	mA
ΔI <sub>Son</sub>	Additional supply current for each output in ON state driving nominal current	V <sub>DD</sub> = 5 V I <sub>OUT0,1</sub> = 5.4 A, I <sub>OUT2,3</sub> = 2.5 A			1.7	mA
I <sub>L(off)</sub>	OFF-state output current	V <sub>DD</sub> = 0 V, V <sub>CC</sub> = 13 V, T <sub>J</sub> = 25 °C	0.00	0.01	0.80	μA
		V <sub>DD</sub> = 0 V, V <sub>CC</sub> = 13 V, T <sub>J</sub> = 125 °C, Ch0,1 (per channel)	0.0		2.8	μA
		V <sub>DD</sub> = 0 V, V <sub>CC</sub> = 13 V, T <sub>J</sub> = 125 °C, Ch2,3(per channel)	0.0		0.7	μA
V <sub>F0,1</sub>	Output V <sub>CC</sub> diode voltage	V <sub>CC</sub> = 13 V, I <sub>OUT</sub> = 5.4 A, T <sub>J</sub> = 150 °C			0.7	V
V <sub>F2,3</sub>		V <sub>CC</sub> = 13 V, I <sub>OUT</sub> = 2.5 A, T <sub>J</sub> = 150 °C			0.7	V

**Table 50. Logic inputs (DI<sub>0,1</sub> pins)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V <sub>IL0,1</sub>	Input low level voltage				0.9	V
I <sub>IL0,1</sub>	Low level input current	V <sub>DIN</sub> = 0.9 V	1			μA
V <sub>IH0,1</sub>	Input high level voltage		2.1			V
I <sub>IH0,1</sub>	High level input current	V <sub>DIN</sub> = 2.1 V			10	μA
V <sub>I(hyst)0,1,2,3</sub>	Input hysteresis voltage		0.2			V
V <sub>ICL0,1</sub>	Input clamp voltage	I <sub>IN</sub> = 1 mA	6.0		8.2	V
		I <sub>IN</sub> = -1 mA		-0.7		V

**Table 51. Protection**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$\Delta T_{PLIM}^{(1)}$	Junction-case temperature difference triggering power limitation protection	$V_{CC} = 16\text{ V}$		80		$^{\circ}\text{C}$
		$V_{CC} = 19\text{ V}$		55		
$\Delta T_{PLIMR}$	Junction-case temperature difference resetting power limitation protection	$V_{CC} = 16\text{ V}$		58		$^{\circ}\text{C}$
		$V_{CC} = 19\text{ V}$		33		
$T_{TSD}$	Shutdown temperature		150	175	210	$^{\circ}\text{C}$
	Shutdown temperature ( $V_{CC}$ decreasing) <sup>(2)</sup>	$V_{CC} = 2.7\text{ V}$	140			
$T_{RS}$	Thermal reset of CHFBSR fault detection <sup>(2)</sup>	$V_{CC} = 13\text{ V}$ , latched off mode disabled	135			$^{\circ}\text{C}$
$T_{HYST}$	Thermal hysteresis ( $T_{TSD} - T_{RS}$ ) <sup>(2)</sup>	$V_{CC} = 13\text{ V}$ , latched off mode disabled		10		$^{\circ}\text{C}$
$T_{CSD}$	Case thermal detection pre-warning <sup>(2)</sup>	$V_{CC} = 13\text{ V}$ (see CTRL)	$T_{CSD\text{ nom}} - 10$	$T_{CSD\text{ nom}}$	$T_{CSD\text{ nom}} + 10$	$^{\circ}\text{C}$
$T_{CR}$	Case thermal detection reset <sup>(2)</sup>	$V_{CC} = 13\text{ V}$		$T_{CSD\text{ nom}} - 10$		$^{\circ}\text{C}$
$V_{DS\_OVL}$	$V_{DS}$ overload detection threshold		$V_{CC} - 2$	$V_{CC} - 1.5$	$V_{CC} - 1$	V
$t_{Blanking}$	Programmable blanking time		14.4		264	ms
$t_{D\_Restart}$	Latch-OFF delay time before automatic restart			55		ms
$t_{D\_GSB\_UPDATE}$	Status register update filtering time		1.4	2.0	2.6	ms
$t_{ON\_MIN}$	Minimum turn-on time per channel to avoid false $V_{DS}$ error flag	Bulb mode, ch0 and ch1			200	$\mu\text{s}$
		LED mode, ch0 and ch1			200	$\mu\text{s}$
		Bulb mode, ch2, ch3			200	$\mu\text{s}$
		LED mode, ch2, ch3			100	$\mu\text{s}$

1.  $Z_{thJC} \times P = \Delta T_{PLIM}$ ,  $Z_{thJC}$  is the thermal impedance,  $P$  is the Power.

2. Parameter specified by design and evaluated by characterization, not tested in production.

**Table 52. Open-load detection ( $7\text{ V} < V_{CC} < 18\text{ V}$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{OL}$	Open-load OFF-state voltage detection threshold	CHx off	$V_{CC} - 2$	$V_{CC} - 1.5$	$V_{CC} - 1$	V
$I_{PU}$	Pull-up current generator for open-load at OFF-state detection	Pull-up current generator active, $V_{OUT} = V_{CC} - 1.0\text{ V}$	-0.5	-1	-1.5	mA
$t_{DOLOFF}$	Delay time after turn off to allow open-load OFF-state detection		85	175	265	$\mu\text{s}$

## 7.5 PWM unit

2.7 V < V<sub>DD</sub> < 5.5 V; -40 °C < T<sub>J</sub> < 150 °C, unless otherwise specified.

**Table 53. PWM unit**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
PWM_Res	PWM resolution	PWM_Divider_1,2,3			0.1	%
		PWM_Divider_4			0.2	%
PWM_Clk	PWM clock range		300	400	500	kHz
PWM_Clk_flbk	PWM clock fallback		300	400	500	kHz
PWM_Clk_flbk_del	PWM clock fallback delay		20		40	µs

Note:

*PWMCLOCKLOW warning flag is set only when the period of the clock signal externally applied to PWM\_CLK pin is longer than PWM\_Clk\_flbk\_del.*

*Consequently, if the PWM\_CLK frequency is higher than 50 kHz (corresponding to a period of 20 µs, which is the minimum value of PWM\_Clk\_flbk\_del) this flag is never set.*

*When PWMCLOCKLOW warning flag is set, an internal fallback clock (at a typical frequency of 400 kHz) is used to substitute the external one.*

## 7.6 Bulb mode

**Table 54. Bulb - power section**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
R <sub>ON_ch0,1</sub> <sup>(1)</sup>	On-state resistance	I <sub>OUT</sub> = 5.4 A, T <sub>J</sub> = 25 °C	-	6.7		mΩ
		I <sub>OUT</sub> = 5.4 A, T <sub>J</sub> = 150 °C	-		14.7	mΩ
		I <sub>OUT</sub> = 5.4 A, V <sub>CC</sub> = 4 V, T <sub>J</sub> = 25 °C	-		12.3	mΩ
		I <sub>OUT</sub> = 2.5 A, V <sub>CC</sub> = 2.7 V, V <sub>CC</sub> decreasing	-		67	mΩ
R <sub>ON_ch2,3</sub> <sup>(1)</sup>	On-state resistance	I <sub>OUT</sub> = 2.5 A, T <sub>J</sub> = 25 °C	-	20		mΩ
		I <sub>OUT</sub> = 2.5 A, T <sub>J</sub> = 150 °C	-		44	mΩ
		I <sub>OUT</sub> = 2.5 A, V <sub>CC</sub> = 4 V, T <sub>J</sub> = 25 °C	-		37	mΩ
		I <sub>OUT</sub> = 1 A, V <sub>CC</sub> = 2.7 V, V <sub>CC</sub> decreasing	-		200	mΩ

1. For each channel.

**Table 55. Bulb - switching (V<sub>CC</sub> = 13 V, normal switch mode)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
t <sub>don</sub> <sup>(1)</sup>	Turn-on delay time Ch0,1 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, Bulb mode, from Dlx rising to 20% V <sub>OUT</sub> , R <sub>L</sub> = 2.4 Ω, SLOPECRx = 00	9	60	94	µs
	Turn-on delay time Ch2,3 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, Bulb mode, from Dlx rising to 20% V <sub>OUT</sub> , R <sub>L</sub> = 5.2 Ω, SLOPECRx = 00	10	35	54	µs
t <sub>doff</sub> <sup>(1)</sup>	Turn-off delay time Ch0,1 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, Bulb mode, from Dlx falling to 80% V <sub>OUT</sub> , R <sub>L</sub> = 2.4 Ω, SLOPECRx = 00	17	55	72	µs
	Turn-off delay time Ch2,3 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, Bulb mode, from Dlx falling to 80% V <sub>OUT</sub> , R <sub>L</sub> = 5.2 Ω, SLOPECRx = 00	11	40	56	µs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{\text{skew}}^{(1)}$	Turn-off turn-on time Ch0,1 at $T_J = 25\text{ }^{\circ}\text{C}$ to $150\text{ }^{\circ}\text{C}$	Differential pulse skew ( $t_{\text{pHL}} - t_{\text{pLH}}$ ), $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 00	-59	-5	41	$\mu\text{s}$
	Turn-off turn-on time Ch2,3 at $T_J = 25\text{ }^{\circ}\text{C}$ to $150\text{ }^{\circ}\text{C}$	Differential pulse skew ( $t_{\text{pHL}} - t_{\text{pLH}}$ ), $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 00	-48	5	51	$\mu\text{s}$
$(dV_{\text{OUT}}/dt)_{\text{on}}^{(1)}$	Turn-on voltage slope Ch0,1 at $T_J = 25\text{ }^{\circ}\text{C}$ to $150\text{ }^{\circ}\text{C}$	$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 00	0.18	0.4	0.74	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 01	0.19	0.42	0.76	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 10	0.20	0.44	0.80	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 11	0.21	0.46	0.82	$\text{V}/\mu\text{s}$
	Turn-on voltage slope Ch2,3 at $T_J = 25\text{ }^{\circ}\text{C}$ to $150\text{ }^{\circ}\text{C}$	$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 00	0.29	0.55	0.90	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 01	0.35	0.62	1.15	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 10	0.37	0.74	1.4	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 2.6\text{ V}$ to $7.8\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 11	0.59	1	1.62	$\text{V}/\mu\text{s}$
$(dV_{\text{OUT}}/dt)_{\text{off}}^{(1)}$	Turn-off voltage slope Ch0,1 at $T_J = 25\text{ }^{\circ}\text{C}$ to $150\text{ }^{\circ}\text{C}$	$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 00	0.10	0.45	0.80	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 01	0.09	0.52	0.87	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 10	0.09	0.6	0.93	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 2.4\text{ }\Omega$ , SLOPECRx = 11	0.08	0.72	1	$\text{V}/\mu\text{s}$
	Turn-off voltage slope Ch2,3 at $T_J = 25\text{ }^{\circ}\text{C}$ to $150\text{ }^{\circ}\text{C}$	$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 00	0.23	0.52	0.87	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 01	0.31	0.6	1.27	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 10	0.4	0.75	1.57	$\text{V}/\mu\text{s}$
		$V_{\text{OUT}} = 10.4\text{ V}$ to $5.2\text{ V}$ , $R_L = 5.2\text{ }\Omega$ , SLOPECRx = 11	0.44	1.12	1.84	$\text{V}/\mu\text{s}$
$W_{\text{ON}}$	Switching losses energy at turn-on Ch0,1	$R_L = 2.4\text{ }\Omega$ , SLOPECRx = 00		1.0	3.6 <sup>(2)</sup>	mJ
	Switching losses energy at turn-on Ch2,3	$R_L = 5.2\text{ }\Omega$ , SLOPECRx = 00		0.17	0.40 <sup>(2)</sup>	mJ
$W_{\text{OFF}}$	Switching losses energy at turn-off Ch0,1	$R_L = 2.4\text{ }\Omega$ , SLOPECRx = 00		0.52	0.90 <sup>(2)</sup>	mJ
	Switching losses energy at turn-off Ch2,3	$R_L = 5.2\text{ }\Omega$ , SLOPECRx = 00		0.14	0.25 <sup>(2)</sup>	mJ

1. See Figure 32. *Switching characteristics.*

2. Parameter specified by design and evaluated by characterization, not tested in production.

**Table 56. Bulb - protection and diagnostic ( $7 < V_{CC} < 18 \text{ V}$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{limH\_ch0,1}^{(1)}$	DC short-circuit current	$V_{CC} = 16 \text{ V}, T_J = -40^\circ \text{C}$	-15%	81.6	15%	A
		$V_{CC} = 16 \text{ V}, T_J = 150^\circ \text{C}$	-15%	56.5	15%	A
$I_{limH2\_ch0,1}^{(2)}$		$V_{CC} = 19 \text{ V}, T_J = -40^\circ \text{C}$	-15%	64	15%	A
		$V_{CC} = 19 \text{ V}, T_J = 150^\circ \text{C}$	-15%	44	15%	A
$I_{limH\_ch0,1}$ at 22 V		$V_{CC} = 22 \text{ V}, T_J = 25^\circ \text{C}$		40% $I_{limH2\_ch0,1}$		A
$I_{limH\_ch2,3}^{(1)}$	DC short-circuit current	$V_{CC} = 16 \text{ V}, T_J = -40^\circ \text{C}$	-15%	38.3	15%	A
		$V_{CC} = 16 \text{ V}, T_J = 150^\circ \text{C}$	-15%	27.3	15%	A
$I_{limH2\_ch2,3}^{(2)}$		$V_{CC} = 19 \text{ V}, T_J = -40^\circ \text{C}$	-15%	29.7	15%	A
		$V_{CC} = 19 \text{ V}, T_J = 150^\circ \text{C}$	-15%	21.1	15%	A
$I_{limH2\_ch2,3}$ at 22 V		$V_{CC} = 22 \text{ V}, T_J = 25^\circ \text{C}$		40% $I_{limH2\_ch2,3}$		A
$V_{DEMAG}$	Turn-off output voltage clamp	$I_{OUT} = 2 \text{ A}, V_{IN0,1} = 0 \text{ V},$ $L = 6 \text{ mH}, 25^\circ \text{C} < T_J < 150^\circ \text{C}$ $I_{OUT} = 1 \text{ A}, V_{IN2,3} = 0 \text{ V},$ $L = 6 \text{ mH}, 25^\circ \text{C} < T_J < 150^\circ \text{C}$	$V_{CC} - 36$	$V_{CC} - 38$	$V_{CC} - 45$	V

1.  $I_{LIMH}$ , guaranteed between 7 V and 16 V,  $-40^\circ \text{C} < T_J < 150^\circ \text{C}$ .

2.  $I_{LIMH2}$ , guaranteed between 16 V and 19 V,  $-40^\circ \text{C} < T_J < 150^\circ \text{C}$ .

**Table 57. Bulb - Digital Current Sense ( $7 \text{ V} < V_{CC} < 18 \text{ V}$ , channel 0,1,  $T_J = -40^\circ \text{C}$  to  $150^\circ \text{C}$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$K_{OL}$	Digital current sense gain: $ADC_{OUT} / I_{OUT}$	$I_{OUT} = 350 \text{ mA}, T_J = -40^\circ \text{C} \text{ to } 150^\circ \text{C}$	-65%	36	65%	1/A
$K_{LED}$	Digital current sense gain: $ADC_{OUT} / I_{OUT}$	$I_{OUT} = 600 \text{ mA}, T_J = -40^\circ \text{C} \text{ to } 150^\circ \text{C}$	-35%	34	35%	1/A
$K_0$	Digital current sense gain: $ADC_{OUT} / I_{OUT}$	$I_{OUT} = 1.1 \text{ A}, T_J = -40^\circ \text{C} \text{ to } 150^\circ \text{C}$	-15%	34	15%	1/A
$K_1$	Digital current sense gain: $ADC_{OUT} / I_{OUT}$	$I_{OUT} = 5.4 \text{ A}$	-8%	34	8%	1/A
$K_2$	Digital current sense gain: $ADC_{OUT} / I_{OUT}$	$I_{OUT} = 16.2 \text{ A}$	-7%	34	7%	1/A
$I_{OUT\_OFFSET}^{(1)}$	Output current offset	$ISENSE = 000H$	-90		90	mA
$I_{OUT\_SAT\_BULB}$	Output saturation current in Bulb mode	$ISENSE = 3FFH$	24.5			A
$t_{ON\_CS(min)}_{Bulb}$	Minimum ON time for digital current sense availability				280	$\mu\text{s}$

1. Parameter specified by design and evaluated by characterization, not tested in production.

**Table 58. Bulb - Digital Current Sense (7 V < V<sub>CC</sub> < 18 V, channel 2,3, T<sub>J</sub> = -40 °C to 150 °C)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K <sub>OL</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 90 mA	-65%	89	65%	1/A
K <sub>LED</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 150 mA	-35%	86	35%	1/A
K <sub>0</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 0.5 A	-15%	83	15%	1/A
K <sub>1</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 2.5 A	-8%	83	8%	1/A
K <sub>2</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 7.5 A	-7%	83	7%	1/A
I <sub>OUT_OFFSET</sub> <sup>(1)</sup>	Output current offset	ISENSE = 000H	-45		45	mA
I <sub>OUT_SAT_BULB</sub>	Output saturation current in Bulb mode	ISENSE = 3FFH	9.5			A
t <sub>ON_CS(min)_Bulb</sub>	Minimum ON time for digital current sense availability				280	μs

1. Parameter specified by design and evaluated by characterization, not tested in production.

## 7.7

### LED mode

7 V < V<sub>CC</sub> < 18 V; -40 °C < T<sub>J</sub> < 150 °C, unless otherwise specified.

**Table 59. LED - power section**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
R <sub>ON</sub> Ch0,1 <sup>(1)</sup>	ON-state resistance	I <sub>OUT</sub> = 1.35 A; T <sub>J</sub> = 25 °C	-	24		mΩ
		I <sub>OUT</sub> = 1.35 A; T <sub>J</sub> = 150 °C	-		52	mΩ
		I <sub>OUT</sub> = 1.35 A; V <sub>CC</sub> = 4 V; T <sub>J</sub> = 25 °C	-		44	mΩ
		I <sub>OUT</sub> = 0.67 A, V <sub>CC</sub> = 2.7 V, V <sub>CC</sub> decreasing	-		240	mΩ
R <sub>ON</sub> Ch2,3 <sup>(1)</sup>	ON-state resistance	I <sub>OUT</sub> = 0.625 A; T <sub>J</sub> = 25 °C	-	80		mΩ
		I <sub>OUT</sub> = 0.625 A; T <sub>J</sub> = 150 °C	-		176	mΩ
		I <sub>OUT</sub> = 0.625 A; V <sub>CC</sub> = 4 V; T <sub>J</sub> = 25 °C	-		148	mΩ
		I <sub>OUT</sub> = 0.31 A, V <sub>CC</sub> = 2.7 V, V <sub>CC</sub> decreasing	-		800	mΩ

1. For each channel.

**Table 60. LED - switching (V<sub>CC</sub> = 13 V; Normal switch mode)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
t <sub>don</sub> <sup>(1)</sup>	Turn-on delay time Ch0,1 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, LED mode, from Dlx rising to 20% V <sub>OUT</sub> ; R <sub>L</sub> = 10 Ω; SLOPECRx = 00	9.5	30	45	μs
	Turn-on delay time Ch2,3 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, LED mode, from Dlx rising to 20% V <sub>OUT</sub> ; R <sub>L</sub> = 21 Ω; SLOPECRx = 00	4	15	28	μs
t <sub>doff</sub> <sup>(1)</sup>	Turn-off delay time Ch0,1 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, LED mode, from Dlx falling to 80% V <sub>OUT</sub> ; R <sub>L</sub> = 10 Ω; SLOPECRx = 00	11.5	37	47	μs
	Turn-off delay time Ch2,3 at T <sub>J</sub> = 25 °C to 150 °C	Fail-safe mode, LED mode, from Dlx falling to 80% V <sub>OUT</sub> ; R <sub>L</sub> = 21 Ω; SLOPECRx = 00	6.5	26	34	μs
t <sub>skew</sub> <sup>(1)</sup>	Turn-off, turn-on time Ch0,1 at T <sub>J</sub> = 25 °C to 150 °C	Differential Pulse skew (tp <sub>HL</sub> -tp <sub>LH</sub> ); R <sub>L</sub> = 10 Ω; SLOPECRx = 00	-48	0	51	μs
	Turn-off, turn-on time Ch2,3 at T <sub>J</sub> = 25 °C to 150 °C	Differential Pulse skew (tp <sub>HL</sub> -tp <sub>LH</sub> ); R <sub>L</sub> = 21 Ω; SLOPECRx = 00	-47	18	52	μs
(dV <sub>OUT</sub> /dt) <sub>on</sub> <sup>(1)</sup>	Turn-on voltage slope Ch0,1 at T <sub>J</sub> = 25 °C to 150 °C	from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 10 Ω; SLOPECRx = 00	0.43	0.81	1.25	V/μs
		from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 10 Ω; SLOPECRx = 01	0.50	0.9	1.55	V/μs
		from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 10 Ω; SLOPECRx = 10	0.57	1.11	1.85	V/μs
		from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 10 Ω; SLOPECRx = 11	0.61	1.55	2.10	V/μs
	Turn-on voltage slope Ch2,3 at T <sub>J</sub> = 25 °C to 150 °C	from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 21 Ω; SLOPECRx = 00	0.61	1.05	1.47	V/μs
		from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 21 Ω; SLOPECRx = 01	0.72	1.15	1.9	V/μs
		from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 21 Ω; SLOPECRx = 10	0.87	1.4	2.3	V/μs
		from V <sub>OUT</sub> = 2.6 V to 7.8 V; R <sub>L</sub> = 21 Ω; SLOPECRx = 11	1.06	2.1	2.86	V/μs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$(dV_{OUT}/dt)_{off}^{(1)}$	Turn-off voltage slope Ch0,1 at $T_J = 25\text{ °C}$ to $150\text{ °C}$	from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 10\text{ }\Omega$ ; SLOPECRx = 00	0.25	0.72	1.25	V/ $\mu$ s
		from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 10\text{ }\Omega$ ; SLOPECRx = 01	0.29	0.85	1.55	V/ $\mu$ s
		from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 10\text{ }\Omega$ ; SLOPECRx = 10	0.34	1.07	1.85	V/ $\mu$ s
		from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 10\text{ }\Omega$ ; SLOPECRx = 11	0.39	1.45	2.08	V/ $\mu$ s
	Turn-off voltage slope Ch2,3 at $T_J = 25\text{ °C}$ to $150\text{ °C}$	from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 21\text{ }\Omega$ ; SLOPECRx = 00	0.25	0.8	1.37	V/ $\mu$ s
		from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 21\text{ }\Omega$ ; SLOPECRx = 01	0.33	1	1.87	V/ $\mu$ s
		from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 21\text{ }\Omega$ ; SLOPECRx = 10	0.42	1.15	2.3	V/ $\mu$ s
		from $V_{OUT} = 10.4\text{ V}$ to $5.2\text{ V}$ ; $R_L = 21\text{ }\Omega$ ; SLOPECRx = 11	0.51	1.85	2.85	V/ $\mu$ s
$W_{ON}$	Switching losses energy at turn-on Ch0,1	$R_L = 10\text{ }\Omega$ ; SLOPECRx = 00		0.07	0.21 <sup>(2)</sup>	mJ
	Switching losses energy at turn-on Ch2,3	$R_L = 21\text{ }\Omega$ ; SLOPECRx = 00		0.025	0.05 <sup>(2)</sup>	mJ
$W_{OFF}$	Switching losses energy at turn-off Ch0,1	$R_L = 10\text{ }\Omega$ ; SLOPECRx = 00		0.07	0.15 <sup>(2)</sup>	mJ
	Switching losses energy at turn-off Ch2,3	$R_L = 21\text{ }\Omega$ ; SLOPECRx = 00		0.025	0.05 <sup>(2)</sup>	mJ

1. See Figure 32. Switching characteristics.

2. Parameter specified by design and evaluated by characterization, not tested in production.

**Table 61. LED - protection and diagnosis**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{limH\_ch0,1}^{(1)}$	DC short-circuit current	$V_{CC} = 16\text{ V}$ , $T_J = -40\text{ }^{\circ}\text{C}$	-15%	20.2	15%	A
		$V_{CC} = 16\text{ V}$ , $T_J = 150\text{ }^{\circ}\text{C}$	-15%	13.6	15%	A
$I_{limH2\_ch0,1}^{(2)}$		$V_{CC} = 19\text{ V}$ , $T_J = -40\text{ }^{\circ}\text{C}$	-15%	15.6	15%	A
		$V_{CC} = 19\text{ V}$ , $T_J = 150\text{ }^{\circ}\text{C}$	-15%	10.5	15%	A
$I_{limH2\_ch0,1}$ at 22 V		$V_{CC} = 22\text{ V}$ , $T_J = 25\text{ }^{\circ}\text{C}$		40% $I_{limH2\_ch0,1}$		A
$I_{limH\_ch2,3}^{(1)}$	DC short-circuit current	$V_{CC} = 16\text{ V}$ , $T_J = -40\text{ }^{\circ}\text{C}$	-15%	9.3	15%	A
		$V_{CC} = 16\text{ V}$ , $T_J = 150\text{ }^{\circ}\text{C}$	-15%	6.8	15%	A
$I_{limH2\_ch2,3}^{(2)}$		$V_{CC} = 19\text{ V}$ , $T_J = -40\text{ }^{\circ}\text{C}$	-15%	7.4	15%	A
		$V_{CC} = 19\text{ V}$ , $T_J = 150\text{ }^{\circ}\text{C}$	-15%	5.3	15%	A
$I_{limH2\_ch2,3}$ at 22 V		$V_{CC} = 22\text{ V}$ , $T_J = 25\text{ }^{\circ}\text{C}$		40% $I_{limH2\_ch2,3}$		A

1.  $I_{LIMH}$ , guaranteed between 7 V and 16 V,  $-40\text{ °C} < T_J < 150\text{ °C}$

2.  $I_{LIMH2}$ , guaranteed between 16 V and 19 V,  $-40\text{ °C} < T_J < 150\text{ °C}$



**Table 62. LED - Digital Current Sense (7 V < V<sub>CC</sub> < 18 V, channel 0, 1; T<sub>J</sub> = -40 °C to 150 °C)**

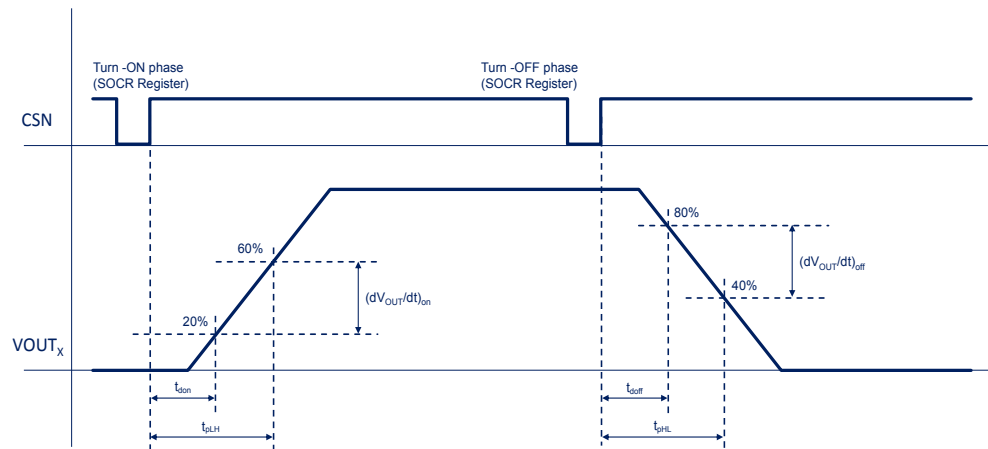
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K <sub>OL</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 90 mA	-65%	148	65%	1/A
K <sub>LED</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 140 mA	-35%	144	35%	1/A
K <sub>0</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 270 mA	-15%	143	15%	1/A
K <sub>1</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 1.35 A	-8%	143	8%	1/A
K <sub>2</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 4 A	-7%	143	7%	1/A
I <sub>OUT_OFFSET</sub> <sup>(1)</sup>	Output current offset	ISENSE = 000H	-25		25	mA
I <sub>OUT_SAT_LED</sub>	Output saturation current in LED mode	ISENSE = 3FFH	6			A
t <sub>ON_CS(min)_LED</sub>	Minimum ON time for digital current sense availability				150	µs

1. Parameter specified by design and evaluated by characterization, not tested in production.

**Table 63. LED - Digital Current Sense (7 V < V<sub>CC</sub> < 18 V, channel 2, 3; T<sub>J</sub> = -40 °C to 150 °C)**

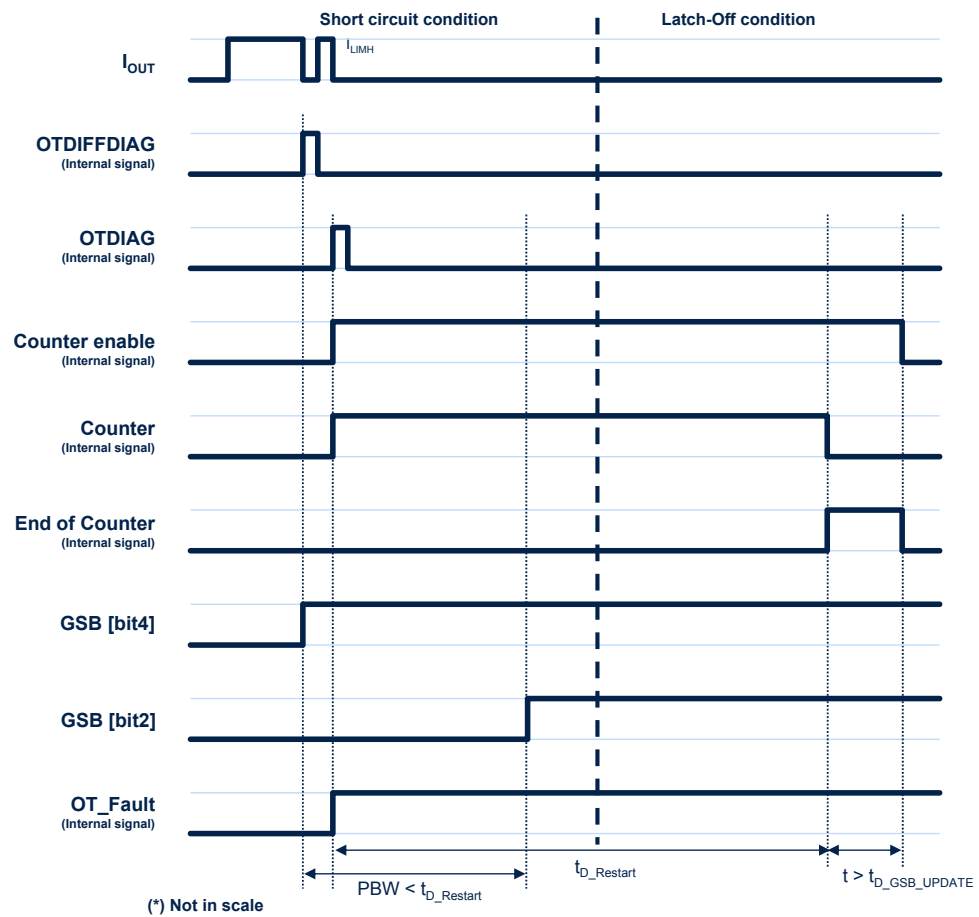
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K <sub>OL</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 25 mA	-65%	354	65%	1/A
K <sub>LED</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 40 mA	-35%	346	35%	1/A
K <sub>0</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 125 mA	-15%	335	15%	1/A
K <sub>1</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 625 mA	-8%	333	8%	1/A
K <sub>2</sub>	Digital current sense gain: ADC <sub>OUT</sub> / I <sub>OUT</sub>	I <sub>OUT</sub> = 1.875 A	-7%	333	7%	1/A
I <sub>OUT_OFFSET</sub> <sup>(1)</sup>	Output current offset	ISENSE = 000H	-15		15	mA
I <sub>OUT_SAT_LED</sub>	Output saturation current in LED mode	ISENSE = 3FFH	2.5			A
t <sub>ON_CS(min)_LED</sub>	Minimum ON time for digital current sense availability				150	µs

1. Parameter specified by design and evaluated by characterization, not tested in production.

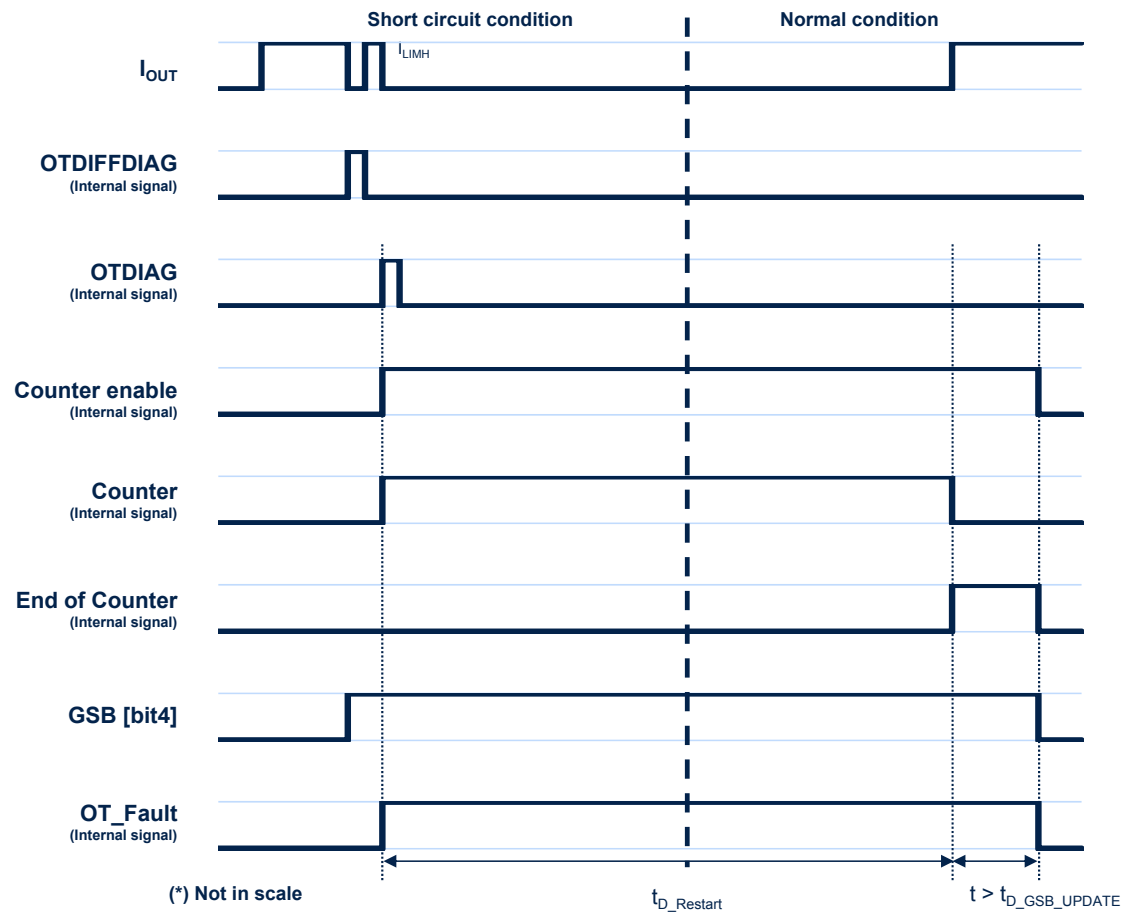
**Figure 32. Switching characteristics**


## 7.8 Waveforms

Figure 33. Normal mode - Short circuit –  $PBW < t_{D\_Restart}$



**Figure 34. Normal mode - Short circuit –  $PBW > t_{D\_Restart}$**



## 8 ISO Pulse

The immunity of the device against transient electrical emissions, conducted along the supply lines and injected into the VCC pin, is tested in accordance with ISO7637-2:2011(E) and ISO 16750-2:2010.

The related function performances status classification is shown in the [Table 64. ISO 7637-2 - electrical transient conduction along supply line](#).

Test pulses are applied directly to DUT (device under test) both in ON and OFF-state and in accordance to ISO 7637-2:2011(E), chapter 4. The DUT is intended as the present device only, with external components, as shown in Figure 1.

"Status II" is defined in ISO 7637-1 Function Performed Status Classification (FPSC) as follows: "The function does not perform as designed during the test but returns automatically to normal operation after the test".

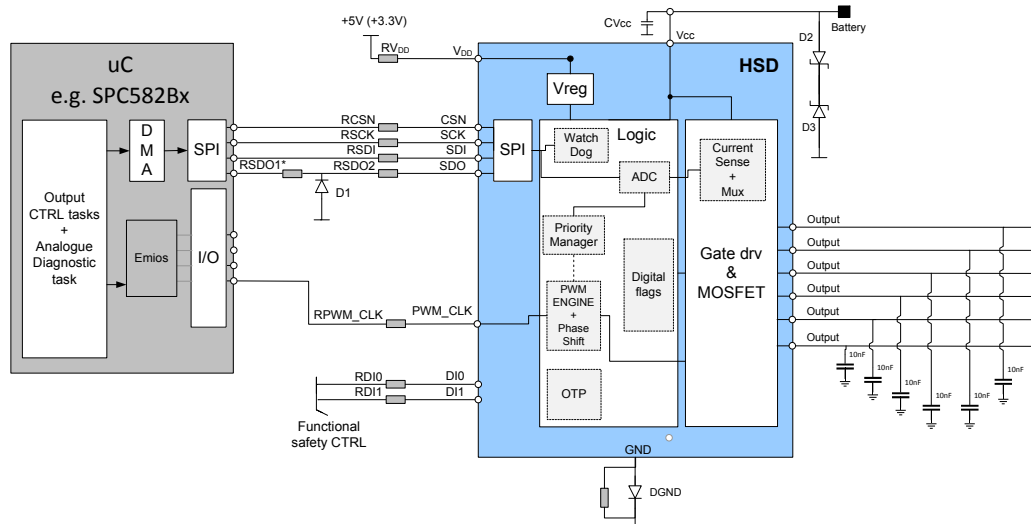
**Table 64. ISO 7637-2 - electrical transient conduction along supply line**

Test pulse 2011(E)	Test pulse severity level with status II functional performance status		Minimum number of pulses or test time	Burst cycle / pulse repetition time		Pulse duration and pulse generator internal impedance
	Level	Us <sup>(1)</sup>				
1 <sup>(2)</sup>	III	-112 V	500 pulses	0.5 s	5 s	2 ms, 10 Ω
2a <sup>(3)</sup>	III	+55 V	500 pulses	0.2 s	5 s	50 μs, 2 Ω
3a	IV	-220 V	1 h	90 ms	100 ms	0.1 μs, 50 Ω
3b	IV	+150 V	1 h	90 ms	100 ms	0.1 μs, 50 Ω
4 <sup>(4)</sup>	IV	-7 V	1 pulse			100 ms, 0.01 Ω
Load dump according to ISO 16750-2:2010						
Test B <sup>(3)</sup>		+87 V	5 pulses	1 min		400 ms, 2 Ω

1. *US is the peak amplitude as defined for each test pulse in ISO 7637-2:2011(E), chapter 5.6.*
2. *With -18 V external suppressor referred to ground (from -40 °C to 150 °C).*
3. *With 35 V external suppressor referred to ground (-40 °C < T<sub>J</sub> < 150 °C).*
4. *Test pulse in ISO 7637-2:2004(E).*

## 9 Application schematics

**Figure 35. M0-9SPI application Schematic**

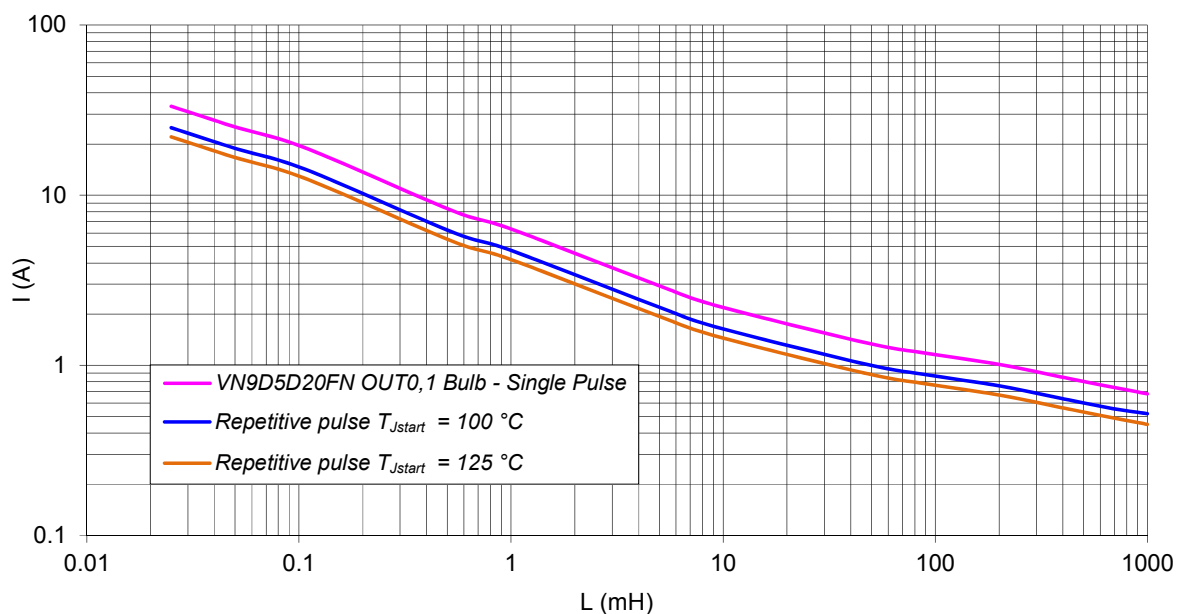


**Table 65. Component values**

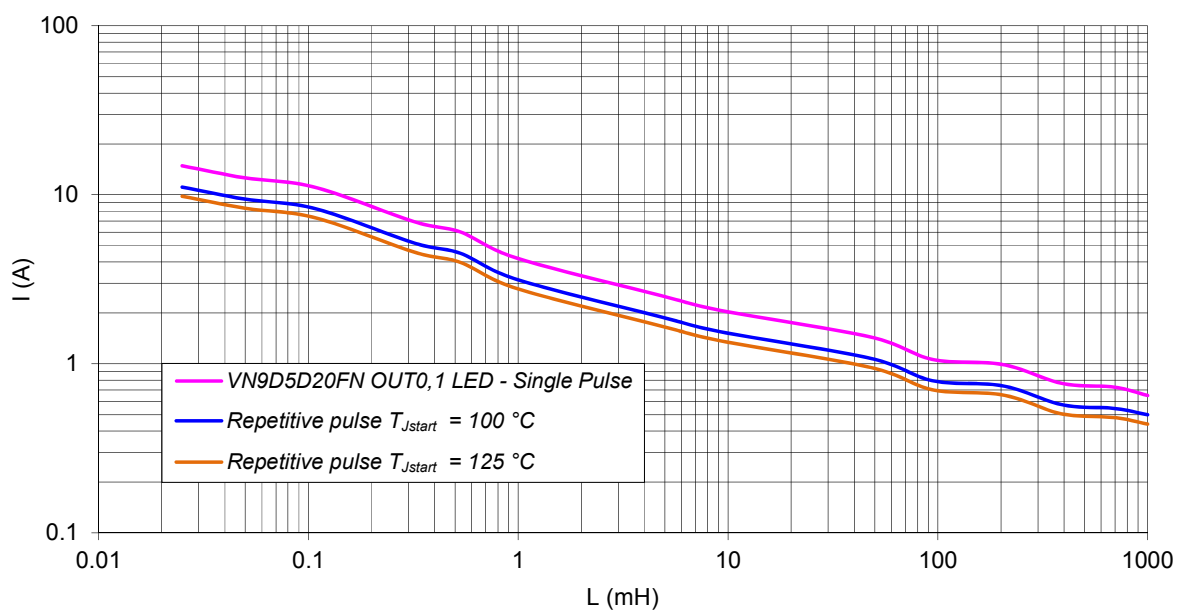
Reference	Value	comment
RV <sub>DD</sub>	330 Ohm for V <sub>DD</sub> =5V 150 Ohm for V <sub>DD</sub> =3.3V	Device logic protection
CV <sub>CC</sub>	100 nF	Battery voltage spikes filtering mounted close to IC
RCSN	2.7 kΩ	Microcontroller protection during overvoltage and reverse polarity
RSCK	2.7 kΩ	Microcontroller protection during overvoltage and reverse polarity
RSDI	2.7 kΩ	Microcontroller protection during overvoltage and reverse polarity
RSDO2	220 Ω	Microcontroller protection during overvoltage and reverse polarity
RSDO1	50 Ω	Optional
D1	BAT54	Microcontroller protection during overvoltage and reverse polarity
RPWM_CLK	2.2 kΩ	Microcontroller protection during: overvoltage, reverse polarity and loss of GND
RDIO	15 kΩ	Microcontroller protection during: overvoltage, reverse polarity and loss of GND
RDI1	15 kΩ	Microcontroller protection during: overvoltage, reverse polarity and loss of GND
D2	Suppressor 20 V	Negative transient protection.
D3	Suppressor 36 V	Overvoltage protection.
RGND	4.7 kΩ	
DGND	BAS21 for V <sub>DD</sub> = 5 V Schottky (i.e., BAT54-Y) for V <sub>DD</sub> = 3.3 V	Reverse polarity protection. Usage of schottky or standard diode dependent on V <sub>DD</sub>

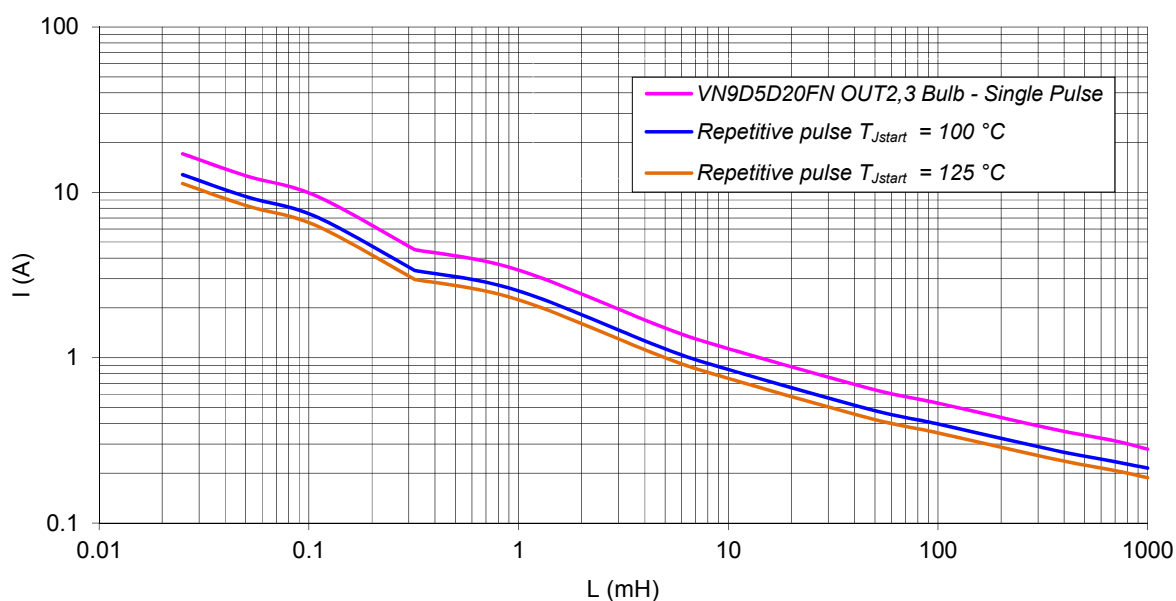
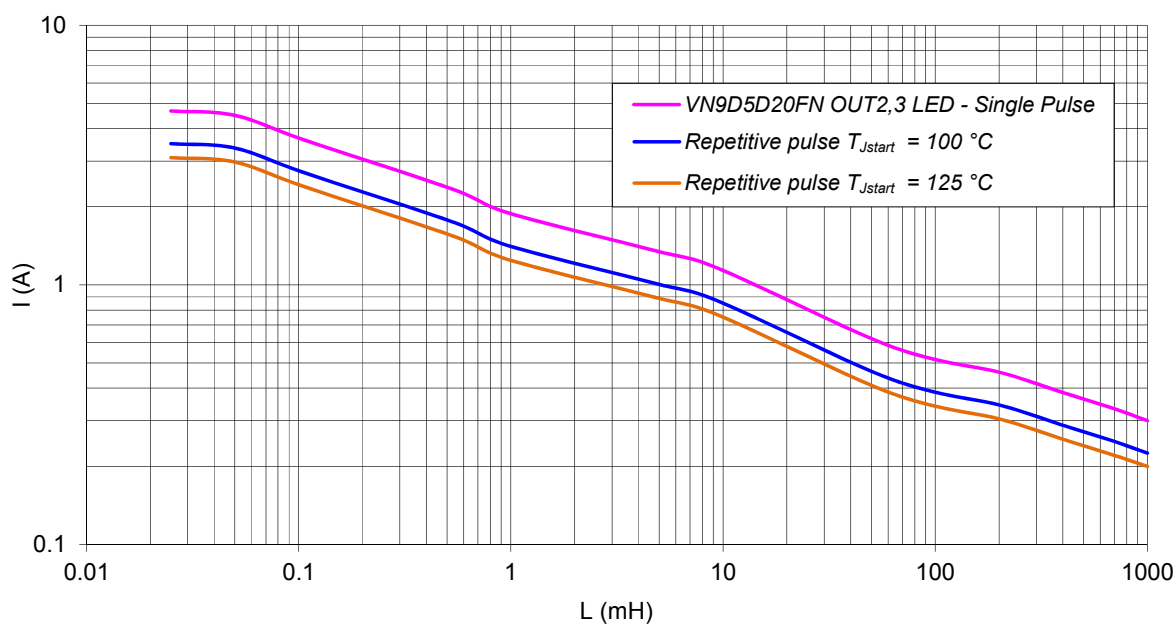
## 10 Maximum demagnetization energy ( $V_{CC} = 16\text{ V}$ )

**Figure 36.** Maximum turn off current versus inductance - Ch0,1 - Bulb Mode



**Figure 37.** Maximum turn off current versus inductance - Ch0,1 - LED Mode



**Figure 38. Maximum turn off current versus inductance - Ch2,3 - Bulb Mode**

**Figure 39. Maximum turn off current versus inductance - Ch2,3 - LED Mode**


**Note:** Values are generated with  $R_L = 0 \Omega$ .

In case of repetitive pulses,  $T_{Jstart}$  (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for repetitive curves.

## 11 Package and PCB thermal data

### 11.1 QFN 6x6 thermal data

Figure 40. QFN 6x6 PCB 2 layers

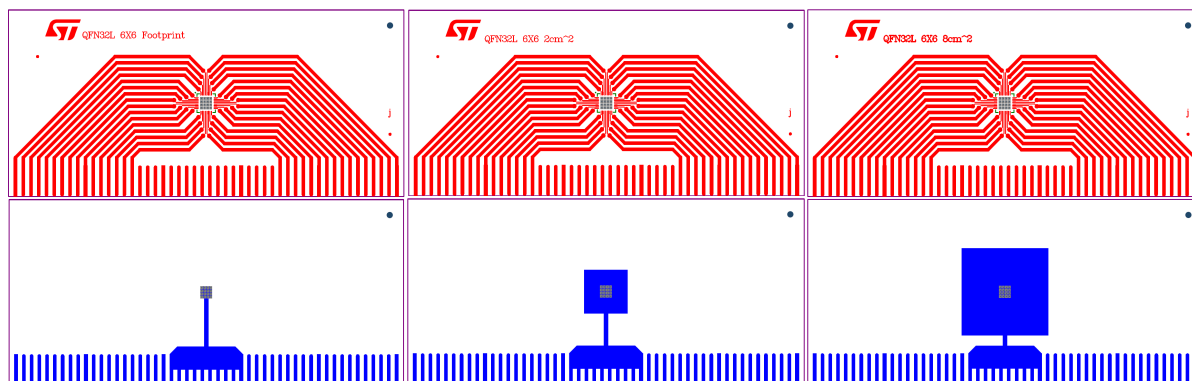
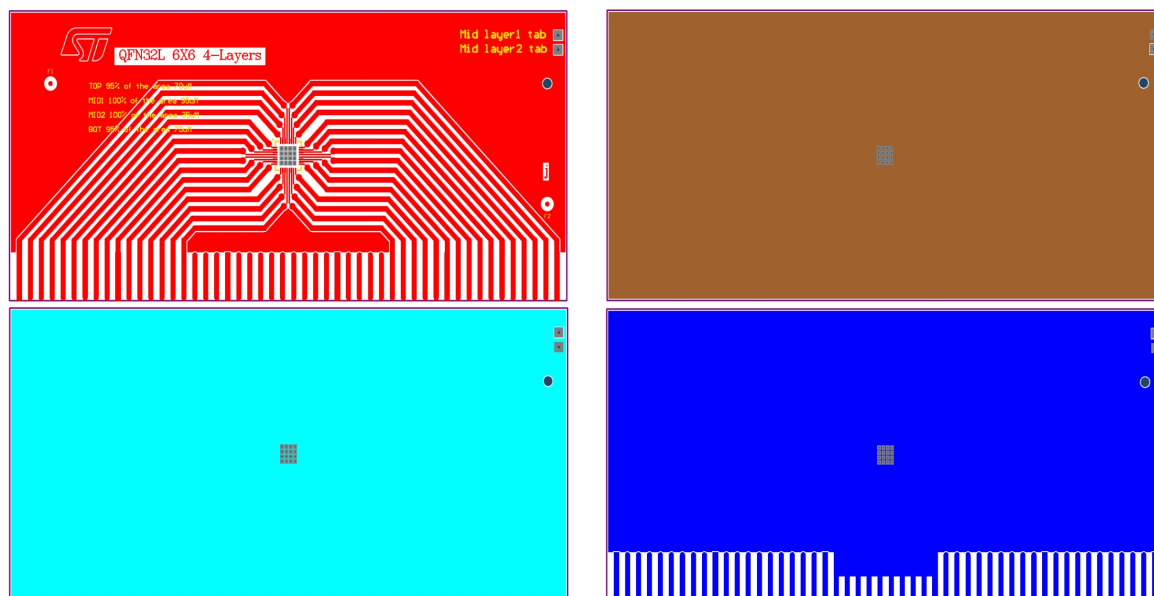


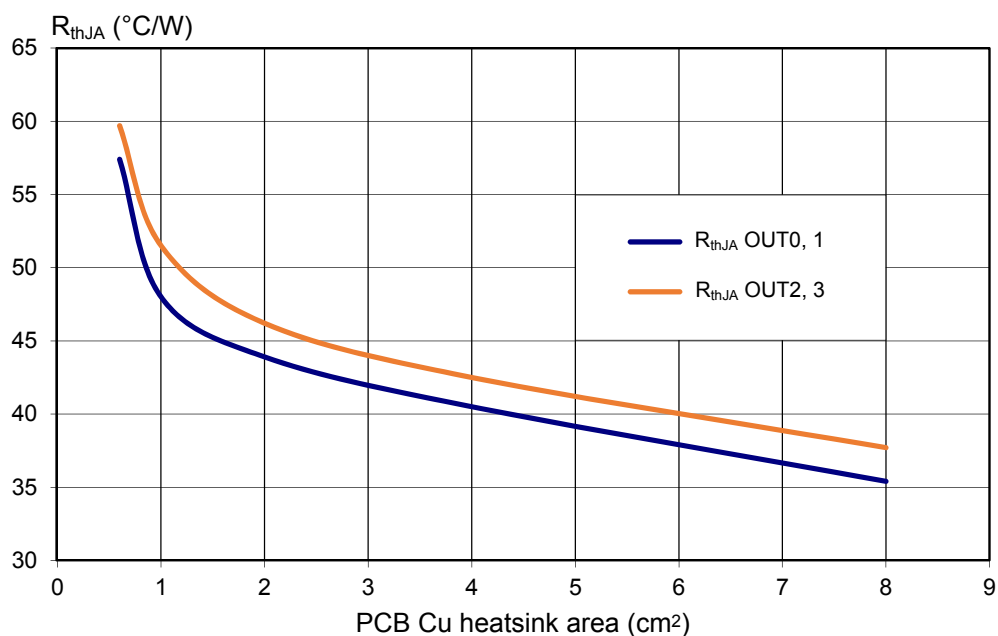
Figure 41. QFN 6x6 PCB 4 layers





**Table 66. PCB properties**

Dimension	Value
Board finish thickness	1.6 mm $\pm 10\%$
Board dimension	129 mm x 60 mm
Board material	FR4
Cu thickness (top and bottom layers)	0.070 mm
Cu thickness (inner layers)	0.035 mm
Thermal vias separation	1.2 mm
Thermal vias diameter	0.3 mm $\pm 0.08$ mm
Cu thickness on vias	0.025 mm
Footprint dimension (top layer)	6 mm x 6 mm
Heatsink copper area dimension (bottom layer)	Footprint, 2 cm <sup>2</sup> or 8 cm <sup>2</sup>

**Figure 42.  $R_{thJA}$  vs PCB copper area in open box free air conditions**

 $R_{thJC} = 8.7 \text{ }^{\circ}\text{C/W}$ 

 OUT0, 1  $R_{thJA}$  on 4Layers PCB: 20.5  $^{\circ}\text{C/W}$ 

 OUT2, 3  $R_{thJA}$  on 4Layers PCB: 24.2  $^{\circ}\text{C/W}$

Figure 43. QFN 6x6 thermal impedance junction ambient for channels 0 and 1

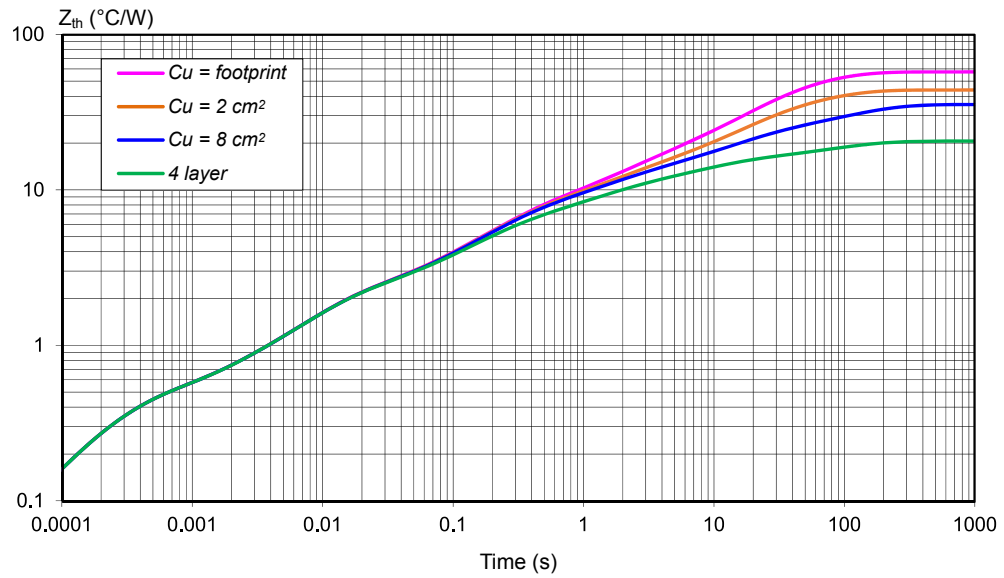
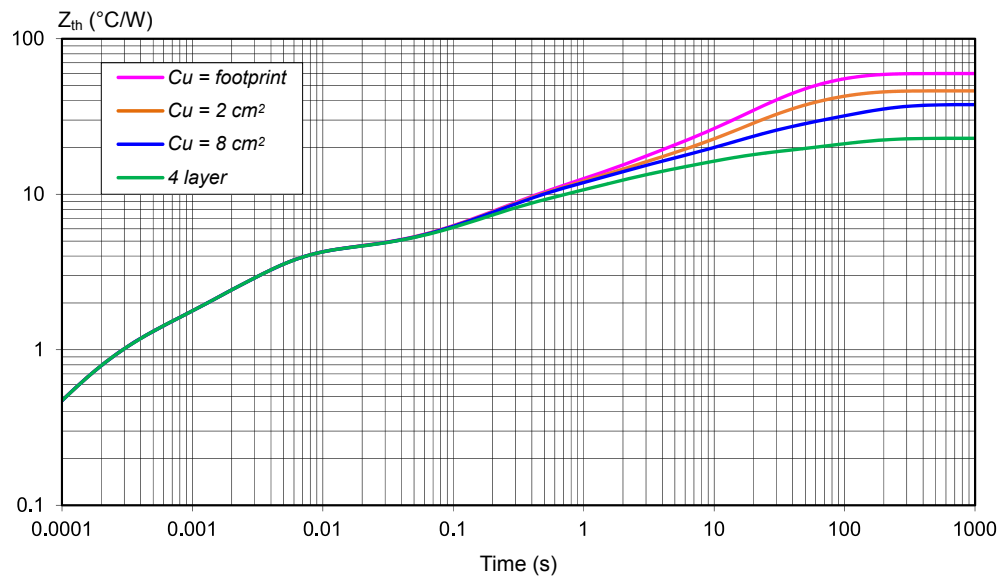
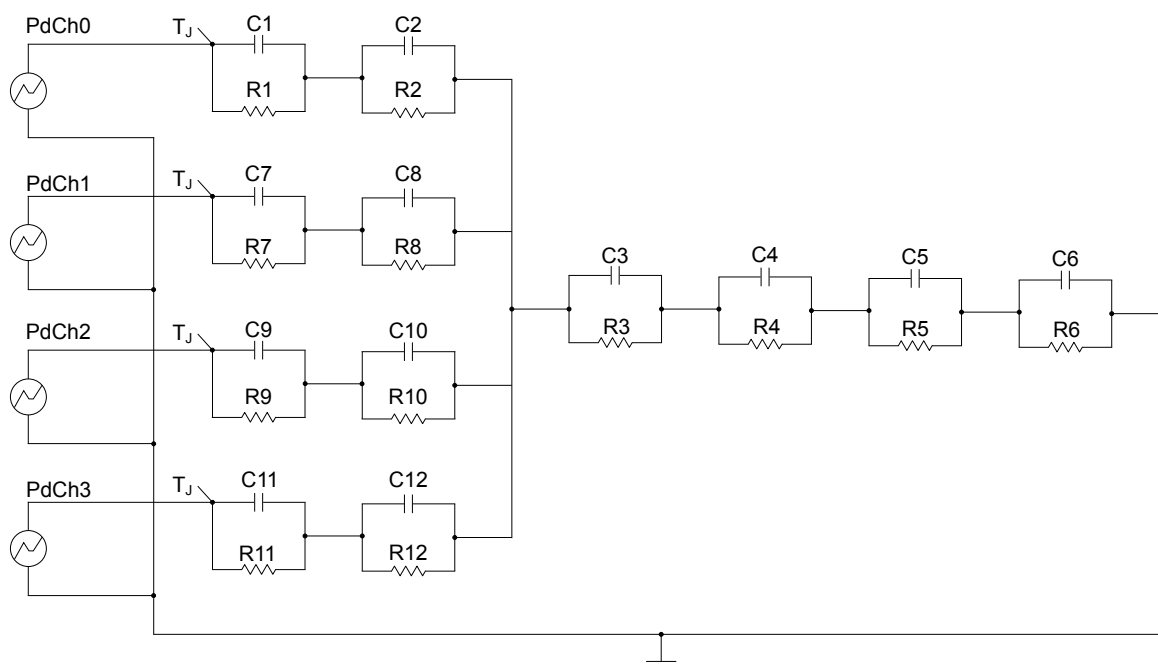


Figure 44. QFN 6x6 thermal impedance junction ambient for channels 2 and 3



$$Z_{th\delta} = R_{th} \cdot \delta + Z_{thp} (1 - \delta)$$

where  $\delta = t_p/T$

**Figure 45. Thermal fitting model**


GADG050120230849GT

**Note:** The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

**Table 67. Thermal parameters**

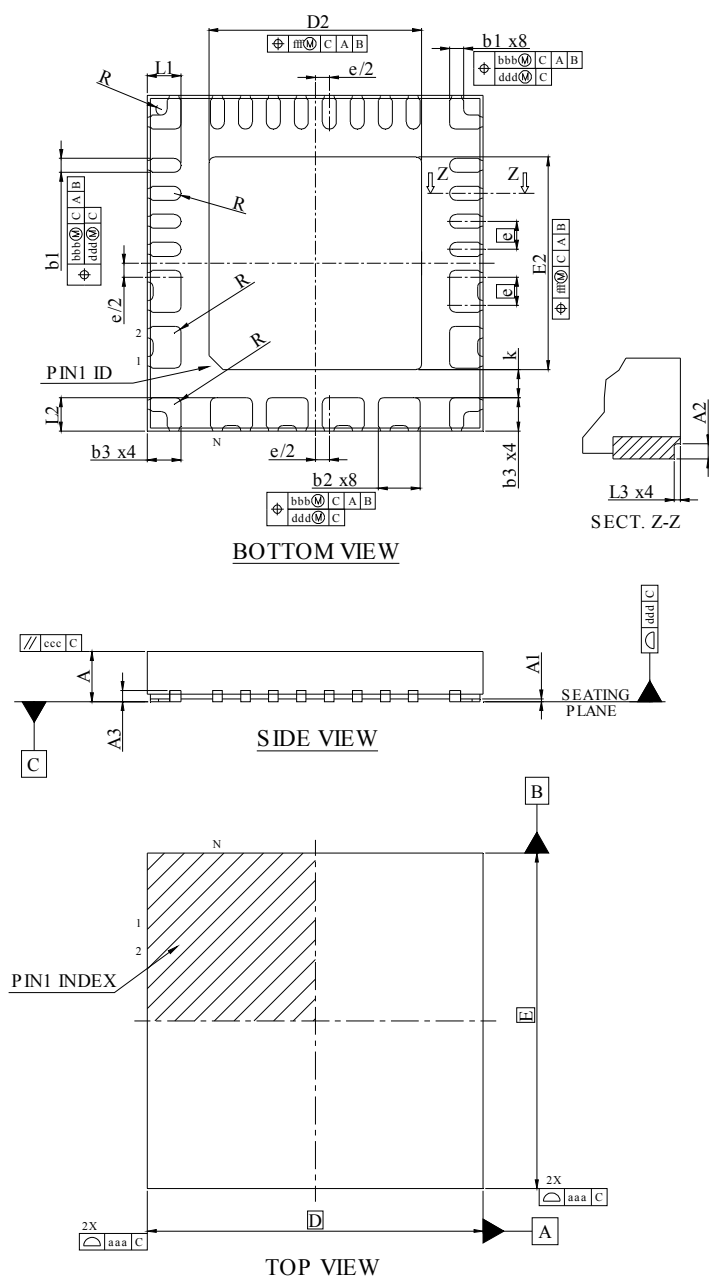
Area/island (cm <sup>2</sup> )	FP	2	8	4L
R1 = R11 (°C/W)	0.4			
R2 = R12 (°C/W)	1.5			
R3 (°C/W)	4.5	4.5	4.5	3.5
R4 (°C/W)	6	5	5	4
R5 (°C/W)	21	15	10	5.5
R6 (°C/W)	24	17.5	14	5.7
R7 = R9 (°C/W)	0.9			
R8 = R10 (°C/W)	3.3			
C1 = C11 (W·s/°C)	0.00055			
C2 = C12 (W·s/°C)	0.006			
C3 (W·s/°C)	0.05			
C4 (W·s/°C)	0.3			
C5 (W·s/°C)	1	1.2	1.4	1.4
C6 (W·s/°C)	2.4	3.5	8	15
C7 = C9 (W·s/°C)	0.0002			
C8 = C10 (W·s/°C)	0.001			

## 12 Package information

To meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions, and product status are available at: [www.st.com](http://www.st.com). ECOPACK is an ST trademark.

### 12.1 QFN 6x6 package information

**Figure 46. QFN 6x6 package outline**



DM00346180\_2

**Table 68. QFN 6x6 mechanical data**

Dim.	Millimeters		
	Min.	Typ.	Max.
A	0.80	0.90	1.00
A1	0.00	-	0.05
A2	0.10		
A3	0.20 REF.		
b1	0.20	0.25	0.30
b2	0.70	0.75	0.80
b3	0.50	0.60	0.70
D	6.00 BSC		
E	6.00 BSC		
e	0.50 BSC		
L1	0.50	0.60	0.70
L2	0.50	0.60	0.70
L3			0.05
k	0.45		
R			0.10
N	32+4		

**Table 69. QFN 6x6 tolerance of form and position**

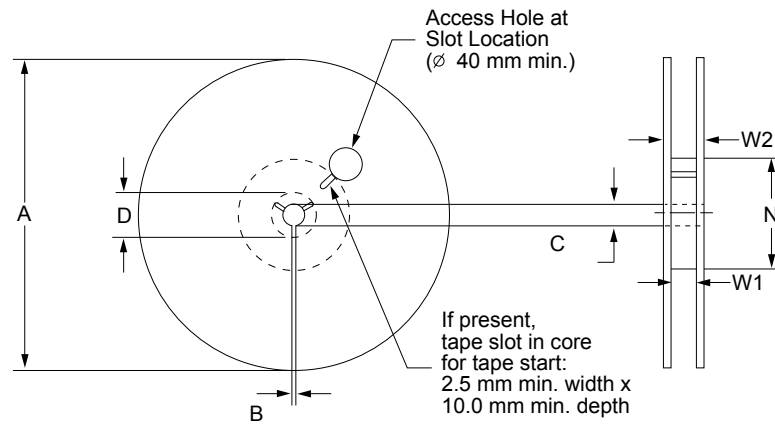
Dim.	Millimeters
aaa	0.15
bbb	0.10
ccc	0.10
ddd	0.05
eee	0.08
fff	0.10

**Table 70. QFN 6x6 variations**

Dim.	Millimeters			OPT.
	Min.	Typ.	Max.	
D2	3.70	3.80	3.90	A
E2	3.70	3.80	3.90	

## 12.2 QFN 6x6 packing information

**Figure 47. QFN 6x6 reel 13"**



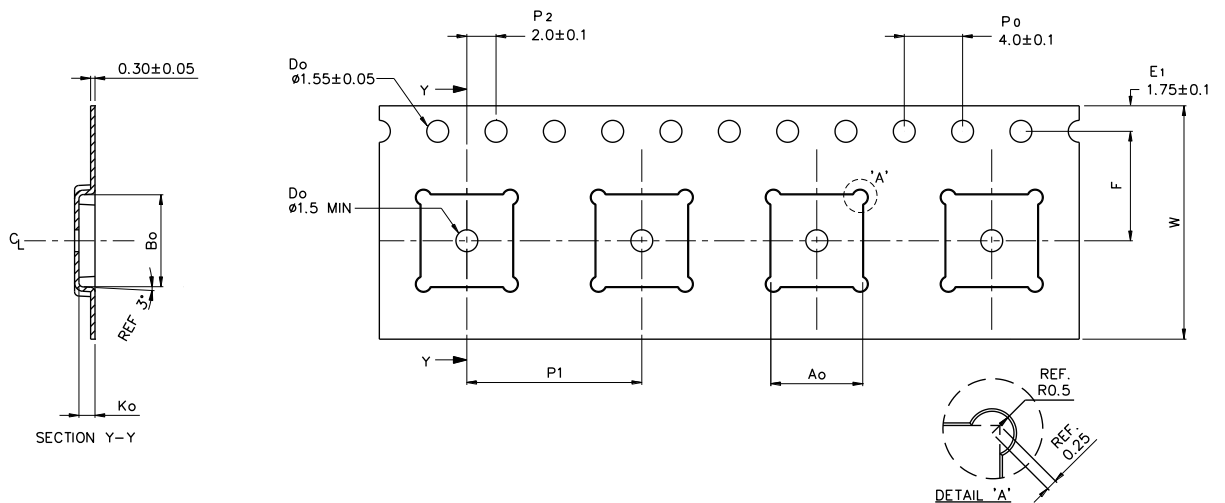
TAPG2004151655CFT

**Table 71. Reel dimensions**

Description	Value <sup>(1)</sup>
Base quantity	3000
Bulk quantity	3000
A (max)	330
B (min)	1.5
C (+0.5, -0.2)	13
D (min)	20.2
N	178
W1	146.4
W2	22.4

1. All dimensions are in mm.

**Figure 48. QFN 6x6 carrier tape**



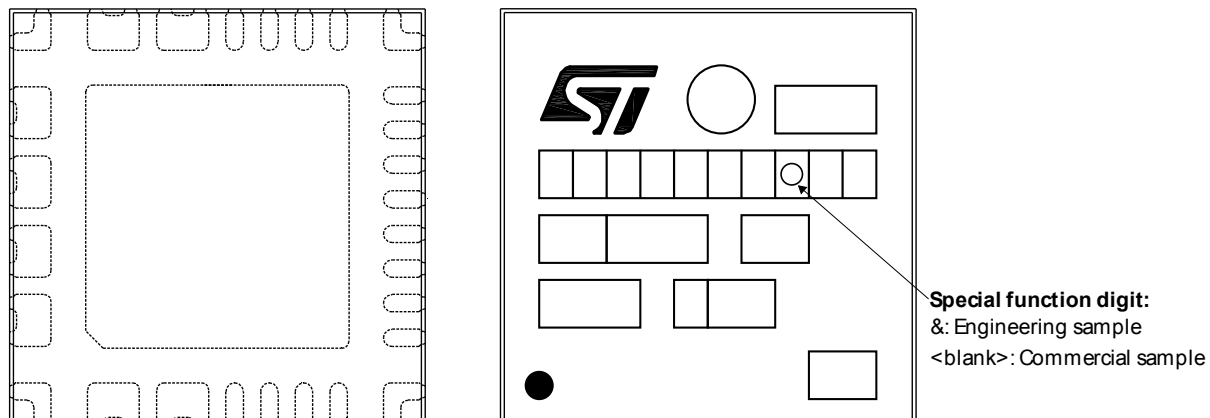
**Table 72. QFN 6x6 carrier tape dimensions**

Description	Value <sup>(1)</sup>
A <sub>0</sub>	6.30 ± 0.1
B <sub>0</sub>	6.30 ± 0.1
K <sub>0</sub>	1.10 ± 0.1
F	7.50 ± 0.1
P <sub>1</sub>	12.00 ± 0.1
W	16.00 ± 0.3

1. All dimensions are in mm.

## 12.3 QFN 6x6 marking information

Figure 49. QFN 6x6 marking information



Parts marked as '&' are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.



## Revision history

**Table 73. Document revision history**

Date	Version	Changes
21-Dec-2020	1	Initial release.
16-Apr-2021	2	Updated <i>VN9D5D20FN package silhouette</i> in cover page. Updated note in <i>Figure 2. Connection diagram (top view)</i> . Updated <i>Table 44. Absolute maximum ratings</i> . Updated <i>Table 51. Protection</i> . Added <i>Section 7.8 Waveforms</i> . Added <i>Section 10 Maximum demagnetization energy (VCC = 16 V)</i> . Added <i>Section 11.2 QFN 6x6 packing information</i> and <i>Section 11.3 QFN 6x6 marking information</i> . Minor text changes.
07-Jun-2021	3	Updated <i>Table 57. Bulb - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 0,1, TJ = -40 °C to 150 °C)</i> and <i>Table 58. Bulb - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 2,3, TJ = -40 °C to 150 °C)</i> . Updated <i>Table 62. LED - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 0, 1; TJ = -40 °C to 150 °C)</i> and <i>Table 63. LED - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 2, 3; TJ = -40 °C to 150 °C)</i> . Minor text changes.
02-Sep-2021	4	Modified <i>Figure 3. Device state diagram</i> . Modified <i>Section 2.2.4 Normal mode</i> and <i>Section 3.2 Junction overtemperature (OT)</i> . Modified <i>Table 57. Bulb - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 0,1, TJ = -40 °C to 150 °C)</i> , <i>Table 58. Bulb - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 2,3, TJ = -40 °C to 150 °C)</i> , <i>Table 62. LED - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 0, 1; TJ = -40 °C to 150 °C)</i> and <i>Table 63. LED - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 2, 3; TJ = -40 °C to 150 °C)</i> .
19-Jan-2022	5	Updated <i>Table 24. ROM memory map</i> , <i>Table 51. Protection</i> , <i>Table 52. Open-load detection (7 V &lt; VCC &lt; 18 V)</i> , <i>Table 54. Bulb - power section</i> , <i>Table 55. Bulb - switching (VCC = 13 V, normal switch mode)</i> , <i>Table 57. Bulb - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 0,1, TJ = -40 °C to 150 °C)</i> , <i>Table 58. Bulb - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 2,3, TJ = -40 °C to 150 °C)</i> , <i>Table 60. LED - switching (VCC = 13 V; Normal switch mode)</i> , <i>Table 61. LED - protection and diagnosis</i> , <i>Table 62. LED - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 0, 1; TJ = -40 °C to 150 °C)</i> and <i>Table 63. LED - Digital Current Sense (7 V &lt; VCC &lt; 18 V, channel 2, 3; TJ = -40 °C to 150 °C)</i> . Modified <i>Section 4.6.1 OUTCTRCRx</i> and <i>Section 4.6.9 ADC9SR</i> .
28-Sep-2022	6	Updated <i>Table 2. Operating modes</i> , <i>Section 2.2.2 Reset mode</i> , <i>Section 2.2.3 Fail Safe mode</i> , <i>Section 2.2.7 Sleep mode 2</i> , <i>Section 3.2 Junction overtemperature (OT)</i> , <i>Section 3.3 Power limitation (PL)</i> , <i>Section 4.3 Register map</i> , <i>Section 4.4 Outputs control</i> , <i>Section 4.4.3 Procedure to Turn-ON the outputs with the Direct Input Dlx</i> , <i>Section 4.6.1 OUTCTRCRx</i> , <i>Section 4.6.2 OUTCFGRx</i> , <i>Section 4.6.3 CHLOFFTCR0</i> , <i>Section 4.6.4 CHLOFFTCR1</i> , <i>Section 4.6.5 SOCR</i> , <i>Section 4.6.7 OUTSRx</i> , <i>Section 4.6.8 ADCxSR</i> , <i>Section 4.6.10 OTP memory map (Reserved)</i> , <i>Section 5.1.2 ADC operating principle</i> , <i>Section 5.3 Digital diagnostic</i> , <i>Section 5.4 Overload (VDS high voltage, Overload (OVL))</i> , <i>Section 5.6 Open-load OFF-state detection</i> , <i>Table 43. Time values written by MCU and their real value in timer register</i> , <i>Section 6.5 Registers</i> and <i>Section 7.5 PWM unit</i> . Minor text changes.
13-Apr-2023	7	Updated <i>Table 24</i> , <i>Table 33</i> , <i>Table 54</i> , <i>Table 55</i> and <i>Table 60</i> . Updated <i>Section 11 Package and PCB thermal data</i> . Minor text changes.
21-Apr-2023	8	Updated <i>Table 44</i> and <i>Table 55</i> .
02-Feb-2024	9	Updated <i>Section 4.4.3: Procedure to turn-ON the outputs with the direct input Dlx</i> .

Date	Version	Changes
20-Dec-2024	10	Updated <a href="#">Table 44. Absolute maximum ratings</a> and <a href="#">Table 45. Thermal data</a> .
10-Apr-2025	11	Updated <a href="#">Table 44. Absolute maximum ratings</a> and <a href="#">Table 65. Component values</a>

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