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June 2013

FLS0116 MOSFET Integrated Smart LED Lamp Driver IC with PFC Function

Features

- Built-in MOSFET(1 A / 550 V)
- . **Digitally Implemented Active-PFC Function**
- No Additional Circuit for Achieving High PF
- Application Input Range: 80 VAC ~ 308 VAC
- Built-In HV Supplying Circuit: Self Biasing
- AOCP Function with Auto-Restart Mode
- Built-In Over-Temperature Protection (CTP)
- Cycle-by-Cycle Current Limit
- Current Sense Pin Open Prot Jon
- Low Operating Current: 0.85 A (Typ. I)
- Under-Voltage Lock h 5 Hystr esis
- . Programmable Osc ation Frequency
- . ProgrammrED vrrent
- . Analog D. min nch
- Soft-S 't Function
- ise ern seference:

lic ions

- LED Lamp for Decorative Lighting
- LED Lamp for Low-Power Lighting Fixture

Ordering Information

| Part Number | Operating Temperature Range | Package | Packing Method |
|-------------|--------------------------------|--|----------------|
| FLS0116MX | -40°C to +125°C | 7-Lead, Small-Outline Integrated Circuit (SOIC), JEDEC MS-012, .150-inch, Narrow Body | Tape & Reel |

Description

mp tribe a simple IC with The FLS011 LED integrater Mc `FET d rC for such the special "adopted nital 'ech' que automatically detects input vol ron, ion sends an internal reference signal power factor. When AC input is applied ac. ?ve. tu the C, the PFC function is a turnatically enabled. Wind D input is applied to the C, the FFC function is auto, atically disabled. The FLS0116 ones not need a bulk (electronic) capacito for supply fail stability, which significantly improves LED lamp life.



Figure 1. Typical Application

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Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

| Symbol | Par | rameter | | | | | Min. | Max. | Unit |
|---------------------------|---|--------------------|------------|------------|-----------|--------|----------------------------------|----------|--------------------|
| V _{CC} | IC Supply Voltage | | | | | | | 20 | V |
| ΗV | High Voltage Sensing | | | | | | | 550 | V |
| DRAIN | Internal Drain Voltage | | | | | | | 550 | V |
| V_{ADIM} | Analog Dimming | | | | | | | 5 | V |
| V _{RT} | RT Pin Voltage | | | | | | | 5 | V |
| V _{CS} | Allowable Current Sensing Detection | n Voltage | | | | | | | V |
| T _A | Operating Ambient Temperature Ra | nge | | | | | -40 | 25 | °C |
| TJ | Operating Junction Temperature | | | | | | -40 | +150 | °C |
| T _{STG} | Storage Temperature Range | | | | | | o5 | +150 | °C |
| θ _{JA} | Thermal Resistance Junction-Air ^(1,2) | | | | | | 1 | 135 | °C/W |
| PD | Power Dissipation | | | | | | $\overline{\boldsymbol{\Sigma}}$ | 660 | mW |
| FOD | | Humar [*] | dy Too | a JESL | 2-A114 | | | 2000 | |
| ESD | Electrostatic Discharge Capability | Ch .g. | Devic | 100.JI, JE | S022-C | 101 | 20 | 1000 | V |
| 3. | nal resistance test board. Size: 76 ne no ambient airflow. | m x 14.3 | mı . x 1.6 | 5 mra (15) | 0P); JEDI | EC sta | ndara: J | ESD51-2, | JESD5 |
| . Therm 3. 2. Assur | | | m x 1.6 | | 0P); JEDI | EC sta | ndirei | ESD51-2, | JESD5 [.] |

FLS0116 — MOSFET Integrated Smart LED Lamp Driver IC with PFC Function

Electrical Characteristics

Typical values are at $T_A = +25^{\circ}$ C. Specifications to -40° C ~ 125° C are guaranteed by design based on final characterization results.

| Symbol | Parameter | Condition | Min. | Тур. | Max. | Unit |
|-------------------------|---|--|-------|-------|-------|--------|
| V _{cc} Bias Se | ection | | | 1 | 1 | |
| Vcc | V _{CC} Regulator Output Voltage | V _{HV} =100 V _{DC} | 14.0 | 15.5 | 17.0 | V |
| V _{CCST+} | UVLO Positive-Going Threshold | V _{cc} Increasing | 12 | 13 | 14 | V |
| V _{CCST-} | UVLO Negative-Going Threshold | V _{cc} Decreasing | 7 | 8 | 9 | V |
| V _{CCHYS} | UVLO Hysteresis | | 4 | 5 | 6 | V |
| I _{HV} | HV Pin Current | V _{HV} =100 V _{DC} , RT=Open | | 0.85 | 1.20 | mA |
| I _{ST} | Startup Current | | | 1 | 07 | μA |
| Switching | Section | | | | | C |
| | | R _T =5.95 kΩ | 206 | 250 | 300 | kHz |
| fosc | Operating Frequency | R _T =87 kΩ | | 20 | 24 | kHz |
| | | R _T Open | | .15.0 | 49.5 | kHz |
| t _{MIN} | Minimum On Time ⁽³⁾ | | | 400 | | ns |
| D _{MAX} | Maximum Duty Cycle | | 747 | 5 ` | 20 | % |
| t _{LEB} | Leading Edge Blanking Time ⁽³⁾ | | 0 | 350 | | ns |
| V _{RT} | Voltage Reference of RT Pin | | 2 | 0.5 | l . | V |
| Soft-Start S | Section | | |) | | |
| | | DC Node | 18 | 60 | 72 | ms |
| t _{ss} | Soft-Start T | AC Mode | | 7 | | Period |
| Reference | Ser' | <u>777201</u> | | | | |
| V _{CS1} | | DC Mode | 0.354 | 0.365 | 0.376 | |
| V | Internal Representation Voltage of CS Pin | A.C. Moce ⁽³⁾ | 0.485 | 0.500 | 0.515 | V |
| rtectic | Se ion S | | | | | |
| C Pvcr | Over-Vollage Protection on VCC Fits | | 17.7 | 18.7 | 19.7 | V |
| VALOP | Abhormal OCE Level at CS Pho ⁽²⁾ | | | 2.5 | | V |
| tAOCP | Abnormal Detection Time ⁽³⁾ | | | 70 | | ns |
| TISOL | Thermal Shutdown Threshold ⁽³⁾ | | 140 | 150 | | °C |
| TISDHY | Thermal Shutdown Threshold Hysteresis ⁽³⁾ | | | 50 | / | °C |
| Dimming S | ection | | | | 1 | |
| V _{ADIM(ST+)} | Analog Dimming Positive Going Threshold ⁽³⁾ | | 3.15 | 3.50 | 3.85 | V |
| V _{ADIM(ST-)} | Analog Dimming Negative Going Threshold ⁽³⁾ | | | 0.50 | 0.75 | V |
| I _{AD} | Internal Current Source for ADIM Pin | | 9 | 12 | 15 | μA |

Continued on the following page...

Electrical Characteristics (Continued)

Typical values are at T_A = +25°C. Specifications to -40°C ~ 125°C are guaranteed by design based on final characterization results.

| Symbol | Parameter | Condition | Min. | Тур. | Max. | Unit |
|----------------------------|---|---|-----------------------|------------|------|------|
| MOSFET S | ection | | | | | |
| BV _{DSS} | Breakdown Voltage | V _{CC} =0 V, I _D =250 µA | 550 | | | V |
| I _{LKMOS} | Internal MOSFET Leakage Current | V_{DS} =550 V_{DC} , V_{GS} =0 V | | | 250 | μA |
| R _{ON(ON)} | Drain-Source On Resistance ⁽³⁾ | V_{GS} =10 V, V_{DGS} =0 V, T_{C} =25°C | | 7.3 | 10.0 | Ω |
| CISS | Input Capacitance ⁽³⁾ | V _{GS} =0 V,V _{DS} =25 V, f=1 MHz | | 135 | | pF |
| Coss | Output Capacitance ⁽³⁾ | V _{GS} =0 V,V _{DS} =25 V, f=1 MHz | | | | pF |
| C _{RSS} | Reverse Transfer Capacitance ⁽³⁾ | V _{GS} =0 V,V _{DS} =25 V, f=1 MHz | | | | pr |
| t _{d(ON)} | Turn-On Delay ⁽³⁾ | V _{DD} =350 V, I _D =1 A | | 10 | | ns |
| tr | Rise Time ⁽³⁾ | V _{DD} =350 V, I _D =1 A | , Ti | 1 4 | 110 | ns |
| t _{d(OFF)} | Turn-Off Delay ⁽³⁾ | V _{DD} =350 V, I _D =1 A | | 14.9 | 110 | ns |
| t _f | Fall Time ⁽³⁾ | V _{DD} =350 V, I _D =1 A | | ડછે.૬ | | ns |
| Note: 3. These p | parameters, although guaranteed, are | not 10° 5 to ted procosion | FO | sem | | 4 |
| | | MAENDER | FO Por NFO | sem RMA | | 2 |
| | | MAENDER | reor reor | RMA | | 2 |
| | | MAENDER | FO Ron NFO | RMA | | 1 |
| | | MAENDER | r or r or v F O | RMA | | 1 |
| | | MAENDER | FOR | RMA | | 1 |
| | | MAENDER | FO Por NFO | RMA | | 1 |
| | | MAENDER | r or r or NFO | RMA | | 1 |
| | | MAENDER | reor pron | RMA | | 1 |
| | | not 10° 'S t. 'ed. procion | FO Ron NFO | RMA | | 1 |

Note:

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Functional Description

The FLS0116 is a basic PWM controller for buck converter topology in Continuous Conduction Mode (CCM) with an intelligent PFC function that uses a digital control algorithm. An internal self-biasing circuit uses the high-voltage switching device. The IC does not need an auxiliary powering path to the VCC pin typical in flyback control ICs or PSR product family.

When the input voltage applied to the HV pin is within operating range (25 V to 500 V), the FLS0116 maintains a 15.5 V DC voltage at the VCC pin for stable operation. The UVLO block functions such that when the V_{CC} voltage rises higher than V_{CCST+}, the internal UVLO block releases and starts operation. Otherwise, the V_{CC} goes down to the V_{CCST} and IC operation stops. Normally, the hysteresis function provides stable operation even if the input voltage is operating under very noisy or unstable circumstances.

The FLS0116 has a "smart" internal digital block for determining input condition: AC or DC. When an AC source with 50 Hz or 60 Hz is applied to the IC, the IC automatically changes its internal reference signal, which is similar to input signal, for creating high power factor. When a DC source connects to the IC, th internal reference immediately changes to DC.

Soft-Start Function

The FLS0116 has an internal soft start inc. o reduce inrush current at startup when the C starts operation following an internal quence the internal reference slowly increase a re-det mined fixed time. After this transier period, the inder reference goes to a steady-stal level. In this time, the IC continually tries and place in imation from the VCC pin. If the IC rece in galling phase information, it automatic. I follows a similar shape reference made durin the times, 7 periods. If net, the IC has a For reference yet.



Figure 4. Soft-Start Function in AC Input Mode

Internal PFC Function: How to Achieve High Power Factor

The FLS0116 has a simple, "smart", internal PFC function that does not require additional pins for detecting input phase information or an electrolytic capacitor for supply voltage stabilization. For achieving high PF, the FLS0116 does not use the rectification capacitor after the bridge diode. This is important because the IC instead uses fluctuation in the signal on the VCC pin. Basically, the VCC pin, which is supplies

power for the IC, has voltage ripple as well as the rectification voltage after bridge, changing voltage level according to the V_{CC} capacitor value. Using this kind of voltage fluctuation on the VCC pin, the IC can detect the time reference and create the internal ZCD signal.

For precise and reliable internal reference for input voltage signal, the FLS0116 uses a digital technique (sigma/delta modulation) and creates a new internal signal (DAC_OUT) that has the same phase as the input voltage, as shown in Figure 5. This signal enters the final comparator and is compared with current information from the sensing resistor



Self-Bicsing Function

The self-biasing function, using an HV device, can supply enough operating current to the IC and guarantee similar startup time across the whole input voltage range ($80 V \sim 308 V_{AC}$). However, self-biasing has a weakness in high-voltage condition. Normally, he HV device acts as constant current source, so the internal HV device has power loss when high input voltage connects to the HV pin. This power loss is proportional to input voltage. To reduce this power loss, one of the possible solutions is an additional resistor between the input voltage source and the HV pin, as shown in Figure 6.



Figure 6. High-Voltage Application

(4)

Dimming Function

The FLS0116 uses the ADIM pin for analog or 0 V to 10 V dimming by using a resistive divider. The peak voltage of internal reference, which is DAC_OUT signal in Figure 5, is changed by the V_{ADIM} level, as shown in Figure 7, and has different peak level according to the operating mode.



Figure 7. VADIM VS. VDAC_OUT(peak)

Inductor Design

The fixed internal duty ratio range is below 50%, or around 400 ns, from a timing point of view. The range is dependent on the input voltage and number of LEP, its string.

Minimum duty is calculated as:

 $n \cdot V$ $D_{\min} =$ $\eta \cdot V_{in(\max)}$ where: = efficiency of stem: η V_{IN(max)}= maxim int voltaç V_{f} = for and drop v per of LED; and in series connection. LE. n rrei. Avera ge 2D Cost ant (a) DC:n Mode current neak at LED current maximum point (ILED(peak)) Average LED Current (I_{LED(ave)}) Current peak at LED average current maximu тp (ILED(ave.peak)) Current min at LED curren maximum point (ILED(n ton tor (b) CCM Mode Figure 8. DCM and CCM Operation

In DCM Mode, inductance is:

$$L_m = \frac{n \cdot V_f \cdot (1 - D_{\min})}{f_s \cdot \Delta i_{rip}} [H]$$
⁽²⁾

If the peak current is fixed at 350 mApk, the formula for the peak current is:

$$I_{LED(ave.peak)} = \Delta i_{con} + \frac{\Delta i_{rip}}{2} \quad [A]$$

In FL7701, the LED RMS current determines the inductance parameter. To drive for CCM Mode, define LED RMS current first, as:

$$I_{LED(rms)} = \frac{I_{LED(ave.peak)}}{\sqrt{2}} \quad [A]$$

Substituting Equation (2) for an on (4) the inductance of inductor in the med.



Figure 9. Typical Performance Characteristics











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