

Series String Pixel Controller for Automotive (Front) Lighting

NCV78343

Introduction

The NCV78343 is a single-chip pixel controller with embedded switches to control individual LEDs in a series LED string, designed for automotive dynamic lighting applications and in particular for high current LEDs. In order to create a pixel lighting solution, the LEDs need to be powered by current sources such as NCV78763 or NCV78723. The NCV78343 pixel controller devices receive the pixel control parameters from the pixel light ECU which translates the required light pattern or light image into individual pixel dimming data.

One pixel controller device can control up-to 12 pixels of 1× or 2× 1.4 A LEDs per pixel. The maximum LED string voltage has to be limited to 60 V.

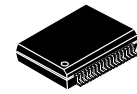
When more than 12 pixels are to be controlled, multiple pixel controllers can be combined in a single system.

The NCV78343 uses two communication interfaces for connection with a microcontroller. A universal asynchronous receiver transmitter (UART), which supports the use of CAN transceiver and multipoint low voltage differential signaling (M-LVDS) for either local connection or connection with the MCU.

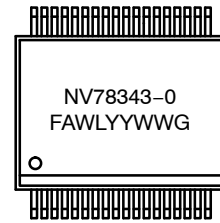
Features

- Single Chip Compatible with IMS Board (Single Layer)
- 12 Integrated Switches with Multiple Configuration Options
- Minimum of External Components
- Communication Interfaces to the Pixel Light ECU via
 - ♦ Integrated M-LVDS
 - ♦ UART over CAN Interface
 - ♦ Integrated Bridge between M-LVDS and UART
 - ♦ Supports up to 32 Devices, 1Mbaud
- No Need for Local MCU and Precise Clock
 - ♦ Interface to External I2C EEPROM
 - ♦ Integrated 8 bit Analog to Digital Converter
- Dimming Controller
 - ♦ PWM + Phase Shift Unit per Channel
- Over Temp Protection
- Individual Open/Short/OV LED Diagnostic Feedback
- Open LED Failure Automatic Bypass
- This is a Pb-Free Device
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC-Q100 Qualified and PPAP Capable

MARKING DIAGRAM



SSOP36 EP
CASE 940AB



NV78343 = Specific Device Code
 F = Fab Indicator
 A = Assembly Location
 WL = Wafer Lot
 YYWW = Year / Work Week
 G = Pb-Free Designator

SAFETY DESIGN – ASIL B

ASIL B Product developed in compliance with ISO 26262 for which a complete safety package is available.

ORDERING INFORMATION

Device	Package	Shipping†
NCV78343DQ0R2G	SSOP36 EP (P-Free)	1500 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

Typical Applications

- Dynamic Adaptive Driving Beam Functions
 - ♦ Glare-free High Beam
 - ♦ Static Swiveling
 - ♦ Beam Shaping
 - ♦ Light Power Adjustment
- Animated Welcome Functions on Signal Lights
- Wiping Blinker

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PACKAGE AND PIN DESCRIPTION

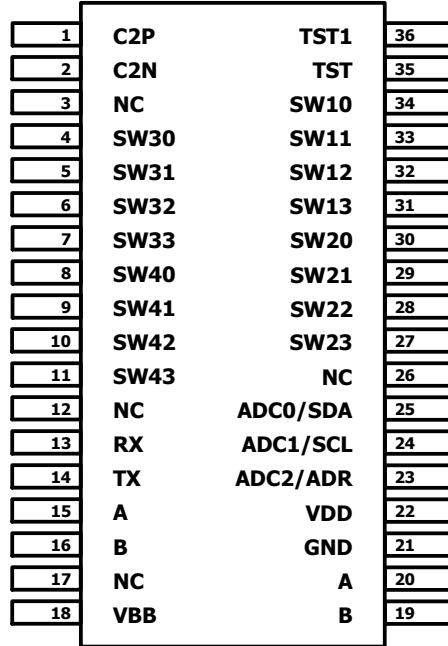


Figure 1. Pin Connections – SSOP36-EP (Top View)

Table 1. PIN DESCRIPTION

Pin No. SSOP36-EP	Pin Name	Description	I/O Type
1	C2P	Switch control capacitor connection	HV in/out
2	C2N	Switch control capacitor connection	HV in/out
3, 12, 17, 26	NC	Not used (to be left floating)	NC
31, 32, 33, 34	SW1y	Power switch to short LED	HV in/out
27, 28, 29, 30	SW2y	Power switch to short LED	HV in/out
4, 5, 6, 7	SW3y	Power switch to short LED	HV in/out
8, 9, 10, 11	SW4y	Power switch to short LED	HV in/out
13	RX	Receive data input (To be tied to GND when not used)	HV60 in
14	TX	Transmit data output (To be tied to GND or left floating when not used)	MV out
15, 20	A	M-LVDS IO pins (internally connected; to be shorted to B when not used)	MV in/out
16, 19	B	M-LVDS IO pins (internally connected; to be shorted to A when not used)	MV in/out
18	VBB	Battery supply	HV60 supply
21	GND	Ground	Ground
22	VDD	3V analog and logic supply	LV supply
23	ADC2/ADR	ADC input 2 / Address	LV in
24	ADC1/SCL	ADC input 1 / I2C clock	LV in/out
25	ADC0/SDA	ADC input 0 / I2C data	LV in/out
35	TST	Internal function. To be tied to GND or left floating	HV70 in
36	TST1	Internal function. To be tied to GND	LV in/out
EP	EP	To be tied to GND	Exposed Pad

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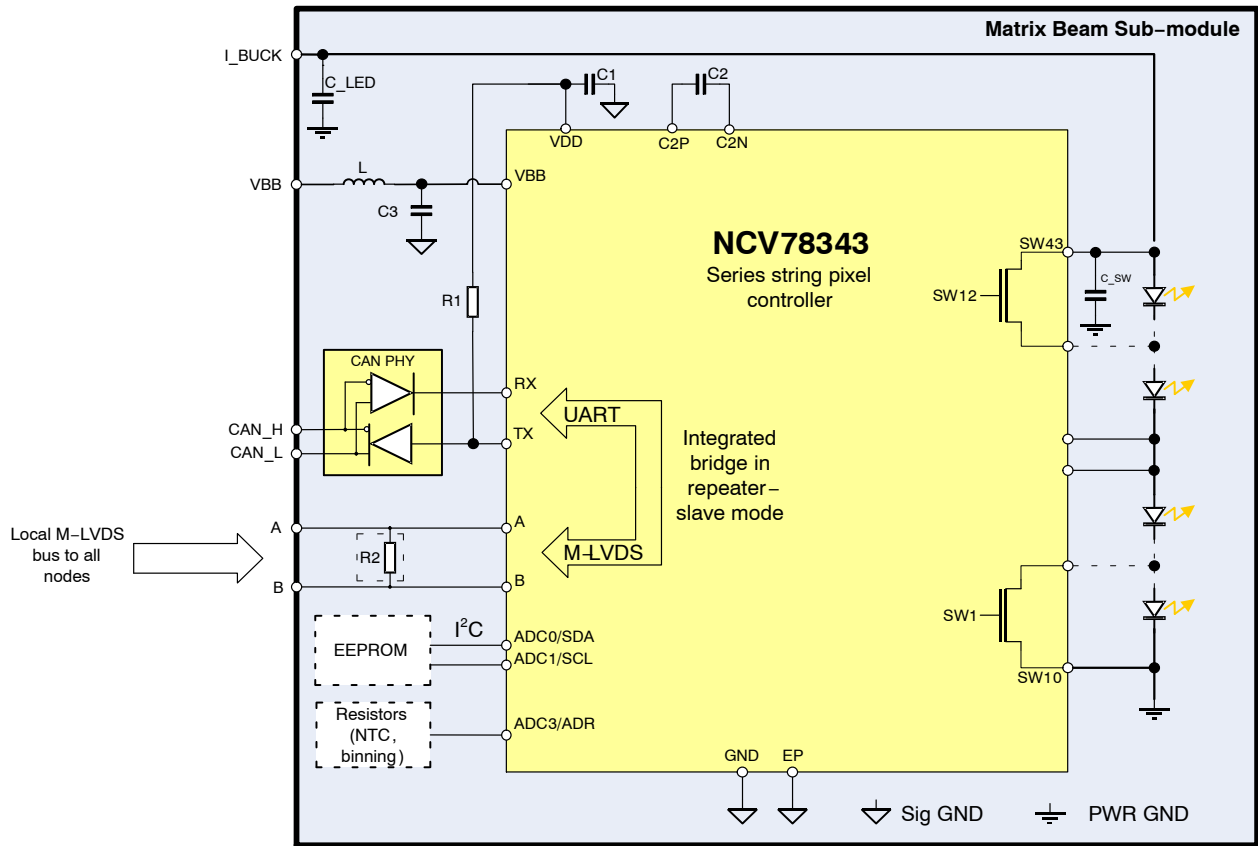


Figure 2. Application Diagram

Table 2. EXTERNAL COMPONENTS

Component	Function	Typ. Value	Unit
C1	Cap. for VDD regulator	470	nF
C2	Cap. for switch control	220	nF
C3	VBB decoupling cap.	100	nF
C_SW	VLED decoupling cap.	22	nF
C_LED	VLED decoupling cap.	22	nF
R1	Tx pull-up resistor	100	kΩ
R2	Terminating resistors (only for the first and last device)	100	Ω
CAN	CAN transceiver	NCV7344	
M-LVDS	M-LVDS transceiver	NBA3N206S	
EEPROM	External EEPROM	CAT24C02	
L	Ferrite bead *	600 @ 100 MHz	Ω

* It is recommended to place a ferrite bead at VBB net close to a VBB decoupling capacitor for a better electromagnetic immunity.

NOTE: Unused switches to be shorted externally. The switches should be grounded if a full section is not used.

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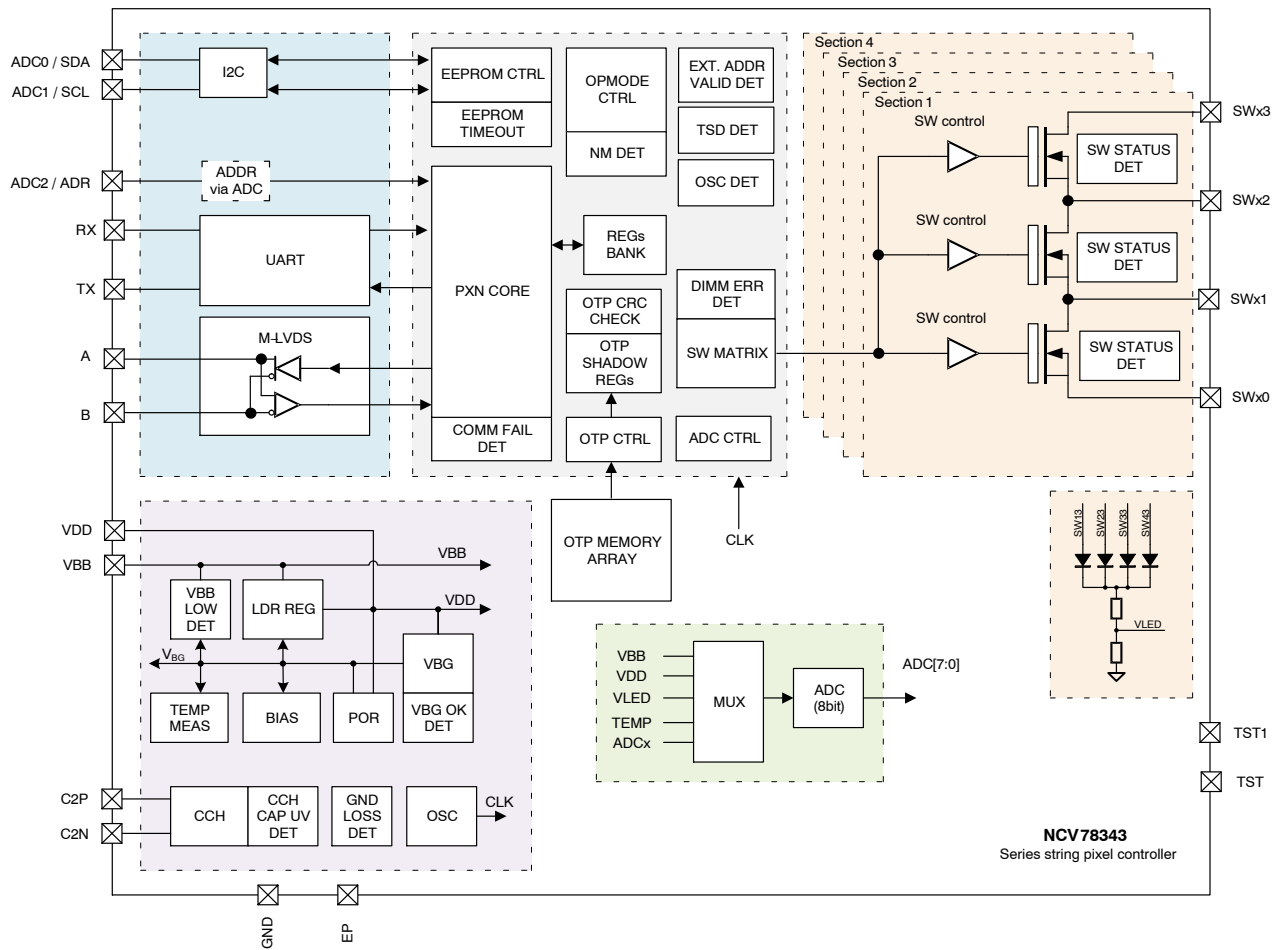


Figure 3. Block Diagram

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The NCV78343 supports two communication interfaces: UART and M-LVDS. It is possible to communicate over both interfaces, where the first example uses the UART interface over CAN physical layer as a master bus from the LED Driver Module to the first NCV78343 chip and the

M-LVDS bus for local connection between submodules of each functional lights such as high beam, low beam, turn indicator, etc. The second example uses the M-LVDS bus only.

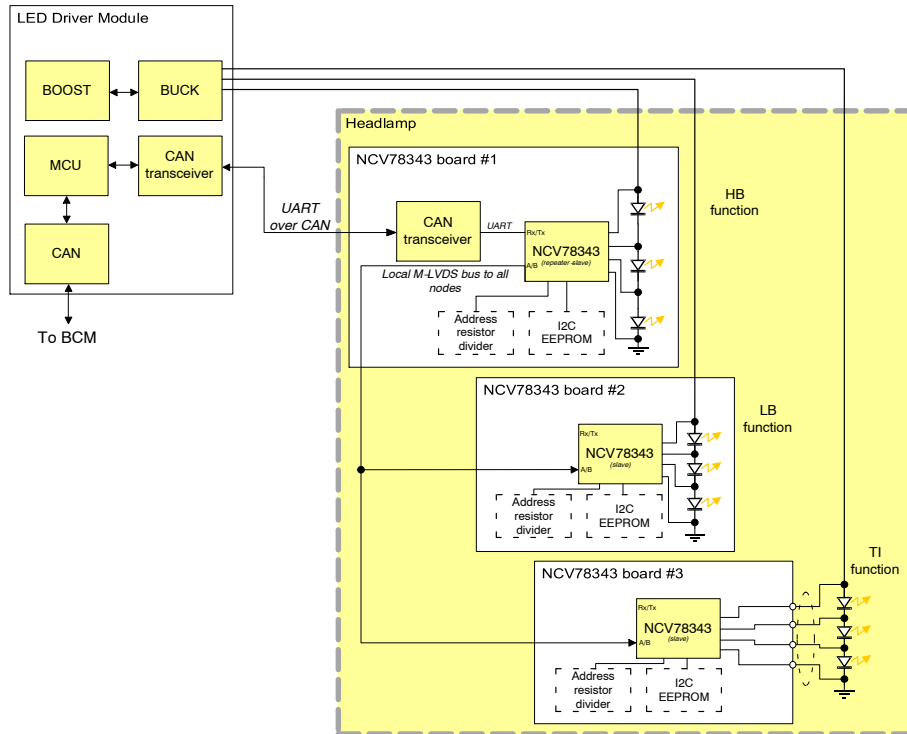


Figure 4. System Architecture using CAN-FD and M-LVDS

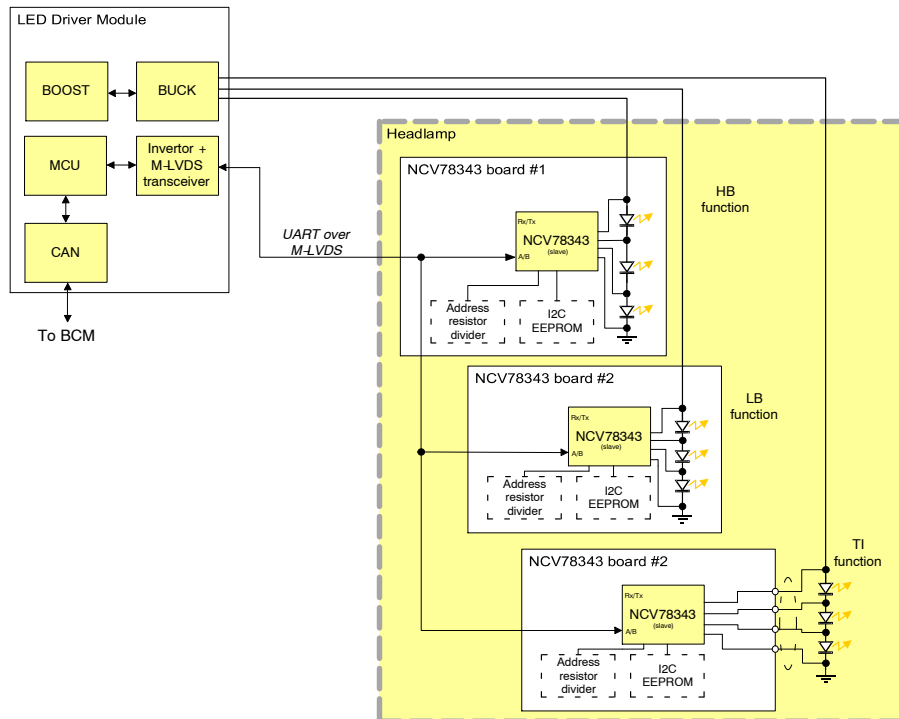


Figure 5. System Architecture using M-LVDS only

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The advantage of sharing common heatsink for higher currents can be reached by placement of the NCV78343 together with the LEDs on same PCB (IMS type of board

supported). This is not necessary for lower currents or application where the LED string is connected over two NCV78343 devices.

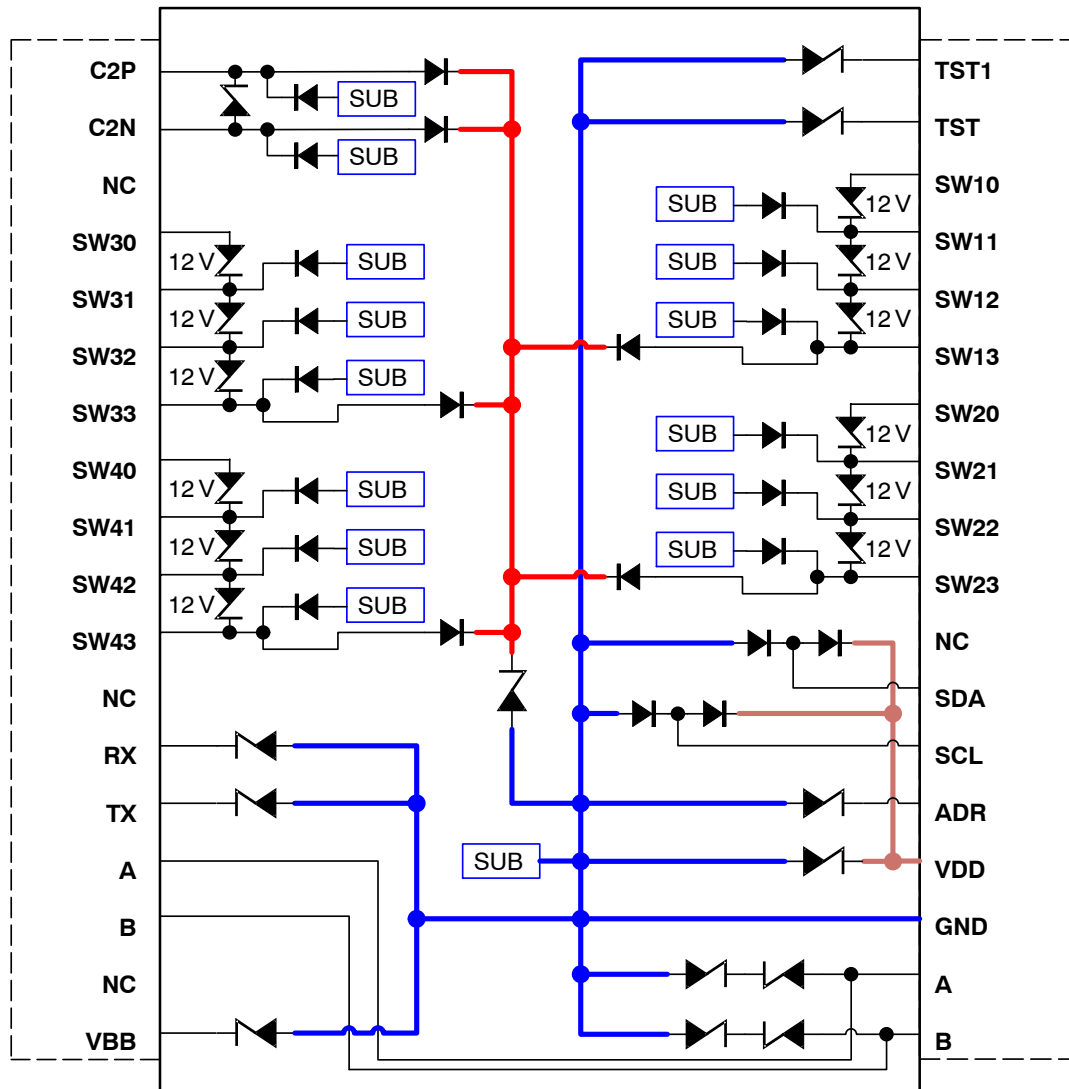


Figure 6. ESD Protection Schematic

Typical Switch Resistance

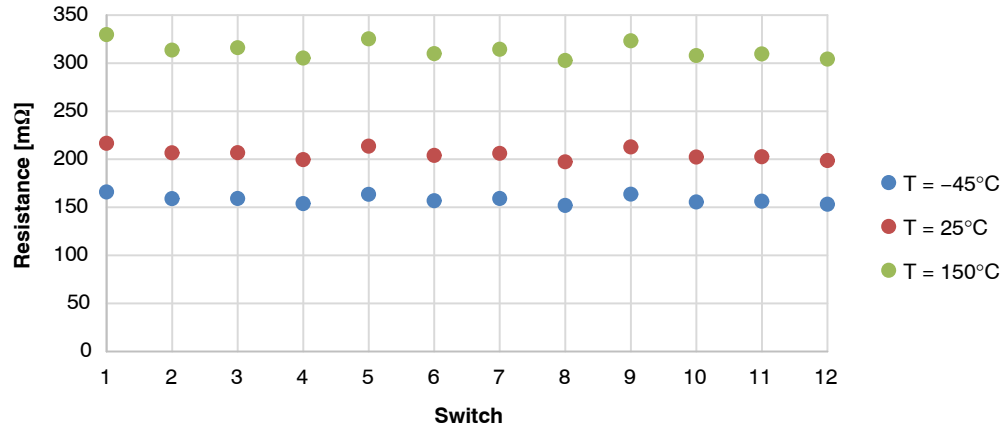


Figure 7. Typical Switch Resistance

Typical Switch Section Resistance

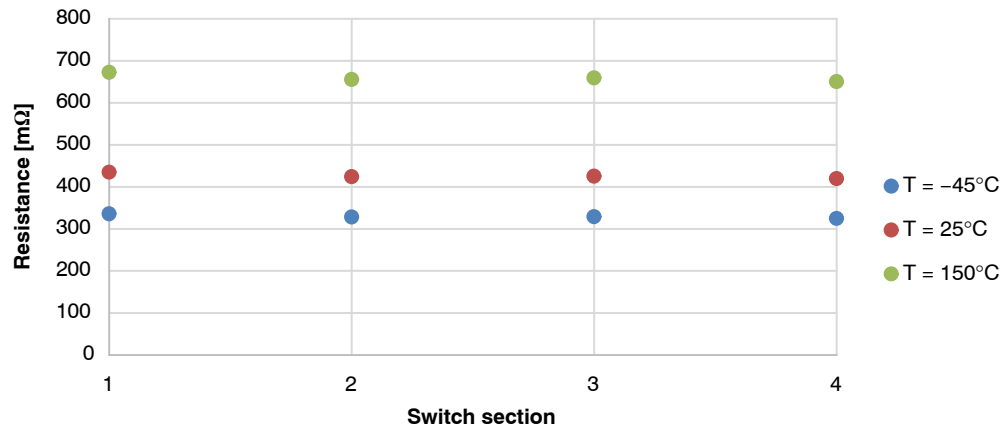


Figure 8. Typical Switch Section Resistance

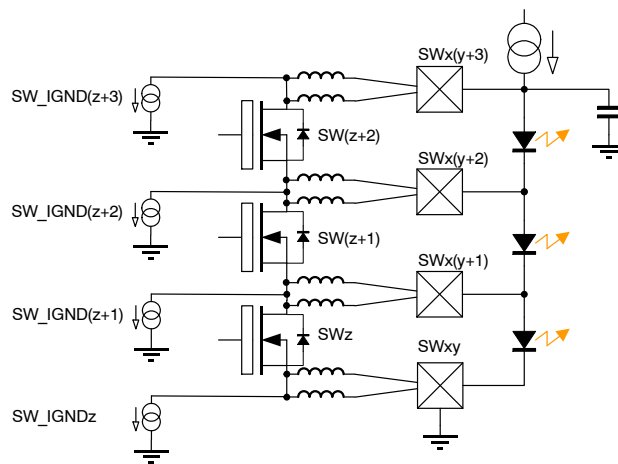


Figure 9. Pixel Switches

Table 3. ABSOLUTE MAXIMUM RATINGS

Characteristic	Symbol	Min	Max	Unit
Battery Supply voltage (Note 1)	V _{BB}	-0.3	60	V
Low voltage supply (Note 2)	V _{DD}	-0.3	3.6	V
High voltage control IO pins (Note 3)	I _{OHV60}	-0.3	60	V
High voltage IO pins (Note 4)	I _{OHV}	-0.3	68	V
Medium voltage IO pins (Note 5)	I _{OMV}	-0.3	6.5	V
Medium voltage IO pins: M-LVDS (Note 6)	I _{OMV_MLVDS}	-1.8	4	V
Low voltage IO pins (Note 7)	I _{OLV}	-0.3	3.6	V
Low voltage supply for switch control: V ₂ = C2P – C2N	V ₂	-0.3	3.6	V
Switch differential voltage (Note 8)	V _{SWxx_DIFF}	-0.3	12	V
Storage Temperature (Note 9)	T _{strg}	-50	150	°C
Electrostatic discharge on component level Human Body Model (Note 10)	V _{ESD_HBM}	-2	+2	kV
Electrostatic discharge on component level Charge Device Model (Note 10)	V _{ESD_CDM}	-500	+500	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Absolute maximum rating for pins: V_{BB}
2. Absolute maximum rating for pins: V_{DD}
3. Absolute maximum rating for pins: RX, TST
4. Absolute maximum rating for pins: C2P, C2N, SWxy for x={4÷1} & y={3÷0}
5. Absolute maximum rating for pins: TX
6. Absolute maximum rating for pins: A, B
7. Absolute maximum rating for pins: TST1, ADC0/SDA, ADC1/SCL, ADC2/ADR
8. Absolute maximum rating for pins: SWx_(y+1) – SWxy for x={4÷1} & y={2÷0}
9. For limited time up to 100 hours. Otherwise the max storage temperature is 85°C.
10. This device series incorporates ESD protection and is qualified per AEC-Q100:
 ESD Human Body Model Classification level H1C in according to the AEC-Q100-002 Rev-E
 ESD Charge Device Model Classification C2b in according to the AEC-Q100-011 Rev-D
 Latch – up Current Maximum Rating: ≤ 100 mA in according to the AEC-Q100-004 Rev-D JEDEC-Class II

Operating ranges define the limits for functional operation and parametric characteristics of the device. A mission profile (Note 11) is a substantial part of the

operation conditions; hence the Customer must contact **onsemi** in order to mutually agree in writing on the allowed missions profile(s) in the application.

Table 4. RECOMMENDED OPERATING RANGES

Characteristic	Symbol	Min	Typ	Max	Unit
Battery supply voltage	V _{BB}	4.5		40	V
Switch differential voltage	V _{SW_DIFF}	0		10	V
LED string voltage	V _{STRING}	0		60	V
Buck switch output current	I _{SW}			1.4	A
PXN communication speed	S _{PXN}	125		1000	kbit
Ambient temperature	T _A	-40		125	°C
Junction temperature range (Note 12)	T _J	-40		150	°C

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

11. The circuit functionality is not guaranteed outside the Operating junction temperature range. A mission profile describes the application specific conditions such as, but not limited to, the cumulative operating conditions over life time, the system power dissipation, the system's environmental conditions, the thermal design of the customer's system, the modes, in which the device is operated by the customer, etc.
12. The circuit functionality is not guaranteed outside the junction temperature range. Also please note that the device is verified on bench for operation up to 170 °C but the production test guarantees 150 °C only.

Table 5. THERMAL RESISTANCE

Characteristic	Package	Symbol	Min	Typ	Max	Unit
Thermal resistance junction to exposed pad (Note 13)	SSOP36-EP	R _{thjp}		3.5		°C/W

13. Includes also typical solder thickness under the Exposed Pad (EP).

ELECTRICAL CHARACTERISTICS

NOTE: All Min and Max parameters are guaranteed over full junction temperature (T_{JP}) range (–40 °C; 150 °C), unless otherwise specified.

Table 6. CURRENT CONSUMPTION

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
The VBB current consumption	I_VBB			19	25	mA
The VBB current consumption UART only device	I_VBB_M-LVDS_ OFF	M-LVDS off; OTP bit M-LVDS_OFF = '1'		6.5	10	mA

Table 7. OSC20M: SYSTEM OSCILLATOR CLOCK

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
Oscillator output frequency (trimmed)	OSC_CLK		18.2	20	21.8	MHz
Oscillator duty cycle	OSC_DC		30	50	70	%

Table 8. VDD: 3.45V LOW VOLTAGE ANALOG AND DIGITAL SUPPLY

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
VDD regulator output voltage	VDD	VBB > 4.5 V	3.15	3.45	3.6	V
VDD regulator current limitation	VDD_ILIM	VBB > 4.5 V	40		300	mA
OUT_OFF_REG comparator voltage	V_OUT_OF_REG		2.7		3.45	V
VDD POR threshold, VDD rising	POR3V_H		2.7		2.95	V
VDD POR threshold, VDD falling	POR3V_L		2.5		2.75	V
VDD POR hysteresis	POR3V_HYST		0.1	0.2	0.3	V
VBB POR threshold, VBB rising	POR_VBB_H		3.8		4.3	V
VBB POR threshold, VBB falling	POR_VBB_L		3.7		4.2	V
VBB POR hysteresis	POR_VBB_HST		0.05	0.1	0.25	V
OTP UV comparator threshold (VBB pin)	OTP_UV		12.5		15	V
VBB supply during the OTP zapping	VBB_ZAP		15		30	V
VBB current limitation for OTP zapping	IBAT_ZAPP		85			mA

Table 9. SWITCH CONTROL

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
V(C2) under voltage threshold, V(C2) rising	CCH_UVH		2.65	2.75	2.85	V
V(C2) under voltage threshold, V(C2) falling	CCH_UVL		2.6	2.72	2.85	V
Current from VBB to charge C2 capacitor	CCH_IBB		2		15	mA
Current limitation from VDD (during start-up)	CCH_ILIM_RST		6	12	20	mA
Current limitation from VDD	CCH_ILIM		8	12	16	mA
Voltage drop between VDD and V(C2)	CCH_VDROP			120	270	mV
V(C2) voltage after recharge CCH_V2 = VDD – CCH_VDROP	CCH_V2			3.33		V
Switch OFF time	SOF_TRISE	5 mA, without decoupling capacitor	1.5	1.6	2.5	µs
Switch gate voltage detection threshold	SOF_VTH_A	At ambient temperature	0.4	0.8	1.6	V
Switch gate voltage detection threshold	SOF_VTH_C	At cold temperature	0.9	1.3	1.7	V
Switch gate voltage detection threshold	SOF_VTH_H	At hot temperature	0.2	0.8	1.2	V
Switch Short detection voltage threshold	SSH_VTH		0.35		1	V
Switch Overvoltage detection threshold	SOV_TH		10		13.5	V

Table 10. PIXEL SWITCHES

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
RON from SWx3 to SWx0 pin (3 switches)	SW_3R	At ambient		0.43	1.1	Ω
RON from SWxy to SWx(y-1) pin (1 switch)	SW_1R	At ambient		0.2	0.6	Ω
Current from SWxy pin to GND (see Figure 9)	SW_IGND		40	53	70	μA

Table 11. ADC FOR MEASURING VBB, VDD, VLED, TEMP, ADC_x

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
ADC Resolution	ADC_RES			8		Bits
Integral Non-linearity (INL)	ADC_INL	Best fitting straight line method	-1.5		+1.5	LSB
Differential Non-linearity (DNL)	ADC_DNL	Best fitting straight line method	-2.0		+2.0	LSB
Full path gain error	ADC_GE	VBB, VDD measurements	-3.25		3.25	%
Offset at output of ADC	ADC_OFFSET	VBB, VDD measurements	-2		2	LSB
Time for 1 SAR conversion	ADC_CONV		6.67	8	10	μs
ADC full scale for VBB measurement	ADC_VBB		33.5	35	36.5	V
ADC full scale for VDD measurement	ADC_VDD		3.87	4	4.13	V
ADC full scale for VLED measurement	ADC_VLED		63.6	66.1	68.6	V
ADC full scale for ADC _x measurement	ADC_ADC _x		1.175	1.205	1.235	V
ADC _x input current	I_ADC _x		0.3	1	1.7	μA
TSD threshold level	ADC_TSD	ADC measurement of junction temperature	163	170	177	$^{\circ}\text{C}$
Accuracy of temperature meas at hot	ADC_TEMP_ACC_HOT	T = 155 $^{\circ}\text{C}$	-7		7	$^{\circ}\text{C}$
Accuracy of temperature meas at cold	ADC_TEMP_ACC_COLD	T = -40 $^{\circ}\text{C}$	-15		15	$^{\circ}\text{C}$

Table 12. GND LOSS DETECTION

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
GND loss comparator threshold; both edges	GNDLOSS_THR		100	120	160	mV
GND loss comparator delay; both falling and rising edge	GNDLOSS_DEL			800	1200	ns

Table 13. UART INTERFACE: RX, TX

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
High-level input voltage	RX_VIH		2			V
Low-level input voltage	RX_VIL				0.8	V
Input voltage hysteresis	RX_Vlhyst		100	200	400	mV
Input pull-down resistance	RX_RPULL		40		160	k Ω
High-level output voltage	TX_VOH	I _{LOAD} = -3mA	2.1		VDD or external pull-up voltage	V
Low-level output voltage	TX_VOL	I _{LOAD} = 3mA			0.4	V
TX pin leakage current in HiZ	TX_ILEAK		-1		1	μA
TX pin capacitance	TX_C			5		pF
Propagation delay	TX_DL_50pF	C _{LOAD} up to 50 pF			40	ns
Propagation delay	TX_DL_200pF	C _{LOAD} up to 200 pF			150	ns

Table 14. M-LVDS INTERFACE: A, B

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
Differential output voltage magnitude	M-LVDS_TX_VAB	$R_{load_A-B} = 49.9\ \Omega \pm 1\%$ $V_{test} = \text{from } -1\text{ V to } 3.4\text{ V}$	480		650	mV
Change in Differential output voltage magnitude between logic states	M-LVDS_TX_DVAB	$R_{load_A-B} = 49.9\ \Omega \pm 1\%$ $V_{test} = \text{from } -1\text{ V to } 3.4\text{ V}$	-50		50	mV
Steady state common mode output voltage	M-LVDS_TX_VOS	$R_{load_A-B} = 49.9\ \Omega \pm 1\%$	1	1.2	1.4	V
Change in Steady state common mode output voltage between logic states	M-LVDS_TX_DVOS	$R_{load_A-B} = 49.9\ \Omega \pm 1\%$	-50		50	mV
Peak-to-peak common-mode output voltage	M-LVDS_TX_VOSPP	$R_{load_A-B} = 49.9\ \Omega \pm 1\%$			150	mV
Maximum steady-state open-circuit output voltage	M-LVDS_TX_VOC	$R_{load} \geq 1.62\text{ k}\Omega$	1.9		2.28	V
Short-circuit output current magnitude	M-LVDS_TX_IOS	$V_{test} = \text{from } -1\text{ V to } 3.4\text{ V}$			43	mA
Voltage overshoot, low-to-high level output	M-LVDS_TX_VPH	$V_{SS} = 2 \cdot V_{AB}$			1.2	VSS
Voltage overshoot, high-to-low level output	M-LVDS_TX_VPL	$V_{SS} = 2 \cdot V_{AB}$	-0.2			VSS
Differential Output rise and fall times	M-LVDS_TX_TE			5	12	ns
Transmitter Propagation delay	M-LVDS_TX_TP		5	10	20	ns
Positive-going Differential Input voltage Threshold for BUS common mode <0; 3.8> V	M-LVDS_RX_VITP				150	mV
Positive-going Differential Input voltage Threshold for BUS common mode <-1.4; 0> V	M-LVDS_RX_VITP_NCMM				160	mV
Negative-going Differential Input voltage Threshold for BUS common mode <0; 3.8> V	M-LVDS_RX_VITN		50			mV
Negative-going Differential Input voltage Threshold for BUS common mode <-1.4; 0> V	M-LVDS_RX_VITN_NCMM		60			mV
Receiver Propagation delay	M-LVDS_RX_TP		20	40	60	ns
A or B pin capacitance	M-LVDS_C			5		pF
Transceiver input current in high impedance state (range 1)	M-LVDS_IOZ_1	$0\text{ V} \leq (V_A \text{ or } V_B) \leq 2.4\text{ V}$, other output at 1.2 V, transmitter in HiZ	-20		20	μA
Transceiver input current in high impedance state (range 2)	M-LVDS_IOZ_2	$-1.4\text{ V} \leq (V_A \text{ or } V_B) \leq 0\text{ V}$ or $2.4\text{ V} \leq (V_A \text{ or } V_B) \leq 3.8\text{ V}$, other output at 1.2V, transmitter in HiZ	-32		32	μA

Table 15. I2C INTERFACE: SDA, SCL

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
High-level input voltage	I2C_VIH		0.7			VDD
Low-level input voltage	I2C_VIL				0.3	VDD
Input voltage hysteresis	I2C_Vlhyst		300		700	mV
Low-level output voltage	I2C_VOL				0.4	V
High-level output voltage	I2C_VOH	$I_{LOAD} = -3\text{ mA}$		VDD-0.1		V
SDA or SCL pin capacitance	I2C_C	$I_{LOAD} = 3\text{ mA}$		5		pF
SCL to SDA and SDA to SCL propagation delay, both edges	I2C_DL		5		60	ns

DETAILED OPERATING AND PIN DESCRIPTION

SUPPLY CONCEPT IN GENERAL

Low operating voltages become more and more required due to the growing use of start stop systems. In order to respond to this necessity, the NCV78343 is designed to support power-up starting from $V_{BB} = 4.5\text{ V}$.

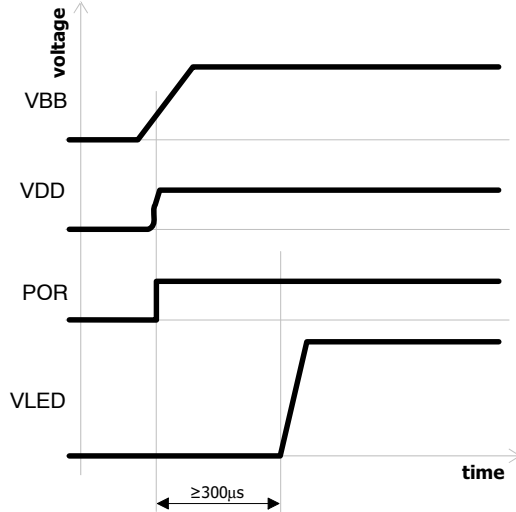


Figure 10. Power-up Sequence

A specific power-up and power-down sequences are shown in the Figure 20 and Figure 21.

There is no special circuit to disable switches in case of VBB power supply disconnection. The gate of the switch is discharged by SW-OFF circuit in case the VBB-LOW threshold is crossed. The gate of the switch is discharged by leakage currents when the supply is suddenly lost. Because of low leakage currents, the switch may stay enabled for a few seconds after power lost. Possible temperature rise speeds up opening the switch by higher leakage current.

VDD Supply

The VDD supply is the low voltage digital and analog supply for the chip, which is powered from VBB. VDD is supplying the internal analog and digital circuits as well as external components like I2C EEPROM and resistor divider on ADC inputs. The POR-circuit is monitoring both the VBB and VDD voltages.

VLED Supply

If the device is running but the LED current source is disconnected, the LEDs can light up because of the bias currents flowing through pins of the switches. Up to $180\mu\text{A}$ (typical) from switch current source may cause the bottom-most LED to shine. If needed, resistors can be connected in parallel to the switches to avoid undesired LED lighting (typically $10\text{ k}\Omega$).

INTERNAL CLOCK GENERATION

The clocks are fully internally generated without the need for any trimming by the user. The accuracy is guaranteed under all operating conditions and independent of external component selection.

OSC20M Clock

The OSC20M clock is the system clock. All the internal timings as well as the internal PWM unit depend on OSC20M accuracy.

Communication Clock

The internal clock is also used for oversampling of UART incoming frame and I2C EEPROM, so there is no need for any external clock.

DIMMING CONTROLLER

Internal (built-in) dimming controller allows change of light intensity of individual LEDs in LED string by means of digital (PWM) dimming.

Dimming Control Parameters

The dimming for all switches is controlled from 1 common 10-bit counter. The ON and OFF events are programmable per channel, each with a 10 bit counter value.

100% duty cycle is generated when ON time is set to min. value (0) and OFF time is set to max value (1023).

0% duty cycle is generated when ON time is equal to OFF time. When more than one 0% or 100% duty cycle is required, the TR (transition) slots must be used.

The dimming frequency is the DIMCLK frequency divided by 1024. The T_{DIMCLK} is the duration of one PWM tick. The duration of one PWM period is T_{PWM} . The required time for one switch ON sequence is $T_{\text{SW_SEQ}}$. The ratio of $T_{\text{SW_SEQ}}$ and T_{DIMCLK} results in number of PWM ticks required for one switch ON sequence. The number of slots available for each DIMCLK is 1024 divided by the ratio. The recommended time for TR slots and recommended step between each switch ON request is shown in Table 43. When the TR slot technique is used, the ON values should not be set within this period.

Dimming Mode

The NCV78343 incorporates two modes of operation – ON/OFF dimming mode and direct mode.

- **ON/OFF mode** – the NCV78343 controls the dimming duty cycle and phase shift for each switch individually. The time of ON event is set by means of $\langle \text{ONx}[9:0] \rangle$ register and the time of OFF event is set by means of $\langle \text{OFFx}[9:0] \rangle$ register.
- **Direct mode** – in addition to ON/OFF dimming mode, the state of the switches can be controlled directly by means of $\langle \text{SWx} \rangle$ register.

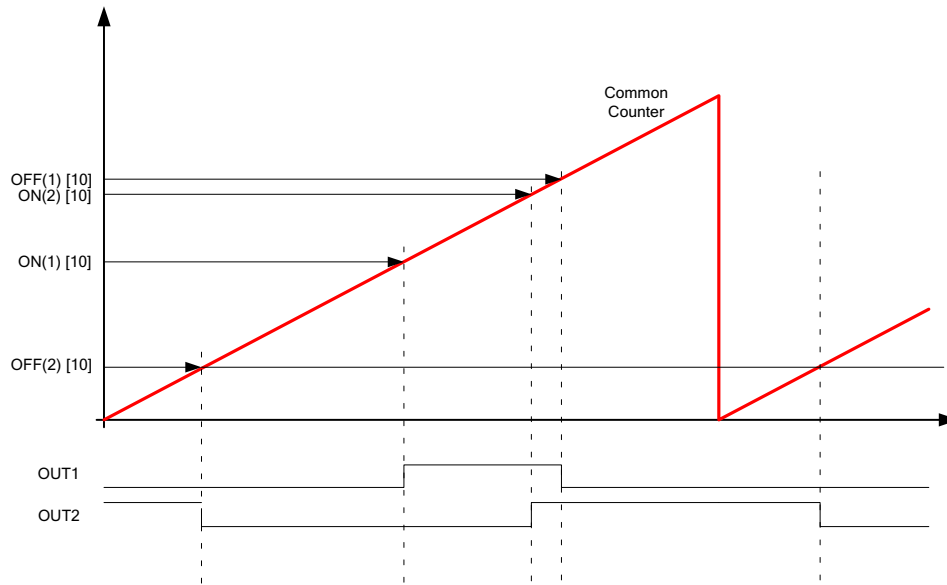


Figure 11. Dimming Operation (dimming ON/OFF event)

Dimming Transition Vector Insertion

Transition vectors are required in case of pattern changes (update of dimming settings) for avoiding multiple switching events at the same time and minimizing brightness error.

Fully closed switch (100% duty cycle) requires ON event equal to 0. It can happen that such switch ON event is required on more switches at the same time, which is not allowed. Therefore a transition slot technique is used for consecutive activation of those switches (which need to be changed to 100% duty cycle). When overlapping multiple switch ON events are invoked despite this, the <DIMERR> error is raised. When overlapping switch OFF events occur, the <DIMWARN> status bit is set and processing of this pattern continues. However, multiple switch OFF events may cause large LED string voltage changes.

Transition vector inserts additional transition either ON or OFF event at the beginning of next PWM period (in transition slots space). This helps to reduce brightness error significantly and the duty cycle is affected only in one period. The error is proportionate to duration of transition slot.

Pattern is updated when common PWM counter overflows and <MAPENA> = '1' (see Table 64) is set.

The NCV78343 contains 12 channels, so with unique settings of <TRx[3:0]> for each switch 12 different Transient Vector values are needed in the worst case ("0x0" to "0xB"). When <TRx[3:0]> = '0xC', '0xD', '0xE' or '0xF', the <TRx[3:0]> is ignored and transition vectors are not applied. In this case the switch status from previous PWM period is kept unchanged until next ON or OFF event into opposite direction.

PWM dimming clock

Selection of internal dimming clock is done by means of <DIMFREQ[4:0]> register, which shall be used to select dimming frequencies in range of 125 kHz to 1 MHz (see Table 43. PWM Frequency Settings).

SWITCH CONFIGURATIONS

The 12 integrated switches are typically organized as 12 × 1 switch of 1.4 A, but can be organized in 6 × 2 switches in parallel to offer 6 × 1 switch of 2.8 A. Examples of switch configurations are shown in Figure 12.

Selection of the switch configuration is done by <CONF_SEL[2:0]> register. Detailed information about switch configuration is available in Table 16.

Table 16. SWITCH CONFIGURATIONS

CONF_SEL [2:0]	Conf. Code Name	Description
000	1, 2, 3, 4	12 × PWM channels
001	1+2, 3, 4	9 × PWM channels (PWM 1=2)
010	1+2, 3+4	6 × PWM channels (PWM 1=2 & 3=4)
011	1, 2+3, 4	9 × PWM channels (PWM 2=3)
100	1, 2, 3+4	9 × PWM channels (PWM 3=4)
101	1+4, 2+3	6 × PWM channels (PWM 1=4 & 2=3)
110	1+4, 2, 3	9 × PWM channels (PWM 1=4)
111	1, 2, 3, 4.	Same as 0000, 12 × PWM channels

In case of configurations with 2× current, PWM signals of sections with higher index are controlled with PWM signals from lower index section. For example in case of

configuration “101”, the PWM signals of section 1 is controlling section 4; control signals of section 2 is controlling section 3.

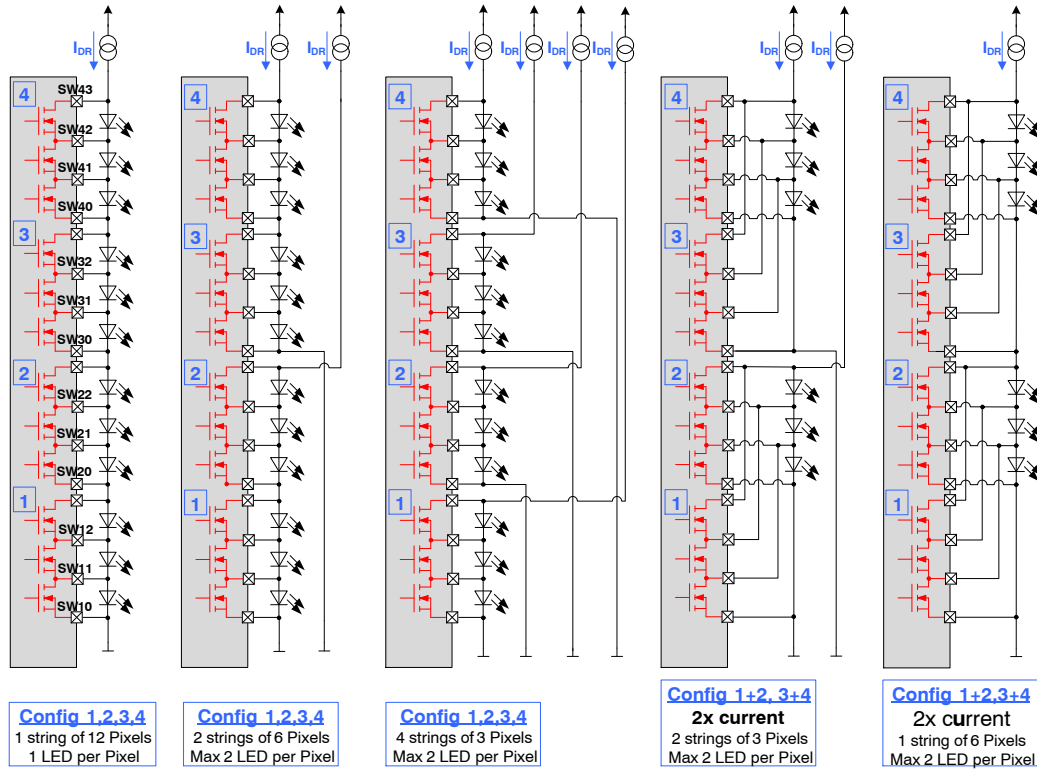


Figure 12. Example of Switch Configurations

Parallel combination is used where the I_{DR} current exceeds maximum switch current 1.4 A. This is not for use in redundant applications.

The following consequence must be taken into account when using parallel switches:

The OTP safe-state bits should be zapped to “0” to avoid sequentially switching ON which might cause that the higher current will flow through one switch.

Analog Input

The analog input AIN is an input channel that can be used for different types of measurements, like e.g. LED temperature or battery voltage. The converted voltage is calculated with the following formula:

$$V_{ADC_x} = ADC_RES_x \cdot \frac{1.205}{255} [V] \quad (\text{eq. 1})$$

where

ADC_RES_x is saved in register 0x11

OPEN, SHORT and FAIL Status Detection

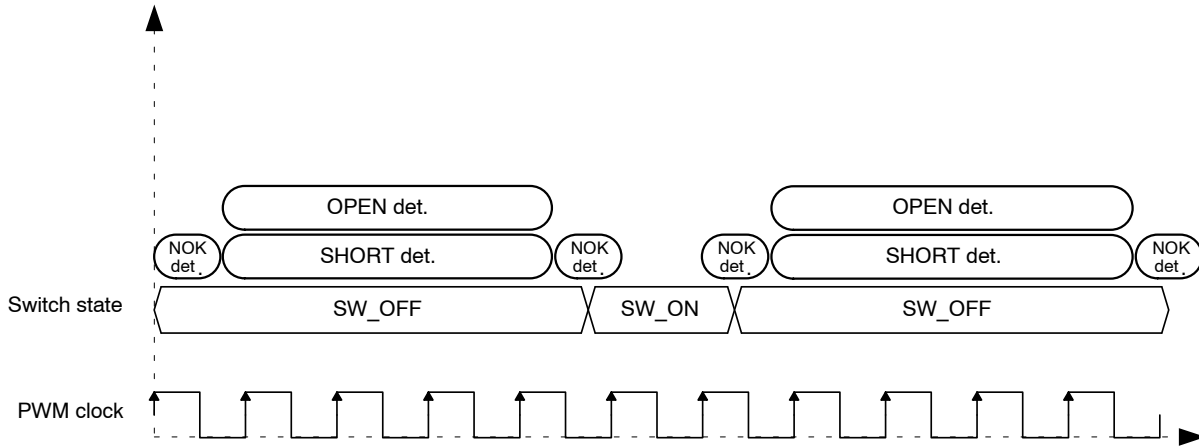
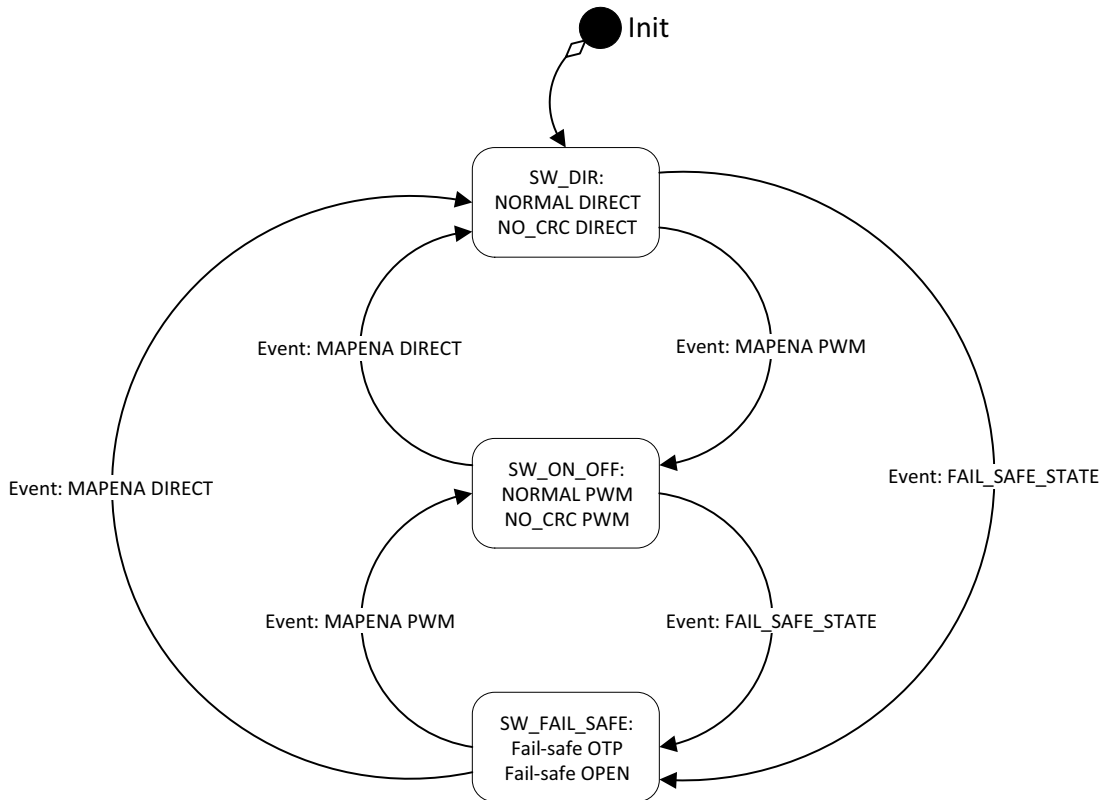


Figure 13. OPEN, SHORT and FAIL Status Detection Timing

Following the figure above, the OPEN and SHORT flags are detected only during the switch OFF state. The On/Off Failed flag detection is triggered by the transition between the switch ON and the switch OFF event. The SHORT and On/Off Failed status flags are cleared upon a successful read

out of register 0x0F. Due to this behavior and the diagram above, the read status might alternate between the SHORT and On/Off Failed, following the duty cycle of the specific switch. When the buck current is disabled, the device reports SHORT status for all switches.



NOTE: MAPENA DIRECT means writing into REG 0x00. MAPENA PWM means either writing into REG 0x0D or sending CF15.

Figure 14. Normal Mode State Machine

Pixel Light Network

The PXN is a proprietary network technology developed primarily for communication with and within the LED matrix head light system (see Figure 15).

The LED matrix head light system may incorporate a various number of sub-systems interconnected using PXN technology. The connection of such sub-system to a local network is realized using M-LVDS physical interface and a twisted-pair cable. Termination is required at both ends of the twisted-pair cable. Nominally, it is 100 Ω across the pair. Transmitter on the bus sees both termination resistors in parallel, thus the nominal bus load is actually 50 Ω .

The LED matrix head light system can be integrated into a superior system through an optional physical interface, e.g. differential low speed CAN. The choice of the external physical interface is application specific.

The PXN protocol for communication over PXN is based on UART communication standard, i.e. one start bit, 8 data bits (LSB first), one stop bit, no parity bit.

Rx pin is 5 V tolerant and has CMOS compatible threshold levels. External pull-up resistor is required on Tx pin.

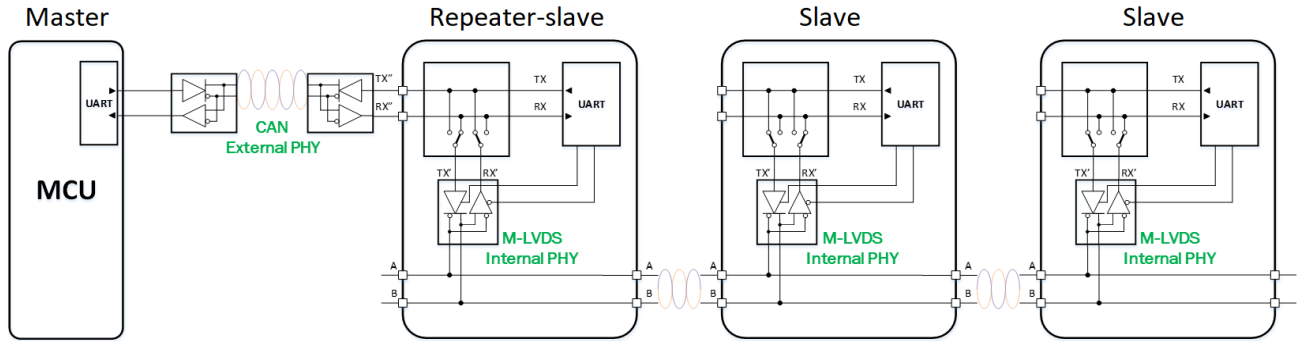


Figure 15. PXN Topology Inside LED Matrix Head Light System

Table 17. THE UART SIGNAL LEVELS TRANSFERRED TO M-LVDS BUS

UART RX Input Pin	M-LVDS Differential Voltage A-B	UART TX Output Pin
LOW	POSITIVE (A-B > 150 mV)	LOW
HIGH	NEGATIVE (A-B < 50 mV; in M-LVDS push-pull mode; valid for repeater-slave)	HIGH

The table above must be taken into account when using only M-LVDS slaves cluster. The master MCU generates UART signal, which is connected to the M-LVDS transceiver, where the A and B pins are connected to the A and B pins on the devices. Since the M-LVDS signal is inverted to the UART signal, there must be placed an inverter on the Tx pin from the MCU to M-LVDS transceiver and another inverter on the Rx pin from the M-LVDS transceiver to the MCU.

PXN Switch

The PXN switch is responsible for PXN frame routing within a particular PXN node connected to network.

PXN Media Access Layer

The MAC layer is responsible for a PXN frame composition on a transmitting side, the PXN frame decomposition on a receiving side, a transmission of composed PXN frames, a reception of PXN frames and PXN network error detection and confinement.

PXN Frame

A message is transferred over PXN bus in a form of PXN frame, which is depicted in Figure 16. The PXN protocol for communication over PXN is based on UART communication standard, i.e. one start bit, 8 data bits (LSB first), one stop bit, no parity bit.

The PXN frame consists of a header and a response. The header is always transmitted by PXN master while the response can either be transmitted by master, in case of write frames or by slave, in case of read frames. The header and the response are separated by in-frame response space.

The header consists of a BREAK field (logic 0 for a certain time), a SYNC field (0x55 byte) and two protected identifiers PID1 and PID2. The response consists of an arbitrary number of DATA bytes within a range from 1 to 12 followed by CRC. The particular bytes are separated by inter-byte space. Minimal delay of 1 Tbit is required before starting new PXN frame. The minimum length for the BREAK field is 13 Tbits (52 μ s for the default

communication speed 250 kbps). The BREAK field stop bit (BREAK field delimiter) is minimum 1 Tbit and maximum according to the selected watchdog time. If the device is not responding through the repeater-slave, the extended break (26 Tbits) can be required to recover communication to slave devices. Such case can occur when Read frame is addressed non assigned address. In case of only M-LVDS

slave cluster, the DE pin on the M-LVDS transceiver (e.g. NBA3N206S) must be set LOW within 1 Tbit after the Header part to allow device response.

The PXN protocol supports two frame types:

- configuration frame
- register bank frame

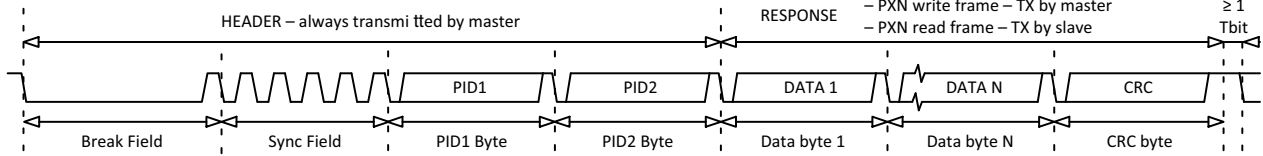


Figure 16. PXN Frame

PXN Configuration Frame

The configuration PXN frame allows activation and monitoring of selected configuration service.

Table 18. PXN CONFIGURATION FRAME

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0]				
2..13	DATAx	DATA[7:0] 0 ..11							
3	CRC	CRC[7:0]							

PID1:

P odd parity bit
SA [4:0] 5-bit slave node address

PID2:

P odd parity bit
CSID[4:0] 5-bit configuration service identifier
DATAx[7:0] 8-bit data, from 1 up to 12 data bytes supported
CRC[7:0] 8-bit CRC

PXN Register Bank Frame

The register bank PXN frame provides an access, both read or write to selected register(s) of internal register bank.

Table 19. PXN REGISTER BANK FRAME

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	FT[1:0]		SA[4:0]				
1	PID2	P	BC[1:0]		RBA[4:0]				
2..13	DATAx	DATA[7:0] 0 ..11							
3	CRC	CRC[7:0]							

PID1:

P odd parity bit

FT [1:0]

2-bit frame type:

“00” – read frame
“01” – write frame to address node only
“10” – write frame to all nodes (broadcast)

SA [4:0]

5-bit slave node address

PID2:

P odd parity bit
BC[1:0] 2-bit byte count
RBA[4:0] 5-bit register bank address
DATAx[7:0] 8-bit data, from 1 up to 12 data bytes supported
CRC[7:0] 8-bit CRC

PXN Register Bank Frame Matched by Length

The PXN network supports devices with different logical organization of internal register bank. The following logical organizations of register bank are supported:

- TYPE1 – up to 32x24 bits
- TYPE2 – up to 32x16 bits
- TYPE3 – up to 32x8 bits

Each of types above has predefined number of data bytes for given PID2.BC parameter in case PID1.FT=“10” (broadcast frame).

Table 20. BROADCAST PXN FRAME DATA BYTE COUNT

PID2.BC[1:0]	Data Byte Count		
	TYPE1	TYPE2	TYPE3
0x0	3	2	1
0x1	6	4	5
0x2	9	8	7
0x3	12	10	11

The NCV78343 supports only the TYPE1 register bank organization, since each register bank consists of 3 bytes.

This means that it is possible to read/write up to 4 registers in one frame, which can be for example used to write ON/OFF times for all 12 switches in only 3 PXN frames.

PXN Error Detection

The PXN network supports detection of these errors:

- frame error
- timeout error
- synchronization error
- local communication error
- global communication error

PXN Application Layer

List of supported configuration services:

Table 21. CONFIGURATION SERVICES

Configuration Service		Configuration Frame			
Name	Code	Name	CSID	Type	Description
Identification	0	CF0	0x00	R	Slave Identification
Ext. EEPROM	1	CF1	0x01	W	Write data to ext. I2C EEPROM
		CF2	0x02	W	Request data from ext. I2C EEPROM
		CF3	0x03	R	Read Data from ext. I2C EEPROM
Auto-addressing	2	CF4	0x04	W	Enable/disable auto-addressing mode
		CF5	0x05	W	Assign address
		CF6	0x06	R	OP mode status
PXN mode	3	CF7	0x07	W	Slave/repeater-slave PXN mode selection
		CF8	0x08	R	Read PXN mode status
OTP	4	CF9	0x09	W	Write data to OTP
		CF10	0x0A	W	Request data from OTP
		CF11	0x0B	R	Read data from OTP
UART	5	CF12	0x0C	W	Set UART communication speed
System	6	CF13	0x0D	W	Switch to normal mode
		CF14	0x0E	W	Reset system
		CF15	0x0F	W	Trigger MAPENA and/or CNTRST

List of supported register bank access:

Table 22. REGISTER BANK ACCESS

Configuration Service		Configuration Frame		
Name	Code	Name	Access Type	Description
Read/Write	0	WF1	Read	Write register bank
		RF1	Write	Read register bank

PXN Communication Modes

The PXN node can operate in one of the two communication modes:

- slave mode
- repeater-slave mode

Depending on the mode selected, the PXN switch is configured to route the PXN frames the respective way. The repeater-slave device works as a bridge between UART bus

and M-LVDS bus. It forwards frames from UART to M-LVDS and back from M-LVDS to UART when reading from a slave device.

Addressing Options

It is possible to set a device address in 3 different ways:

- Multi-level address pin
- Auto-addressing procedure
- OTP node address bits

Addressing using OTP memory is recommended for final application. Some of the other device parameters are saved in memory as well, which speeds up device setup after each power-on.

Multi-level Address Pin

The PXN node address can be determined by connecting ADC2/ADR input to a voltage divider. The voltage divider, represented by resistors R1 and R2 are supplied from regulated 3.3 V VDD supply. The voltage space is divided into 10 ranges where only 8 of them are associated with valid address. The corresponding thresholds are calculated as follows:

Table 23. MULTI-LEVEL ADDRESS PIN

PXN	Resistor Divider		ADC2/VDD	
Addr	R1 (kΩ)	R2 (kΩ)	min. (–)	max. (–)
7	91	27	0.75	0.79
6	68	15	0.59	0.63
5	91	15	0.46	0.49
4	82	10	0.35	0.38
3	91	8.2	0.27	0.29
2	51	3.3	0.20	0.21
1	82	3.6	0.14	0.15
0	51	1.3	0.08	0.09

In case of valid address the node can process both the addressed and the broadcast frames. In case of invalid address the node can process the broadcast frames only.

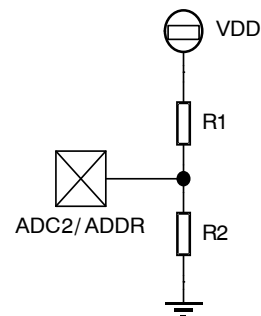


Figure 17. Voltage divided connected to ADC2/ADR Pin

The Multi-level addressing procedure requires stable voltage level at ADC2/ADR pin in 200 μs after POR. If the application cannot ensure this time, please follow the

Multi-level addressing procedure with long time delay recommendation in Application notes. If the Multi-level addressing is successful, a device stays in the OTP config mode (see Figure 23).

Auto-addressing Procedure

Regardless the node address assigned after the measurement of ADC2/ADR multi-level input, the auto-addressing procedure can still be invoked using the auto-address enable PXN frame. The auto-addressing is enabled/disabled on the PXN node upon receiving CF4 PXN configuration frame. The CF4 frame is accepted in the OTP_CONFIG mode only (see Table 53).

The PXN node address, when auto-addressing is enabled is assigned upon receiving CF5 PXN configuration frame. The CF5 frame is accepted in AUTO_ADDR mode only (see Table 54).

The auto-addressing procedure is described in the APPLICATION RELATED INFORMATION section.

OTP Node Address

The OTP node address can be zapped by customer at EoL (End of Line) after the PXN node address was determined either by means of multi-level address pin measurement or by means of auto-addressing procedure. The value of OTP node address and OTP bank lock bit is obtained each time the PXN node is powered up and the custom OTP bank is read out. Loading of other device settings from OTP memory speeds up device setup after power-on. OTP memory zapping is necessary to fulfil ASIL B safety requirements.

PXN Communication Speed

The PXN node can communicate at following speed:

- 125 kb/s
- 250 kb/s (default)
- 500 kb/s
- 1 Mb/s

Communication speed is changed upon receiving CF12 PXN communication frame in OTP_CONFIG, AUTO_ADDR and NORMAL modes only (see Table 61).

OTP Bank – Custom Data

The custom OTP bank is typically zapped by customer at EOL and stored values are used for system operation customization.

Table 24. OTP BANK

OTP#	OTP Name
0	OTP lock bit
1	OTP node address lock bit
2	OTP node address bit 0
3	OTP node address bit 1
4	OTP node address bit 2
5	OTP node address bit 3
6	OTP node address bit 4
7	Fail safe state lock bit

8	Fail safe state of LEDs in string 1
9	Fail safe state of LEDs in string 2
10	Fail safe state of LEDs in string 3
11	Fail safe state of LEDs in string 4
12	PXN lock bit
13	Mode (slave/repeater-slave)
14	Communication speed bit 0
15	Communication speed bit 1
16	Global bit error detection disable
17	M-LVDS OFF
18	UART OFF
19	EEPROM lock bit (write protect)
20	CRC bit0
21	CRC bit1
22	CRC bit2
23	CRC bit3
24	CRC bit4
25	CRC bit5
26	CRC bit6

<OTP_LOCK_BIT> – custom OTP bank general lock bit. When zapped, any further zapping attempt of custom OTP bank is declined.

<OTP_NODE_ADDR_LOCK_BIT> – PXN node address lock bit. When zapped, any further zapping attempt of <OTP_NODE_ADDR> bits of custom OTP bank is declined.

<OTP_NODE_ADDR [4:0]> – 5-bit PXN node address. This address is taken into account only when the <OTP_NODE_ADDR_LOCK_BIT> is zapped.

<FAIL_SAFE_STATE_LOCK_BIT> – fail safe state of LED string lock bit. When zapped, any further zapping attempt of <FAIL_SAFE_STATE_LED_STRINGx> bits of custom OTP bank is declined.

<FAIL_SAFE_STATE_LED_STRINGx> – state of the LED string x, x={1,2,3,4}, in case one of the following conditions is detected:

- NORMAL mode is entered or
- Timeout error occurred

The bits set directly the switch state. The fail safe state is taken into account only when

<FAIL_SAFE_STATE_LOCK_BIT> is zapped.

<PXN_LOCK_BIT> – PXN settings lock bit. When zapped, any further zapping attempt of PXN settings related bits of custom OTP bank is declined.

<PXN_MODE> – PXN communication mode selection bit. Set '0' for slave mode and '1' for repeater-slave mode. The mode selection is taken into account only when the <PXN_LOCK_BIT> is set and the CRC is correct.

<PXN_COMMUNICATION_SPEED [1:0]> – 2-bit PXN communication speed selection bit. The communication speed selection is taken into account only when the **<PXN_LOCK_BIT>** is set and the CRC is correct.

Table 25. PXN COMMUNICATION SPEED

COMMUNICATION_SPEED[1:0] OTP Setting	PXN Communication Speed [kb/s]
0x0	125
0x1	250
0x2	500
0x3	1000

The default communication speed, when the **<PXN_LOCK_BIT>** is not zapped is 250 kbps.

<GLOBAL_BIT_ERR_DTC_DIS> – global bit error detection disable. In case the global bit error detection is enabled (**<GLOBAL_BIT_ERR_DTC_DIS>** is not zapped) the data transmitted on chip's TX output must be echoed back into chip's RX input. There are two ways to generate the TX echo. Either it can be ensured by a CAN transceiver connected between a chip and an MCU on UART bus or it can be generated by the MCU itself. When the echo is not present, chip stops transmitting and will wait for new incoming frame. When more than one device is present on the UART bus, the repeater-slave will always report **<PXN_GLOBAL_COMM_ERR>** bit when reading from another device on the UART bus. In case the global bit error detection is disabled, the reading from an address not assigned to a physical device will block the further operation of the repeater-slave device. Such device must be unplugged from a battery voltage in order to make it operational again. The global bit error detection is enabled by default.

The device configured to act as a PXN repeater-slave shall always have the global bit error detection enabled, i.e. **<GLOBAL_BIT_ERR_DTC_DIS>** shall be not zapped. The setting of **<GLOBAL_BIT_ERR_DTC_DIS>** bit does not affect the communication over M-LVDS bus. The **<GLOBAL_BIT_ERR_DTC_DIS>** bit is taken into account only when the **<PXN_LOCK_BIT>** is set and the CRC is correct.

<UART_OFF> – UART interface disable. The **<UART_OFF>** selection is taken into account only when the **<PXN_LOCK_BIT>** is set and the CRC is correct.

<M-LVDS_OFF> – M-LVDS interface disable. The **<M-LVDS_OFF>** selection is taken into account only when the **<PXN_LOCK_BIT>** is set and the CRC is correct. Unused M-LVDS transceiver can be disabled to reduce current consumption by 12.5 mA (typical).

<EEPROM_LOCK_BIT> – external EEPROM lock bit. The EEPROM lock bit acts as an EEPROM write protection. When zapped, any further write attempts to external EEPROM is declined.

OTP Memory Zapping

The OTP zapping process is one-time programming process during which the OTP memory is written. This process cannot be undone. The OTP zapping is possible only in the OTP config mode. To ensure correct OTP zapping, the VBB voltage must be in range of 16 to 30 V with the current capability at least 85 mA during the OTP zapping process. The OTP memory zapping should be done at the EoL to fulfil the ASIL B requirements.

External MCU can read content of OTP memory. To do this, device must first receive CF10 PXN configuration frame followed by CF11 PXN configuration frame. This process is similar to reading from external I2C EEPROM.

Operating Modes

The PXN node can operate in following modes:

- OTP Config mode
- Auto-addressing mode
- Normal Direct mode
- Normal PWM mode
- Normal Fail-safe OTP mode
- Normal Fail-safe OPEN mode
- NO_CRC Direct mode
- NO_CRC PWM mode
- Fail-safe OTP mode
- Fail-safe OPEN mode

OTP Config mode – the chip enters this mode under the following circumstances: when the OTPs are not zapped and the voltage divider at ADR pin is in a valid range, or the OTPs are zapped but the OTP CRC BANK2 is wrong, or after successful auto-addressing process. Please see the following flow diagram 'Flow chart after POR' in the Application notes.

Chip with non-zapped OTP memory starts with both UART and M-LVDS interfaces enabled. To determine which one will be used, there is a 60 ms timer (typical) that starts once device enters OTP Config mode. After timer elapses, chip reads state of UART RX pin to determine if UART bus should remain enabled (RX pulled high) or disabled (RX pulled low). During this period, M-LVDS devices might be unable to communicate if their UART RX pin is pulled low. Timer can be stopped before elapsing by leaving OTP Config mode, typically by receiving CF13.

Auto-addressing mode – when the chip receives CF4 (see Table 53) configuration frame.

Normal Direct/PWM mode – this is the normal working mode, the chip enters this mode after the POR when the OTPs are zapped and the CRC is correct. Direct means that the switches are controlled directly by writing to the register 0x00. The PWM means that the switches are controlled by the PWM by writing ON and OFF values and sending CF15 (see Table 64).

NO_CRC Direct/PWM mode – the device enters the NO_CRC mode after receiving CF13 (see Table 62) in the OTP Config mode. The functionality of NO_CRC Direct

and PWM modes is same as Normal Direct and Normal PWM modes. NO_CRC prefix means that the device detected invalid CRC in OTP memory bank 2 during power on, most likely because OTP memory is not written. Please note that only device with written OTP memory achieves ASIL B safety rating.

Fail-safe OTP mode – when this fail-safe state is entered after watchdog timeout, OTP fail-safe data are loaded and applied on switches. Chip also enters this mode after DIMERR. To leave this mode, clear TIMEOUT or DIMERR flags set in register 0x10.

Fail-safe OPEN mode – this is the fail-safe state when the switches are automatically open under the following circumstances: TSD or CAP_UV or VBB_LOW appears. To leave this mode, please read out the register 0x10.

EEPROM

The external I2C EEPROM can be connected to SDO and SCL pins and supplied from the VDD net.

The PXN node writes data to EEPROM upon receiving CF1 PXN configuration frame (see Table 50). The PXN

node reads data from EEPROM upon receiving CF2 PXN configuration frame (see Table 51). The EEPROM data are stored in the internal buffer.

The PXN node provides the data, previously read from the EEPROM and stored in the internal buffer upon receiving CF3 configuration frame (see table 52). The EEPROM address is valid for “1010” + <EESA> .

The write and read operations are shown in the application notes chapter (EEPROM write and read operations).

Cyclic Redundancy Check

All PXN frames are covered by an 8-bit CRC, where the polynomial is 0x83 (Koopman’s notation; $x^8+x^2+x^1+1$) with 0xFF seed. The input bytes are bit swapped and LSB first. The CRC is calculated over PID1, PID2 and DATAx bytes.

The OTP bank 2 is covered by a 7-bit CRC, where the polynomial is 0x5B (Koopman’s notation; $x^7+x^5+x^4+x^2+x^1+1$) with 0x7F seed. The input is MSB first. The CRC is calculated over bits 0–19.

The code examples for both CRCs are shown in the APPLICATION RELATED INFORMATION chapter.

Table 26. PXN REGISTER BANK ADDRESS MAP

Address	Access	BIT																									
		23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
		BYTE2								BYTE1								BYTE0									
		7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0		
0x0	R/W	SERVICE=0x0	-	-	-	-	-	-	-	-	-	-	-	SW12	SW11	SW10	SW9	SW8	SW7	SW6	SW5	SW4	SW3	SW2	SW1		
		SERVICE=0x1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	SW_SEL					
		SERVICE=0x2	RESERVED_1																								
		SERVICE=0x3	RESERVED_2																								
0x1	R/W	ON1										OFF1										TR1					
0x2	R/W	ON2										OFF2										TR2					
0x3	R/W	ON3										OFF3										TR3					
0x4	R/W	ON4										OFF4										TR4					
0x5	R/W	ON5										OFF5										TR5					
0x6	R/W	ON6										OFF6										TR6					
0x7	R/W	ON7										OFF7										TR7					
0x8	R/W	ON8										OFF8										TR8					
0x9	R/W	ON9										OFF9										TR9					
0xA	R/W	ON10										OFF10										TR10					
0xB	R/W	ON11										OFF11										TR11					
0xC	R/W	ON12										OFF12										TR12					
0xD	R/W	-	-	-	-	-	-	-	-	TW_CODE[7:0]								-	-	-	-	ADC_SEL[1:0]		CRC_CLR	MAP_ENA		
0xE	R/W	-	-	-	-	-	-	-	-	-	-	-	T1_CONF	-	-	-	-	DIMFREQ[4:0]				CONF_SEL[2:0]					
0xF	R	SW12.STATUS [1:0]		SW11.STATUS [1:0]		SW10.STATUS [1:0]		SW9.STATUS [1:0]		SW8.STATUS [1:0]		SW7.STATUS [1:0]		SW6.STATUS [1:0]		SW5.STATUS [1:0]		SW4.STATUS [1:0]		SW3.STATUS [1:0]		SW2.STATUS [1:0]		SW1.STATUS [1:0]			
0x10	R	PXN_CRC_ERR_CNT[3:0]				OTP_CRC_FAIL_BANK 0	OTP_CRC_FAIL_BANK 2	TIME_OUT	PXN_SYNC_ERR	PXN_FRAME_ERR	PXN_LOCAL_COMM_ERR	PXN_GLOBAL_COMM_ERR	MAP_ENA_STATUS	PWM_CNT_OVF	GND_LOSS	VBB_LOW	OTP_ZAP_UV	CAP_UV	HWR	0	DIM_ERR	DIM_WARN	GSW_ERR	TSD	TW		
0x11	R	ADCX_RES[7:0]								VDD_RES[7:0]								TEMP_RES[7:0]									
0x12	R	TSD_CODE[7:0]								VLED_RES[7:0]								VBB_RES[7:0]									

NOTE: Default value of all registers after POR is 0x00 if not specified explicitly.

REGISTER DESCRIPTION

Table 27. REGISTER 0x00

Register 0x00													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x00	Name	SW12	SW11	SW10	SW9	SW8	SW7	SW6	SW5	SW4	SW3	SW2	SW1
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x00	Name	SERVICE[23:22]		–	–	–	–	–	–	–	–	–	–
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

SERVICE[1:0] – specify the behavior of the last [21:0] bits:

“00” – Direct control – Direct control of all switches all together

“01” – OPEN clear request – De-activation of selected switch in on state due to OPEN fault previously detected

“10” – Reserved

“11” – Reserved

SW[11:0] – Direct control of the switches ON/OFF or OPEN clear request:

SERVICE = “00”: direct control of the switches ON/OFF; valid SW[11:0], others are “0”.

SERVICE = “01”: OPEN clear request; Switch is chosen by valid SW_SEL[3:0] from “0001” to “1100” (Switch 1 to 12), others are “0”.

Table 28. REGISTER 0x01

Register 0x01													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x01	Name	OFF1[11:4]									TR1[3:0]		
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x01	Name	ON1[23:14]										OFF1[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON1[9:0] – 10-bit switch ON threshold.

OFF1[9:0] – 10-bit switch OFF threshold.

TR1[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 29. REGISTER 0x02

Register 0x02													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x02	Name	OFF2[11:4]									TR2[3:0]		
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x02	Name	ON2[23:14]										OFF2[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON2[9:0] – 10-bit switch ON threshold.

OFF2[9:0] – 10-bit switch OFF threshold.

TR2[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 30. REGISTER 0x03

Register 0x03													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x03	Name	OFF3[11:4]								TR3[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x03	Name	ON3[23:14]										OFF3[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON3[9:0] – 10-bit switch ON threshold.

OFF3[9:0] – 10-bit switch OFF threshold.

TR3[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 31. REGISTER 0x04

Register 0x04													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x04	Name	OFF4[11:4]								TR4[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x04	Name	ON4[23:14]										OFF4[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON4[9:0] – 10-bit switch ON threshold.

OFF4[9:0] – 10-bit switch OFF threshold.

TR4[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 32. REGISTER 0x05

Register 0x05													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x05	Name	OFF5[11:4]								TR5[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x05	Name	ON5[23:14]										OFF5[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON5[9:0] – 10-bit switch ON threshold.

OFF5[9:0] – 10-bit switch OFF threshold.

TR5[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 33. REGISTER 0x06

Register 0x06													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x06	Name	OFF6[11:4]								TR6[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x06	Name	ON6[23:14]										OFF6[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON6[9:0] – 10-bit switch ON threshold.

OFF6[9:0] – 10-bit switch OFF threshold.

TR6[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 34. REGISTER 0x07

Register 0x07													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x07	Name	OFF7[11:4]								TR7[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x07	Name	ON7[23:14]										OFF7[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON7[9:0] – 10-bit switch ON threshold.

OFF7[9:0] – 10-bit switch OFF threshold.

TR7[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 35. REGISTER 0x08

Register 0x08													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x08	Name	OFF8[11:4]								TR8[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x08	Name	ON8[23:14]										OFF8[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON8[9:0] – 10-bit switch ON threshold.

OFF8[9:0] – 10-bit switch OFF threshold.

TR8[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 36. REGISTER 0x09

Register 0x09													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x09	Name	OFF9[11:4]								TR9[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x09	Name	ON9[23:14]										OFF9[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON9[9:0] – 10-bit switch ON threshold.

OFF9[9:0] – 10-bit switch OFF threshold.

TR9[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 37. REGISTER 0x0A

Register 0x0A													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0A	Name	OFF10[11:4]								TR10[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x0A	Name	ON10[23:14]										OFF10[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON10[9:0] – 10-bit switch ON threshold.

OFF10[9:0] – 10-bit switch OFF threshold.

TR10[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 38. REGISTER 0x0B

Register 0x0B													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0B	Name	OFF11[11:4]								TR11[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x0B	Name	ON11[23:14]										OFF11[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON11[9:0] – 10-bit switch ON threshold.

OFF11[9:0] – 10-bit switch OFF threshold.

TR11[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 39. REGISTER 0x0C

Register 0x0C													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0C	Name	OFF12[11:4]								TR12[3:0]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x0C	Name	ON12[23:14]										OFF12[13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

ON12[9:0] – 10-bit switch ON threshold.

OFF12[9:0] – 10-bit switch OFF threshold.

TR12[3:0] – Transition vector duration, it is prolonging the duration of ON resp. OFF value at the end of PWM period by $\langle \text{TRx}[3:0] \rangle \times \text{Time slot between switch activations}$.

Table 40. REGISTER 0x0D

Register 0x0D													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0D	Name	TW_CODE[11:8]				–	–	–	–	ADC_SEL[3:2]		CRC_CLR	MAPENA
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x0D	Name	–	–	–	–	–	–	–	–	TW_CODE[15:12]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

MAPENA – Register bank map enable request. When MAPENA request is written ‘1’ the internal mapena register bit is set. It remains set until PWM counter overflows. The internal mapena is cleared upon the PWM counter overflows. The $\langle \text{MAPENA} \rangle$ is always read as ‘0’.

CRC_CLR – $\langle \text{PXN_CRC_ERR_CNT}[3:0] \rangle$ clear request. When $\langle \text{CRC_CLR} \rangle$ bit is set to ‘1’ the $\langle \text{PXN_CRC_ERR_CNT}[3:0] \rangle$ bits are cleared immediately. The $\langle \text{CRC_CLR} \rangle$ bit is always read as ‘0’.

ADC_SEL[1:0] – 2-bit ADC measurement channel selection for ADCx A/D conversion (see Table 11. ADC for measuring VBB, VDD, VLED, TEMP, ADCx). The measurement channel is selected according to the following table:

Table 41. ADC MEASUREMENT CHANNEL SELECTION

ADC_SEL[1:0]	Measurement Channel
0x0	ADC0
0x1	ADC1
0x2	ADC2
0x3	Reserved

NOTE: The ADCx measurement result can be obtained by reading $\langle \text{ADCx_RES}[7:0] \rangle$ status bits.
In case the $\langle \text{ADC_SEL}[1:0] \rangle$ bits are set to “11”, the returned measured value is always “00000000”.

TW_CODE – 8-bit thermal warning threshold. The default value is calculated as follows:

$$\text{TW_CODE}[7:0] = \text{TSD_CODE}[7:0] - 9$$

Table 42. REGISTER 0x0E

Register 0x0E													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0E	Name	–	–	–	–	DIMFREQ[7:3]					CONF_SEL[2:0]		
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x0E	Name	–	–	–	–	–	–	–	–	–	–	–	T1_CONF
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW	RW

CONF_SEL[2:0] – Selects the switch configuration. NCV78343 supports the switch configurations listed in Table 16.

DIMFREQ[4:0] – Defines the DIMCLK frequency register, which shall be used to select dimming frequencies in range of 125 kHz to 1 MHz.

Table 43. PWM FREQUENCY SETTING

DIMFREQ [4:0]	f _{DIMCLK} [kHz]	T _{DIMCLK} [μs]	f _{PWM} [Hz]	T _{PWM} [ms]	T _{SW_SEQ} [μs]	TR	SW SLOT	T _{SW_SEQ} /T _{DIMCLK}
0	125.00	8.00	122.07	8.19	16	24	41	2
1	133.33	7.50	130.21	7.68	15	24	41	2
2	142.86	7.00	139.51	7.17	14	24	41	2
3	152.85	6.50	149.27	6.70	13	24	41	2
4	166.67	6.00	162.76	6.14	12	24	41	2
5	181.82	5.50	177.56	5.63	11	24	41	2
6	200.00	5.00	195.31	5.12	10	24	41	2
7	222.22	4.50	217.01	4.61	9	24	41	2
8	250.00	4.00	244.14	4.10	16	48	20	4
9	266.67	3.75	260.42	3.84	15	48	20	4
10	285.71	3.50	279.01	3.58	14	48	20	4
11	307.69	3.25	300.48	3.33	13	48	20	4
12	333.33	3.00	325.52	3.07	12	48	20	4
13	363.64	2.75	355.12	2.82	11	48	20	4
14	400.00	2.50	390.63	2.56	10	48	20	4
15	444.44	2.25	434.02	2.30	9	48	20	4
16	500.00	2.00	488.28	2.05	16	96	9	8
17	571.43	1.75	558.04	1.79	14	96	9	8
18	666.67	1.50	651.04	1.54	12	96	9	8
19	800.00	1.25	781.25	1.28	10	96	9	8
20	1000.00	1.00	976.56	1.02	16	192	4	16
21 .. 31	125.00	8.00	122.07	8.19	16	24	41	2

T_{DIMCLK} – the duration of one dimming clock tick.

T_{PWM} – the duration of 1024 dimming ticks.

T_{SW_SEQ} – the duration of one switch ON event.

T_{SW_SEQ}/T_{DIMCLK} – the number of clocks required for one switch ON event.

TR – the number of ticks for all transient vectors. This space should be reserved for TR when this technique is used. Calculated as

$$(T_{SW_SEQ}/T_{DIMCLK}) \cdot 12.$$

SW SLOT – recommended distance (in ticks of dimming clock) between two switch ON events.

T1_CONF – The bit defines the switch ON time (switching slope), where the “1” means steeper slope. For better EMC results it is recommended to set this value to “0”.

Table 44. REGISTER 0x0F

Register 0x0F													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0F	Name	SW6.STATUS [11:10]		SW5.STATUS [9:8]		SW4.STATUS [7:6]		SW3.STATUS [5:4]		SW2.STATUS [3:2]		SW1.STATUS [1:0]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x0F	Name	SW12.STATUS [23:22]		SW11.STATUS [21:20]		SW10.STATUS [19:18]		SW9.STATUS [17:16]		SW1.STATUS [1:0]		SW7.STATUS [13:12]	
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R

SWx.STATUS[2] – Reflects status of internal SWx flags:

- “00” – On/Off OK
- “01” – On/Off Failed
- “10” – Open
- “11” – Short

The bit is cleared upon a successful readout over PXN (clear by read bit).

Table 45. REGISTER 0x10

Register 0x10													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x10	Name	PWM_CNT_OVF	GND_LOSS	VBB_LOW	OTP_ZAP_UV	CAP_UV	HWR	–	DIMERR	DIMWARN	GSWERR	TSD	TW
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x10	Name	PXN_CRC_ERR_CNT[23:20]				OTP_CR_C_FAIL_BANK_0	OTP_CR_C_FAIL_BANK_2	TIME_OUT	PXN_SYNC_ERR	PXN_FRAME_ERR	PXN_LOCAL_COM_M_ERR	PXN_GLOBAL_COM_M_ERR	MAPEN_A_STATUS
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R

TW – Thermal warning flag. <TW> flag is set when Tj above Thermal Warning Threshold is detected. The bit is cleared upon a successful readout over PXN.

TSD – Thermal shutdown flag. <TSD> flag is set when Tj above Thermal Shutdown Threshold is detected. When the flag is set, the device enters FAIL status mode and the switches are switched OFF.

The bit is cleared upon a successful readout over PXN.

GSWERR – Global switch error indicator. The bit is set high under the following conditions:

- at least one switch is shorted or
- at least one switch is open or
- ext. capacitor charging has failed

The bit is cleared upon a successful readout over PXN.

DIMWARN – Dimming warning indicator. The bit is set high in case of an overlapping switch OFF sequences are detected.

The bit is cleared upon a successful readout over PXN (clear by read bit).

DIMERR – Dimming error indicator. The bit is set high in case an overlapping switch ON sequences are detected. When the flag is set, the device enters the FAIL status mode. The bit is cleared upon a successful readout over PXN (clear by read bit).

HWR – HWR flag is set after POR. The bit is cleared upon a successful readout over PXN (clear by read bit).

CAP_UV – Status bit indicating that charging process of external capacitor failed. When this bit is set, the <GSWERR> flag is set to ‘1’ and the device enters the FAIL status mode and the switches are switched ON. The bit is cleared upon a successful readout over PXN (clear by read bit).

OTP_ZAP_UV – The bit is set if the battery voltage during OTP zapping is lower than 15 V. The bit is cleared upon a successful readout over PXN (clear by read bit).

VBB_LOW – The bit is set if the battery voltage is lower than 4.5 V. The bit is cleared upon a successful readout over PXN (clear by read bit).

GND_LOSS – The GND loss comparator detects Ground connection loss. The TST1 pin is used as reference ground. The TST1 pin is connected to ground on application PCB level (see Table 12. GND Loss Detection). The bit is cleared upon a successful readout over PXN (clear by read bit).

PWM_CNT_OVF – When PWM counter overflows, flag is set to ‘1’. It should be used to detect that PWM control (PWM counter) is running/functional. The bit is cleared upon a successful readout over PXN (clear by read bit).

MAPENA_STATUS – MAPENA request status. It corresponds to the state of internal MAPENA register bit.

PXN_GLOBAL_COMM_ERR – PXN global communication error. The flag is set in case of global communication error is detected and not disabled by “Global bit error detection DIS” in OTP bank (see Table 24. OTP Bank). For correct functionality it is mandatory to ensure an automatic echo from Tx to Rx. The PXN global communication error is detected under the following circumstance: REPEATER–SLAVE node, where the TX” differs from RX”. The TX” is monitored by PXN node in REPEATER–SLAVE mode through TX_ECHO_UART input.

The bit is cleared upon a successful readout over PXN (clear by read bit).

PXN_LOCAL_COMM_ERR – PXN local communication error. The flag is set in case of local communication error detected. The PXN local communication error is detected under the following circumstance: SLAVE node, where the TX’ differs from RX’. The bit is cleared upon a successful readout over PXN (clear by read bit).

PXN_FRAME_ERR – PXN frame error. The flag is set whenever one of the following errors is detected:

- Parity error – the parity calculated over bits 0 .. 6 of either PID1 or PID2 is not matching the corresponding parity bit P
- CRC error – the CRC calculated over PID1, PID2 and all data bytes do not match the received CRC
- STOP BIT error – ‘0’ received at the expected stop bit position

The bit is cleared upon a successful readout over PXN (clear by read bit).

PXN_SYNC_ERR – The PXN synchronization error is detected in case the duration of eight synchronization field Tbits, when counted in 20 MHz domain is outside the following limits:

Table 46. PXN SYNCHRONIZATION ERROR LIMITS

PXN Communication Speed [kb/s]	Eight_Tbits_min	Eight_Tbits_max
125	1098	1484
250	549	742
500	274	371
1000	140	181

The bit is cleared upon a successful readout over PXN (clear by read bit).

TIMEOUT – The PXN timeout error is detected in case neither MAPENA nor MAPENA_DIR is activated within a TIMEOUT period. The timeout error detection starts when the NORMAL mode is entered and either MAPENA or MAPENA_DIR is activated for the first time. When the flag is set, the device enters the FAIL status mode and LED’s are switched ON/OFF following the OTP memory bits 8–11 in Table 24. OTP Bank. The reported OPMODE status is NORMAL_DIRECT mode. The bit is cleared upon a successful readout over PXN (clear by read bit).

OTP_CRC_FAIL_BANK2 – CRC over mirrored OTP BANK2 bit failure. The bit is cleared upon a successful readout over PXN (clear by read bit).

OTP_CRC_FAIL_BANK0 – CRC over mirrored OTP BANK0 bit failure. The device should not operate when this bit is set. The bit is cleared upon a successful readout over PXN (clear by read bit).

PXN_CRC_ERR_CNT[3:0] – 4-bit PXN frame CRC error counter. The counter is incremented each time a CRC error is detected on incoming PXN frame. Once a maximum number of errors is reached, the counter remains clamped at value of 15. The bit is cleared upon a successful write ‘1’ to register < CRC_CLR>.

Table 47. REGISTER 0x11

Register 0x11													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x11	Name	VDD_RES[11:8]				TEMP_RES[7:0]							
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x11	Name	ADCX_RES[23:16]								VDD_RES[15:12]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R

TEMP_RES[7:0] – the TEMP measured value (read only bits).

VDD_RES[7:0] – the last VDD measured value (read only bits).

ADCX_RES[7:0] – the last value measured at selected ADC measurement channel (read only bits). In case the <ADC_SEL[1:0]> bits are set to “11” the returned measured value is always “00000000”.

Table 48. REGISTER 0x12

Register 0x12													
Address		Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x12	Name	VLED_RES[11:8]				VBB_RES[7:0]							
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R
Address		Bit 23	Bit 22	Bit 21	Bit 20	Bit 19	Bit 18	Bit 17	Bit 16	Bit 15	Bit 14	Bit 13	Bit 12
0x12	Name	TSD_CODE[23:16]								VLED_RES[15:12]			
	Reset	0	0	0	0	0	0	0	0	0	0	0	0
	Access	R	R	R	R	R	R	R	R	R	R	R	R

VBB_RES[7:0] – the last VBB measured value (read only bits).

VLED_RES[7:0] – the last VLED measured value (read only bits). Because of the internal ESD structure, the minimum measured VLED voltage is VDD minus forward diode voltage Vf.

TSD_CODE[7:0] – thermal shutdown threshold (read only bits).

CONFIGURATION FRAMES

Table 49. CF0 – SLAVE IDENTIFICATION

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x00				
2	DATA1	DEV_ID[7:0]							
3	DATA2	ANA_ID[7:0]							
4	DATA3	DIG_ID[7:0]							
5	DATA4	0	0	0	DLT_W[4:0]				
6	DATA5	DLT_X_AXS[7:0]							
7	DATA6	DLT_Y_AXS[7:0]							
8	CRC	CRC[7:0]							

DATA1

DEV_ID[7:0] device ID, L343 = 1

DATA2

ANA_ID[7:0] analog version ID, 35

DATA3

DIG_ID[7:0] digital version ID, 1

DATA4

DLT_W[4:0]

DATA5

DLT_X_AXS[7:0]

DATA6

DLT_Y_AXS[7:0]

The DLT_W, DLT_X_AXS and DLT_Y_AXS parameters describe the wafer specifications.

Table 50. CF1 – EEPROM WRITE DATA

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x01				
2	DATA1	B	0	0	0	0	EESA[2:0]		
3	DATA2	EEBA[7:0]							
4	DATA3	BYTE0 @(EEBA+0)[7:0]							
5	DATA4	BYTE1 @(EEBA+1)[7:0]							
6	DATA5	BYTE2 @(EEBA+2)[7:0]							
7	DATA6	BYTE3 @(EEBA+3)[7:0]							
8	DATA7	BYTE4 @(EEBA+4)[7:0]							
9	DATA8	BYTE5 @(EEBA+5)[7:0]							
10	DATA9	BYTE6 @(EEBA+6)[7:0]							
11	DATA10	BYTE7 @(EEBA+7)[7:0]							
12	CRC	CRC[7:0]							

DATA1

B broadcast bit:
1 – broadcast frame
0 – addressed frame

EESA[2:0] EEPROM slave address

DATA2

EEBA[7:0] EEPROM byte address

DATA3 – DATA10

BYTEx[7:0] bytes to be written

Table 51. CF2 – EEPROM REQUEST DATA

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x02				
2	DATA1	B	0	0	0	0	EESA[2:0]		
3	DATA2	EEBA[7:0]							
4	CRC	CRC[7:0]							

DATA1

B broadcast bit:
1 – broadcast frame
0 – addressed frame

EESA[2:0] EEPROM slave address

DATA2

EEBA[7:0] EEPROM byte address

Table 52. CF3 – EEPROM READ DATA

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x03				
2	DATA1	WP	EES[1:0]		0	0	EESA[2:0]		
3	DATA2	EEBA[7:0]							
4	DATA3	BYTE0 @(EEBA+0)[7:0]							
5	DATA4	BYTE1 @(EEBA+1)[7:0]							
6	DATA5	BYTE2 @(EEBA+2)[7:0]							
7	DATA6	BYTE3 @(EEBA+3)[7:0]							
8	DATA7	BYTE4 @(EEBA+4)[7:0]							
9	DATA8	BYTE5 @(EEBA+5)[7:0]							
10	DATA9	BYTE6 @(EEBA+6)[7:0]							
11	DATA10	BYTE7 @(EEBA+7)[7:0]							
12	CRC	CRC[7:0]							

DATA1

EESA[2:0] EEPROM slave address

EES[1:0]

EEPROM status:
0x0 – EEPROM busy, access denied
0x1 – EEPROM busy, data transfer ongoing
0x2 – EEPROM ready, data transfer completed
0x3 – EEPROM ready, data transfer failed

WP

EEPROM write protect status

DATA2

EEBA[7:0] EEPROM byte address

DATA3 – DATA10

BYTE_x[7:0] read bytes

Table 53. CF4 – ENABLE/DISABLE AUTO-ADDRESSING MODE

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x04				
2	DATA1	B	0	0	0	0	0	0	AAC
3	CRC	CRC[7:0]							

DATA1

B broadcast bit: 1 – broadcast frame
0 – addressed frame

AAC

auto-addressing control bit: 1 – enable
0 – disable

The CF4 frame is accepted in the OTP_CONFIG mode only.

Table 54. CF5 – ASSIGN ADDRESS

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x05				
2	DATA1	B	0	0	AA_ADR[4:0]				
3	DATA2	AA_THR[7:0]							
4	CRC	CRC[7:0]							

DATA1
B broadcast bit:
1 – broadcast frame
0 – addressed frame

AA_ADR[4:0] 5-bit address to assign

DATA2
AA_THR[7:0] 8-bit auto-address threshold value

The CF5 frame is accepted in AUTO_ADDR mode only.

Table 55. CF6 – OP MODE STATUS

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x06				
2	DATA1	PXN_FRAME_CNT[3:0]				OPMODE[3:0]			
3	CRC	CRC[7:0]							

DATA1
PXN_FRAME_CNT[4:0] PXN frame counter – 4-bit counter which is incremented each time any valid PXN frame is processed. The counter overflows to 0 upon the increment.

OPMODE[3:0] OP mode status:
 0x0 – not valid
 0x1 – OTP config mode
 0x2 – auto-addressing mode
 0x4 – normal direct mode
 0x5 – normal pwm mode
 0x6 – normal fail-safe otp mode
 0x7 – normal fail-safe open mode
 0xC – NO_CRC direct mode
 0xD – NO_CRC pwm mode
 0xE – fail safe OTP mode
 0xF – fail safe OPEN mode

Table 56. CF7 – SLAVE/REPEATER–SLAVE PXN MODE SELECTION

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x07				
2	DATA1	B	0	0	0	0	0	0	PMC
3	CRC	CRC[7:0]							

DATA1

B broadcast bit:
 1 – broadcast frame
 0 – addressed frame

PMC PXN mode control bit:
 1 – repeater–slave mode
 0 – slave mode

Overwrites device mode loaded from the OTP memory.

Table 57. CF8 – READ PXN MODE STATUS

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x08				
2	DATA1	0	0	0	0	0	0	0	PMS
3	CRC	CRC[7:0]							

DATA1

PMS PXN mode status bit:
 1 – repeater–slave mode
 0 – slave mode

Table 58. CF9 – WRITE DATA TO OTP

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x09				
2	DATA1	B	0	0	0	0	0	OTPBS[1:0]	
3	DATA2	BYTE0 @(OTPBA+0)							
4	DATA3	BYTE1 @(OTPBA+1)							
5	DATA4	BYTE2 @(OTPBA+2)							
6	DATA5	BYTE3 @(OTPBA+3)							
7	DATA6	0x00							
8	DATA7	CRC[7:0]							

DATA1

B broadcast bit:
1 – broadcast frame
0 – addressed frame

OTPBS[1:0] 2-bit OTP bank selection
0x2 – custom OTP bank
0x0, 0x1, 0x3 – no bank selected

DATA2 – DATA6

BYTE_x[7:0] bytes to be written

The CF9 frame is accepted in OTP_CONFIG mode only.

Table 59. CF10 – REQUEST DATA FROM OTP

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x0A				
2	DATA1	B	0	0	0	0	0	OTPBS[1:0]	
3	CRC	CRC[7:0]							

DATA1

B broadcast bit:
1 – broadcast frame
0 – addressed frame

OTPBS[1:0] 2-bit OTP bank selection:
0x2 – custom OTP bank
0x0, 0x1, 0x3 – no bank selected

Table 60. CF11 – READ DATA FROM OTP

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x0B				
2	DATA1	LOCKB	OTPS[1:0]		0	0	OTPBS[2:0]		
3	DATA2	BYTE0 @(OTPBA+0)							
4	DATA3	BYTE1 @(OTPBA+1)							
5	DATA4	BYTE2 @(OTPBA+2)							
6	DATA5	BYTE3 @(OTPBA+3)							
7	DATA6	BYTE4 @(OTPBA+4)							
8	CRC	CRC							

DATA1

LOCKB custom OTP bank general lock status bit:
 1 – locked
 0 – unlocked, further OTP programming possible

OTPS[1:0] 2-bit OTP status:
 0x0 – idle, no data transfer ongoing
 0x1 – busy, data transfer ongoing
 0x2 – ready, data transfer OK
 0x3 – ready, data transfer failed

OTPBS[1:0] 2-bit OTP bank selection:
 0x2 – custom OTP bank
 0x0 or
 0x1 or
 0x3 – no bank selected

DATA2 – DATA6

BYTEx[7:0] read bytes

Table 61. CF12 – COMMUNICATION SPEED

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x0C				
2	DATA1	B	0	0	0	0	0	CSPEED[1:0]	
3	CRC	CRC[7:0]							

DATA1

B broadcast bit:
 1 – broadcast frame
 0 – addressed frame

CSPEED[1:0] communication speed:
 0x0 – 125 kbps
 0x1 – 250 kbps (default)
 0x2 – 500 kbps
 0x3 – 1000 kbps

Overwrites device communication speed loaded from the OTP memory.

Table 62. CF13 – SWITCH TO NORMAL MODE

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x0D				
2	DATA1	B	0	0	0	0	0	0	NMD
3	CRC	CRC[7:0]							

DATA1

B broadcast bit:
 1 – broadcast frame
 0 – addressed frame

NMD request to enter NORMAL mode from OTP_CONFIG mode:
 1 – go to NO_CRC or normal mode; according to the OTP bank 2 lock bit
 0 – no effect

The CF13 frame is accepted in OTP_CONFIG mode only.

Table 63. CF14 – RESET SYSTEM

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x0E				
2	DATA1	B	0	0	0	0	0	0	SWRST
3	CRC	CRC							

DATA1

B broadcast bit:
 1 – broadcast frame
 0 – addressed frame

SWRST software reset bit:
 1 – perform software reset
 0 – no effect

The CF14 frame is accepted in NORMAL mode only.

Table 64. CF15 – TRIGGER MAPENA

Byte	Name	Contents							
		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	PID1	P	1	1	SA[4:0]				
1	PID2	P	0	0	CSID[4:0] = 0x0F				
2	DATA1	B	0	0	0	WDT_SEL[1:0]		0	MAPENA
3	CRC	CRC[7:0]							

DATA1

B broadcast bit:
 1 – broadcast frame
 0 – addressed frame

MAPENA trigger MAPENA
 1 – perform MAPENA
 0 – no effect

WDT_SEL[1:0] watchdog timeout selector; valid only with <MAPENA> = '1'
 0x0 – 250 ms
 0x1 – 150 ms
 0x2 – 100 ms
 0x3 – 60 ms

THERMAL WARNING, ERROR DETECTION AND DIAGNOSTICS FEEDBACK

The NCV78343 offers a wide range of device-integrated diagnostic features. Their description follows.

Thermal Warning and Shutdown

The junction temperature can be calculated from ADC code as follows:

$$T_j = \frac{ADC_{CODE} + 184}{TSD_{CODE} + 184} \cdot (TSD_{TEMPERATURE} + 273) - 273 \quad (\text{eq. 2})$$

The <TSD_CODE> is trimmed in production to 170°C (TSD_TEMPERATURE). Typical value of <TSD_CODE> is 186 and exact trimmed value can be read from Register 0x12. The <TW> status bit is set high in case the measurement result is greater than or equal to TW_CODE[7:0] value. The <TSD> status bit is set high in case the measurement result is greater than or equal to TSD_CODE[7:0].

The TW and TSD status bits are cleared in case the <TEMP_RES> register value is smaller than both TSD_CODE and TW_CODE values and TW and TSD status bits are successfully read out over PXN.

Overlapping switch ON/OFF events

Overlapping switch ON events is forbidden, the NCV78343 needs time slot between two switch ON events (see Table 43). Superior system has to ensure that overlapping switch ON events do not appear in patterns.

When overlapping switch ON events are despite this invoked, the NCV78343 incorporates protective feature, in which the <DIMERR> error is raised.

When overlapping switch OFF events occur, the <DIMWARN> status bit is set and processing of this pattern continues. However, it has to be taken into account, that overlapping switch OFF events can lead to large fluctuations of LED string voltage.

Pixel Switches diagnostic

Embedded diagnostic covers a wide range of possible failure situations on switches. Each switch contains two comparators – short comparator and over voltage comparator. With the help of these two comparators a several fail situations can be detected and distinguished on each switch individually.

Overall status of switches errors are indicated by the Global Switch Error <GSWERR> status flag. Status of individual switches (whether a switch is ON or OFF) can be read in the Switch Status <SWx.STATUS> flag. The failure state of the individual switches (Short, Open) is indicated by corresponding status in <SW.STATUS[2:0]> register.

When the device operates in the PWM mode and a short is present, the device reports alternating On/Off Failed and SHORT status according to set ON/OFF dimming duration (see details in Figure 13).

Dedicated OPEN clear request

When the OPEN state or overvoltage is detected, the switch is automatically closed and the status is shown in the SWx.STATUS register 0x0F. Further attempts to control this switch using PWM or DIRECT mode don't have any effect. The switch is released upon a successful OPEN clear request frame.

In direct mode, the released switch is updated upon a successful write to register 0x00 with service "00". In PWM mode, the released switch is updated according to the ON/OFF values.

PWM_CNT_OVF

When PWM Counter overflows, <PWM_CNT_OVF> flag is set to '1'. It is clear by read flag. It should be used to detect that PWM control (PWM counter) is running/functional.

CAP_UV

The <CAP_UV> status bit indicates that charging process of external capacitor failed. When this bit is set, the <GSWERR> flag is set to '1' and all switches are switched OFF. It is clear by read flag and the switches are set according to last successful write to the register 0x00.

VBB_LOW

The <VBB_LOW> status bit indicates low battery voltage < 4.5 V (see Table 4). The bit is not detected when the battery voltage level is lower than 4.5 V during the start-up sequence.

Power-on Reset

After a power-on a flag <HWR> in the register is set.

APPLICATION RELATED INFORMATION

PXN CRC Code Example

```
// Byte reverse
uint8_t l343_byte_reverse(uint8_t b)
{
    b = (b & 0xF0) >> 4 | (b & 0x0F) << 4;
    b = (b & 0xCC) >> 2 | (b & 0x33) << 2;
    b = (b & 0xAA) >> 1 | (b & 0x55) << 1;
    return b;
}

// Calculate the PXN CRC
uint8_t l343_pxn_crc(uint8_t *data_bits, uint8_t length)
{
    uint8_t CRC8 = 0;

    for (uint8_t a=0; a<length; ++a)
    {
        CRC8 = CRC8 ^ data_bits[a];

        for (uint8_t i=0; i<8; ++i)
        {
            if (CRC8 & 0x80)
            {
                CRC8 = ((0x7F & CRC8) * 2) ^ 0x07;
            }
            else
            {
                CRC8 = CRC8 * 2;
            }
        }
    }

    return CRC8;
}

Function call example:
int main(void)
{
    // MAPENA as broadcast; SEED, PID1, PID2, DATA0; CRC = 0x28
    uint8_t data_bits[] = {0xFF, 0x61, 0x8F, 0x81};

    // Write to REG00; SEED, PID1, PID2, DATA0, DATA1, DATA2; CRC = 0x14
    // uint8_t data_bits[] = {0xFF, 0xA1, 0x80, 0x55, 0x05, 0x00};

    // Get the number of bytes
    int length = sizeof(data_bits)/sizeof(data_bits[0]);

    // Invert the input
    for (uint8_t a = 0; a<length; ++a) data_bits[a] = l343_byte_reverse(data_bits[a]);

    // Get the result
    uint8_t result = l343_pxn_crc(data_bits, length);
}
```

Parity Bit Calculation

```
uint8_t l343_parity(uint8_t val)
{
    bool Par;

    Par = ((val>>0)&1) ^ ((val>>1)&1) ^ ((val>>2)&1) ^ ((val>>3)&1) ^ ((val>>4)&1) ^
    ((val>>5)&1) ^ ((val>>6)&1);
    Par = (Par ^ 1) & 1;

    return (uint8_t)Par;
}
```

Go to NMD Frame (CF13)

```
int32_t l343_normal_mode(uint8_t addr)
{
    uint8_t p;
    uint8_t pdata[5];

    // SYNC
    pdata[0] = 0x55;

    // PID1
    uint8_t PID1 = 0;
    PID1 = 3<<5 | addr;
    p = l343_parity(PID1);
    pdata[1] = (p << 7) | PID1;

    // PID2
    uint8_t PID2 = 0;
    PID2 = 0x0D;
    p = l343_parity(PID2);
    pdata[2] = (p << 7) | PID2;

    // DATA bytes
    pdata[3] = 1;          // NMD

    uint8_t pdata_crc[4];

    // Invert the input
    for (uint8_t a = 0; a<4; ++a) pdata_crc[a] = l343_byte_reverse(pdata[a]);

    // Calculate the CRC
    pdata_crc[0] = 0xFF;
    uint8_t crc = l343_pxn_crc(pdata_crc, 4);

    pdata[4] = crc;

    // Send data
    return serial_pxn_set_data(pdata, 5);
}
```

OTP Write Code Example

```
typedef struct OTP_t
{
    unsigned lb: 1;           // Lock bit
    unsigned na_lb: 1;        // Node Address Lock bit
    unsigned na: 5;           // Note Address
    unsigned fss_lb: 1;       // Fail Safe State Lock bit
    unsigned fss: 4;          // Fail Safe State of LEDs
    unsigned pxn_lb: 1;       // PXN Lock bit
    unsigned mode: 1;         // Mode
    unsigned cs: 2;           // Communication speed
    unsigned gbed: 1;         // Global bit error detection
    unsigned m_lvds_off: 1;   // M-LVDS off
    unsigned uart_off: 1;     // UART off
    unsigned ee_lb: 1;        // EEPROM lock bit
    unsigned crc: 7;          // CRC
} OTP_t;

// Calculate the OTP CRC
uint8_t l343_otp_crc(uint8_t *data_bits, char length)
{
    uint8_t CRC7 = 0;

    for (uint8_t a = 0; a<length; ++a)
    {
        CRC7 = CRC7 ^ data_bits[a];

        for (uint8_t i = 0; i < 8; ++i)
        {
            if (CRC7 & 0x80)
            {
                CRC7 = ((0x7F & CRC7) * 2) ^ (0x37 * 2);
            }
            else
            {
                CRC7 = (CRC7 * 2);
            }
        }
    }

    return (CRC7 / 2);
}

int32_t l343_otp_zapping(OTP_t otp)
{
    uint8_t data_bits[4];
    data_bits[0] = 0x07;
    data_bits[1] = 0x0F<<4 | otp.ee_lb<<3 | otp.uart_off<<2 | otp.m_lvds_off<<1 | otp.gbed<<0;
    data_bits[2] = otp.cs<<6 | otp.mode<<5 | otp.pxn_lb<<4 | otp.fss<<0;
    data_bits[3] = otp.fss_lb<<7 | otp.na<<2 | otp.na_lb<<1 | otp.lb<<0;

    // Get the number of bytes
    int length = sizeof(data_bits)/sizeof(data_bits[0]);

    // Get the result
    otp.crc = l343_otp_crc(data_bits, length);

    // PXN OTP write frame
```

```

uint8_t p;
uint8_t pdata[10];

// SYNC
pdata[0] = 0x55;

// PID1
uint8_t PID1 = 0;
PID1 = 3<<5 | otp.na;
p = 1343_parity(PID1);
pdata[1] = (p << 7) | PID1;

// PID2
uint8_t PID2 = 0;
PID2 = 0x09;
p = 1343_parity(PID2);
pdata[2] = (p << 7) | PID2;

// DATA bytes
const uint32_t data = otp.lb |
    otp.na_lb << 1 |
    otp.na << 2 |
    otp.fss_lb << 7 |
    otp.fss << 8 |
    otp.pxn_lb << 12 |
    otp.mode << 13 |
    otp.cs << 14 |
    otp.gbed << 16 |
    otp.m_lvds_off << 17 |
    otp.uart_off << 18 |
    otp.ee_lb << 19 |
    otp.crc << 20;

pdata[3] = 0x02;
pdata[4] = ((data >> 0) & 0xFF);
pdata[5] = ((data >> 8) & 0xFF);
pdata[6] = ((data >> 16) & 0xFF);
pdata[7] = ((data >> 24) & 0xFF);
pdata[8] = 0x00;

uint8_t pdata_crc[9];

// Invert the input
for (uint8_t a = 0; a<9; ++a) pdata_crc[a] = 1343_byte_reverse(pdata[a]);

// Calculate the CRC
pdata_crc[0] = 0xFF;
uint8_t crc = 1343_pxn_crc(pdata_crc, 9);

pdata[9] = crc;

// Send data
return serial_pxn_set_data(pdata, 10);
}

```

Function call example:

```
int main(void)
```

```
{
```

```
    // Zap the OTP
```

```
    //          lb  nalb  na  fss_lb fss  pxn_lb  mode  CS  GBED  lOff  uOff  eeLb  crc
```

```
    OTP_t otp = {0x01, 0x01, 0x01, 0x01, 0x0F, 0x01, 0x00, 0x01, 0x00, 0x00, 0x01, 0x00, 0x00};
```

```
    l343_otp_zapping(otp);
```

```
}
```

Auto-addressing (AA) Process

This example is valid for two devices, where the first one is in repeater–slave mode and the second one is in slave mode. The first device is connected to the MCU via UART and the second device is connected to the first device via M–LVDS.

The slave address (SA) for the first device will be set to ‘1’ and the SA for the second device will be set to ‘7’.

The AA process combines benefits of current source and connected LED string. The application does not need any additional wires. When the device is connected to the LED string and the current source for this LED string is enabled, the voltage drop across the LED string will occur. The LED string voltage VLED is measured by the device. Thus the address may be assigned to specific device.

In general, the MCU sends a broadcast frame CF4 (see Table 53) to all node devices and the second broadcast frame CF5 (see Table 54) with the VLED threshold and new device address as parameters. Doing this, all devices on the node will be in AA mode and only the device with VLED higher than set threshold will assign new address.

For this example, the LED string voltage is 33 V (127 ADC code). The auto-addressing threshold will be set to 80.

1. Disable all buck outputs, thus the LEDs are not shining.
2. Enable buck output 1, so the LED string for the device 1 is shining.
3. Enable AA mode by sending CF4 as broadcast (B=1; AAC=1); see Table 53.

4. Assign the address for the device 1 by sending CF5 as broadcast (B=1; ADR=1; THR=80); see Table 54.
5. Disable buck output 1, so the LED string for the device 1 is not shining.
6. Disable the AA mode for the first device SA=1 by sending CF4 (B=0; AAC=0); see Table 53.
7. Force the normal mode for the first device SA=1 by sending CF13 (NMD=1); see Table 62.
8. Set the first device as repeater–slave SA=1 by sending CF7 (PMC=1); see Table 56.
9. Enable buck output 2, so the LED string for the device 2 is shining.
10. Enable AA mode by sending CF4 as broadcast (B=1; AAC=1); see Table 53.
11. Assign the address for the device 2 by sending CF5 as broadcast (B=1; ADR=7; THR=80); see Table 54.
12. Disable buck output 2, so the LED string for the device 2 is not shining.
13. Disable the AA mode for the second device SA=7 by sending CF4 (B=0; AAC=0); see Table 53.
14. Force the normal mode for the second device SA=7 by sending CF13 (NMD=1); see Table 62.

For multiple devices connected to the first one via M–LVDS, please repeat steps 9–14 with different ADR, THR, SA values.

Dimming Control Patterns

The simplest way to set ON, OFF and TR values is to keep the ON time at fixed value and calculate the OFF value according to the required duty. The following table (see Table 65) shows an example of fixed ON values. It is calculated for each switch 0–11 and for each DIMFREQ value 0–20.

The values are calculated as follows:

$$TR_SLOT = 12 \times T_SW_ON_SEQ \quad (\text{eq. 3})$$

$$SW_SLOT = \frac{1024 - TR_SLOT}{T_SW_ON_SEQ} \quad (\text{eq. 4})$$

$$ON_VALUE = TR_SLOT + SWITCH \times SW_SLOT \times T_SW_ON_SEQ \quad (\text{eq. 5})$$

where

SWITCH goes from 0 to 11

T_SW_ON_SEQ is number of ticks required for one switch ON sequence

Table 65. DIMMING CONTROL PATTERN ON VALUES

DIMFREQ	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
T_SW_ON_SEQ	2	2	2	2	2	2	2	2	4	4	4	4	4	4	4	4	8	8	8	8	16
TR_SLOT	24	24	24	24	24	24	24	24	48	48	48	48	48	48	48	48	96	96	96	96	192
SW_SLOT	41	41	41	41	41	41	41	41	20	20	20	20	20	20	20	20	9	9	9	9	4
SWITCH	ON values																				
0	24	24	24	24	24	24	24	24	48	48	48	48	48	48	48	48	96	96	96	96	192
1	106	106	106	106	106	106	106	106	128	128	128	128	128	128	128	128	168	168	168	168	256
2	188	188	188	188	188	188	188	188	208	208	208	208	208	208	208	208	240	240	240	240	320
3	270	270	270	270	270	270	270	270	288	288	288	288	288	288	288	288	312	312	312	312	384
4	352	352	352	352	352	352	352	352	368	368	368	368	368	368	368	368	384	384	384	384	448
5	434	434	434	434	434	434	434	434	448	448	448	448	448	448	448	448	456	456	456	456	512
6	516	516	516	516	516	516	516	516	528	528	528	528	528	528	528	528	528	528	528	528	576
7	598	598	598	598	598	598	598	598	608	608	608	608	608	608	608	608	600	600	600	600	640
8	680	680	680	680	680	680	680	680	688	688	688	688	688	688	688	688	672	672	672	672	704
9	762	762	762	762	762	762	762	762	768	768	768	768	768	768	768	768	744	744	744	744	768
10	844	844	844	844	844	844	844	844	848	848	848	848	848	848	848	848	816	816	816	816	832
11	926	926	926	926	926	926	926	926	928	928	928	928	928	928	928	928	888	888	888	888	896

Dimming Algorithm Code

```
// Variables for dimming algorithm
static const uint8_t TR_SLOT[] = {24, 24, 24, 24, 24, 24, 24, 24, 24, 48, 48, 48, 48, 48, 48, 48, 96, 96, 96, 96, 192};
static const uint8_t SW_SLOT[] = {41, 41, 41, 41, 41, 41, 41, 41, 20, 20, 20, 20, 20, 20, 20, 20, 9, 9, 9, 9, 4};
static const uint8_t T_SW_SEQ_RATIO[] = {2, 2, 2, 2, 2, 2, 2, 2, 2, 4, 4, 4, 4, 4, 4, 4, 8, 8, 8, 8, 16};

bool l343_dimming(uint8_t dimfreq, uint16_t *DCs, uint16_t *onRet, uint16_t *offRet, uint16_t *trRet)
{
    const uint8_t LEDcount = 12;

    for (uint8_t i = 0; i < LEDcount; ++i)
    {
        if (DCs[i] > 1023) DCs[i] = 1023;

        // Input is brightness of LEDs, but the ON/OFF values are for switches, thus inverted
        DCs[i] = 1023 - DCs[i];
        trRet[i] = i; // set TR 0..11

        if (DCs[i] == 1023) // Fully ON 100% DUTY cycle, thus set dedicated values 0 and 1023
        {
            onRet[i] = 0;
            offRet[i] = 1023;
        }
        else
        {
            onRet[i] = TR_SLOT[dimfreq] + i*SW_SLOT[dimfreq]*T_SW_SEQ_RATIO[dimfreq];

            if (DCs[i] == 0) // Fully OFF 0% DUTY cycle, where OFF is equal to ON value
            {
                offRet[i] = onRet[i];
            }
            else
            {
                offRet[i] = (onRet[i] + DCs[i]) % 1024;
            }
        }

        DCs[i] = 1023 - DCs[i]; // Invert back to keep the values as they were
    }

    return true;
}

// Number of devices in a cluster
#define DEVICES X // [1; 31]

#define REGISTERS 12 // number of registers to be written
```

```

// LED brightness; the length should be DEVICES*REGISTERS; or two-dimensional array might be used
uint16_t DC[DEVICES*REGISTERS];          // values are in a range of [0; 1023]
// uint16_t DC[DEVICES][REGISTERS];

// Drimfreq for each device
uint8_t dimfreq[DEVICES]

// Function call example:
// Send ON, OFF, TR values to all devices
void l343_send(uint16_t *DC)
{
    for (uint8_t dev = 0; dev<DEVICES; ++dev)
    {
        uint16_t ON[REGISTERS];
        uint16_t OFF[REGISTERS];
        uint16_t TR[REGISTERS];

        // Calculate the dimming values
        l343_dimming(dimfreq[dev], DC+REGISTERS*dev, ON, OFF, TR);
//        l343_dimming(dimfreq[dev], DC[dev], ON, OFF, TR);          // when two-dimensional array is used

        // Fill the registers values
        uint32_t reg[REGISTERS];

        for (uint8_t r = 0; r<REGISTERS; ++r)
        {
            reg[r] = (ON[r]<<14) | (OFF[r]<<4) | (TR[r]&0xF);
        }

        // 36 bytes need to be sent in 3 frames
        uint8_t RBA[3] = {1, 5, 9};

        for (uint8_t r = 0; r<3; ++r)
        {
            uint8_t p;
            uint8_t pdata[16];

            // SYNC
            pdata[0] = 0x55;

            // PID1
            uint8_t PID1 = 0;
            PID1 = 1<<5 | SA[dev];
            p = l343_parity(PID1);
            pdata[1] = p<<7 | 1<<5 | PID1;

            // PID2
            uint8_t PID2 = 0;
            PID2 = 3<<5 | (RBA[r]);
            p = l343_parity(PID2);

```

```

pdata[2] = p<<7 | PID2;

// DATA bytes
pdata[3] = reg[r*4+0]&0xFF;
pdata[4] = (reg[r*4+0]>>8)&0xFF;
pdata[5] = (reg[r*4+0]>>16)&0xFF;

pdata[6] = reg[r*4+1]&0xFF;
pdata[7] = (reg[r*4+1]>>8)&0xFF;
pdata[8] = (reg[r*4+1]>>16)&0xFF;

pdata[9] = reg[r*4+2]&0xFF;
pdata[10] = (reg[r*4+2]>>8)&0xFF;
pdata[11] = (reg[r*4+2]>>16)&0xFF;

pdata[12] = reg[r*4+3]&0xFF;
pdata[13] = (reg[r*4+3]>>8)&0xFF;
pdata[14] = (reg[r*4+3]>>16)&0xFF;

uint8_t pdata_crc[16];

// Invert the input
for (unsigned char a = 0; a<15; ++a) pdata_crc[a] = 1343_byte_reverse(pdata[a]);

// Calculate the CRC
pdata_crc[0] = 0xFF;
char crc = 1343_pxn_crc(pdata_crc, 15);

pdata[15] = crc;

// Send data
serial_pxn_set_data(2, pdata, 16, 11);
}
}
}

```

OTP Read Data Interpretation

The table below shows an example of OTP read data and its interpretation in the OTP table (see Table 24).

[illegible]

BANK (2)	
OTP_lock_bit	☒
OTP_node_adrlckbit	☒
OTP_node_addr[4:0]	1
Fail_safe_st_lck_bit	☒
Fail_safe_st_LED[3:0]	0
PXN_lock_bit	☒
Mode	☒
Communi_speed[1:0]	1
Globa_bit_err_DIS	☐
LVDS_OFF	☐
UART_OFF	☐
EEPROM_lock_bit	☐
CRC1[6:0]	83

Figure 18. Read OTP Data Interpretation

Return to Normal Operation after Fail-Safe Mode

Once a device detects one of the following status bits: TSD or CAP_UV or VBB_LOW or DIMERR or TIMEOUT, the device enters the fail-safe mode (see Operating Modes section). To leave this mode, the superior system shall read out the REG 0x10 (see Table 45) and/or handle the error if required.

The device enters FAIL-SAFE OTP mode when DIMERR or TIMEOUT appears. When TIMEOUT is set and this mode is entered, switches are set according to the OTP memory values. If the OTP memory is not zapped, the switches are switched OFF. Once the error status bit is cleared, the switches remain unchanged. When DIMERR is

set and this mode is entered, switches operation is unaffected.

The device enters FAIL-SAFE OPEN mode when TSD or CAP_UV or VBB_LOW appears. In this mode, the switches are automatically switched OFF. Once the error status bit is cleared, the switches are set according to the values in REG 0x00 (see Table 27).

The TSD/CAP_UV/VBB_LOW group of bits (hardware fail) have higher priority to the DIMERR/TIMEOUT group of bits (application fail). When these two groups appear at the same time, the device enters the FAIL-SAFE OPEN mode.

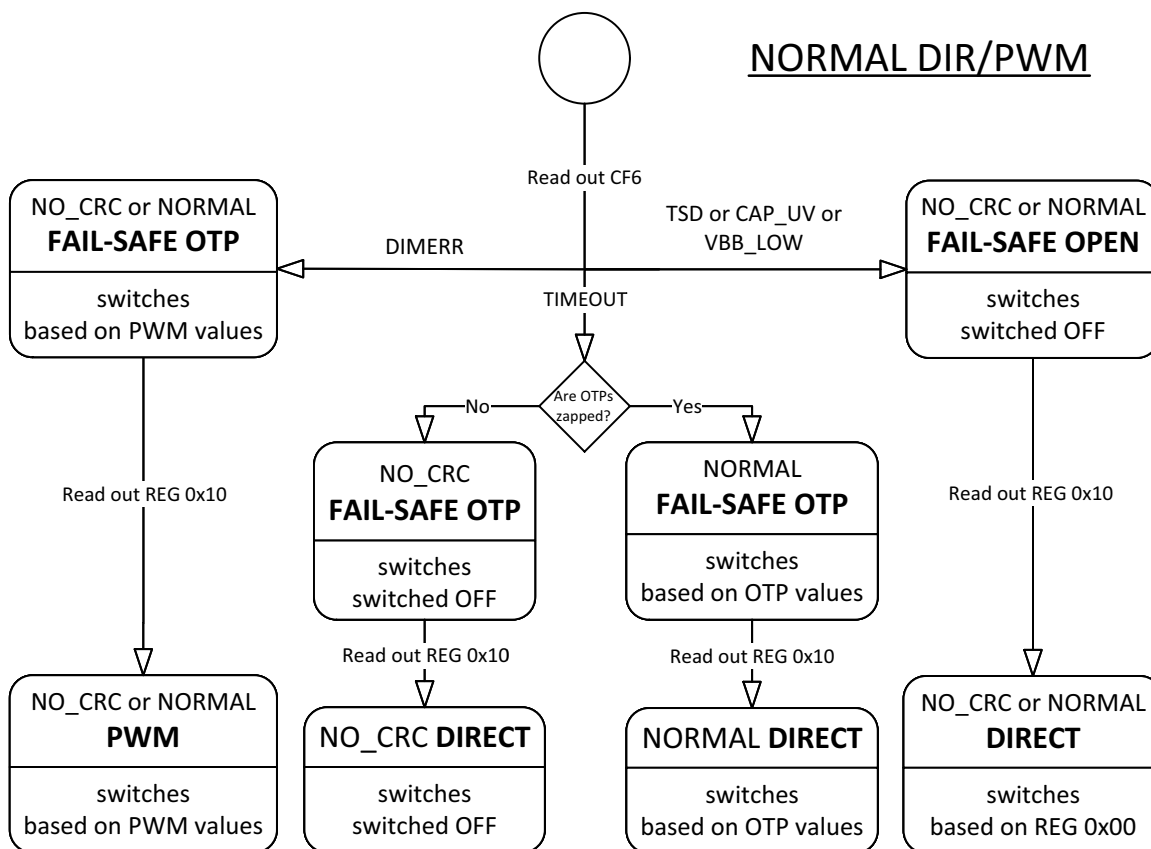


Figure 19. FAIL-SAFE Modes

Power Up and Down Sequences

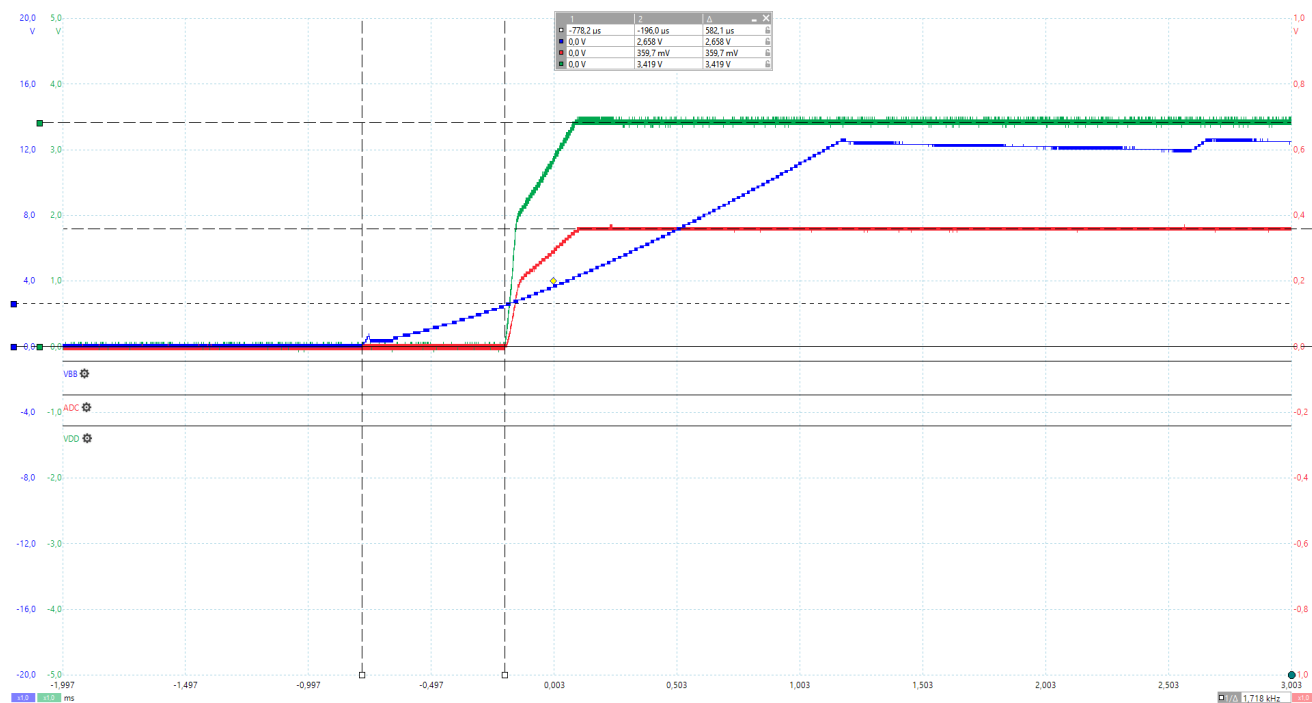


Figure 20. Power-up Sequence with 10 nF at ADC2/ADR Pin

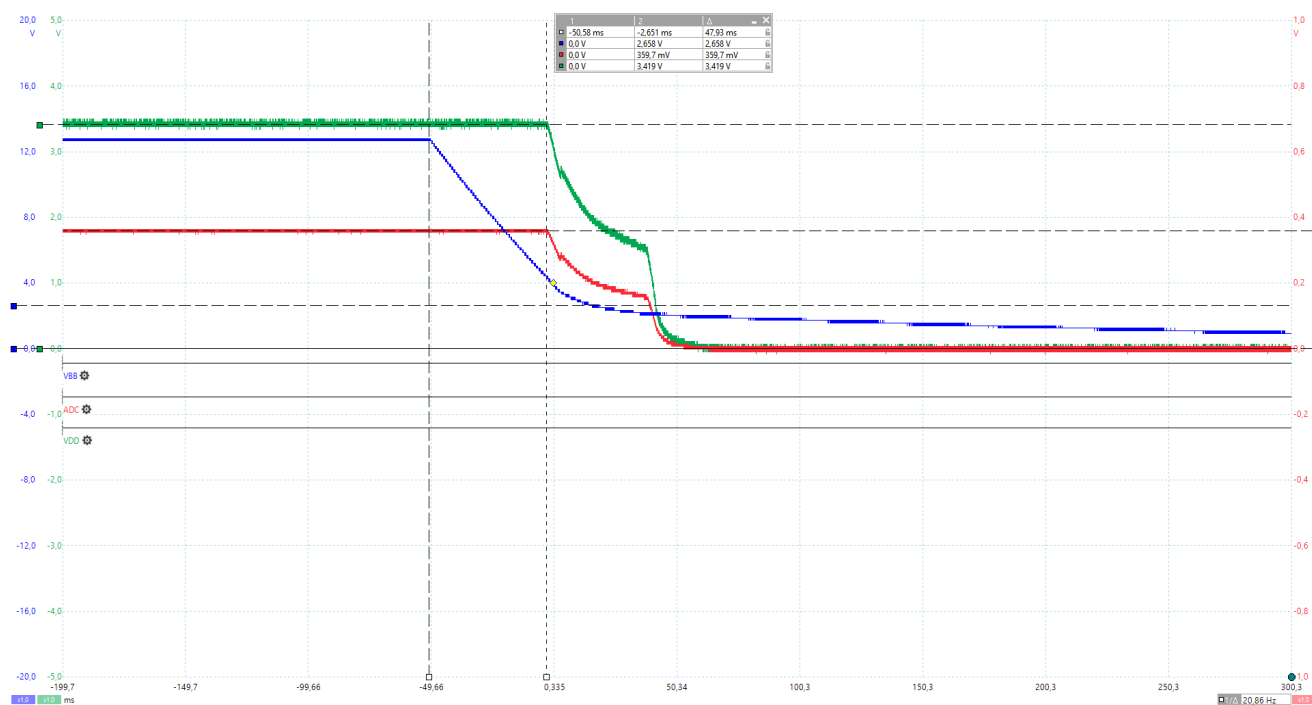


Figure 21. Power-down Sequence with 10 nF at ADC2/ADR Pin

Example Flow Chart Diagram for the Normal Operational Mode

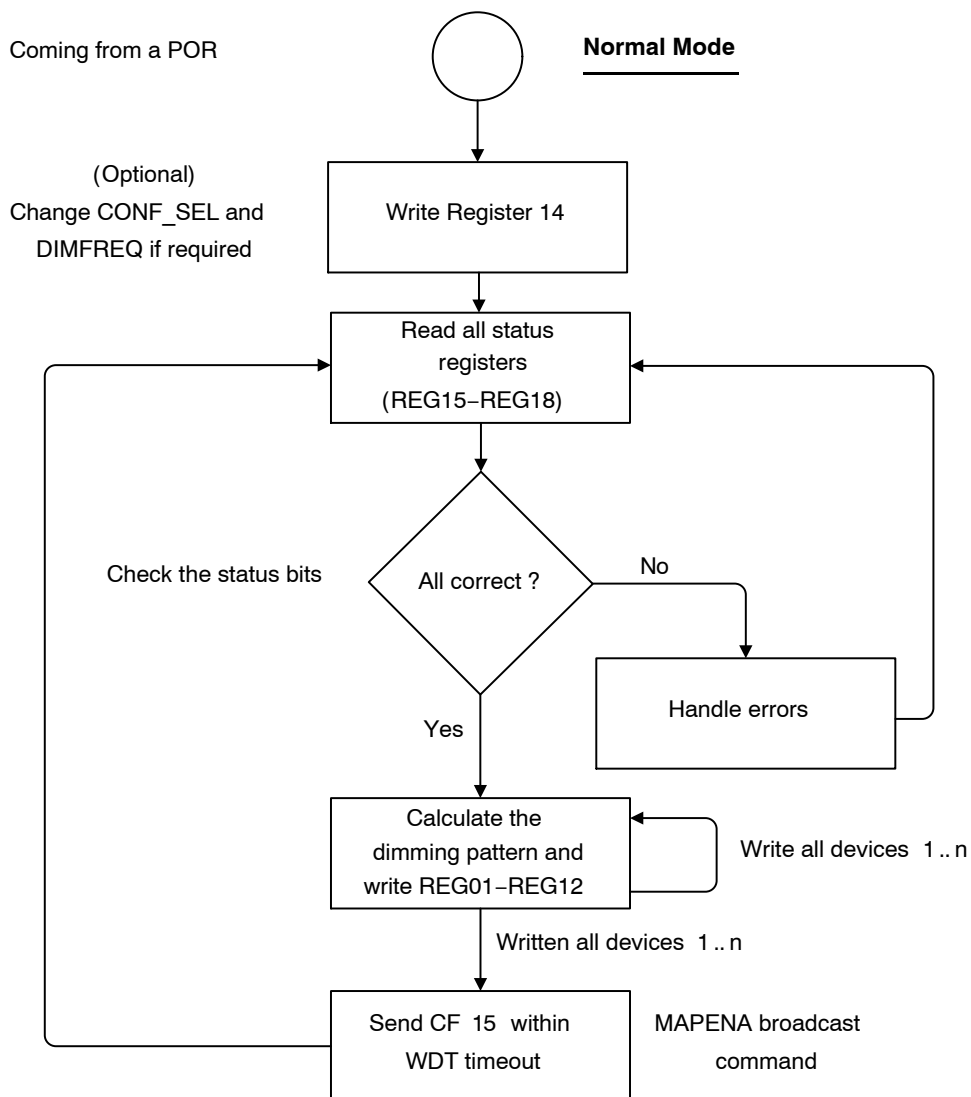


Figure 22. Normal Operation Mode Flow Chart Diagram

The main loop should consist of checking all status bits and handling them if necessary. Set a refresh rate for common headlamp lighting functions (e.g. High beam) as well as fulfill watchdog timeout. The MCU should calculate

all ON/OFF/TR values within this time and send broadcast MAPENA (see Table 64) frame once the values are sent into devices.

Flow Chart after POR

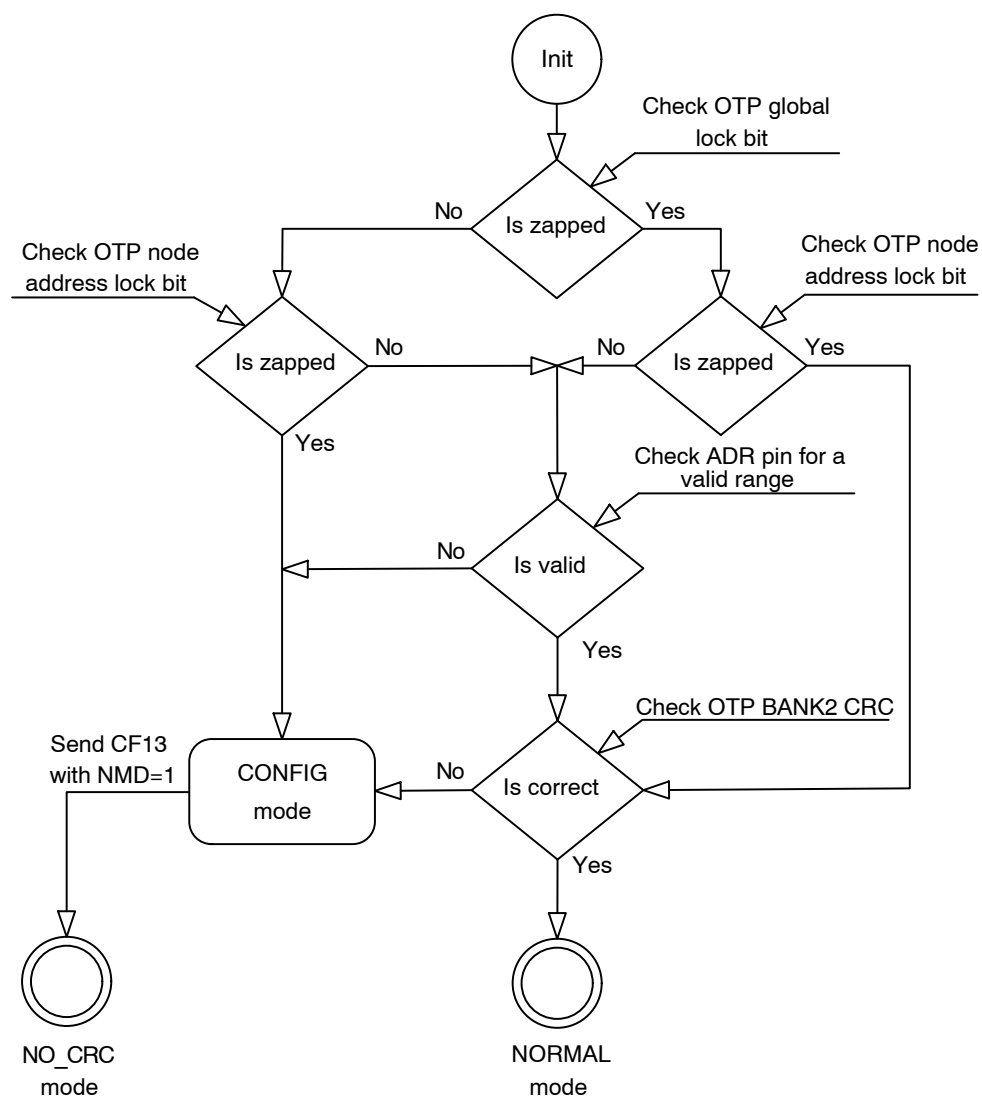


Figure 23. Flow Chart Diagram after POR

The diagram above is an automatic flow after each POR. A device might end up in either CONFIG or NORMAL or NO_CRC mode according to zapped OTP bits.

A new device will end up in CONFIG mode, because OTP bits are not zapped. An address is set by either

Auto-addressing process or by Multi-level addressing using a voltage divider or zapped in the OTP memory.

Multi-level Addressing Procedure with Long Time Delay at ADC2/ADR Pin

The following flow chart is valid for the repeater-slave and slaves cluster, where the repeater-slave communicates

through CAN-PHY layer and slaves are connected via local M-LVDS bus.

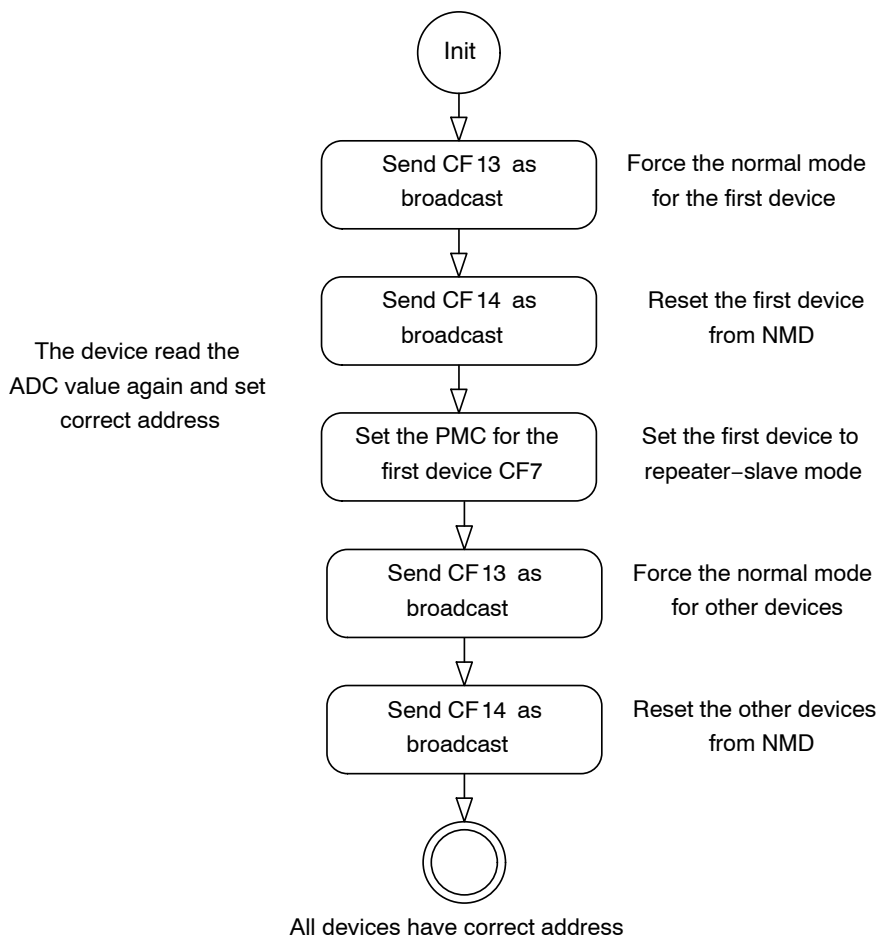


Figure 24. Multi-level Addressing Procedure with Long Time Delay at ADC2/ADR Pin

EEPROM Write and Read Operations

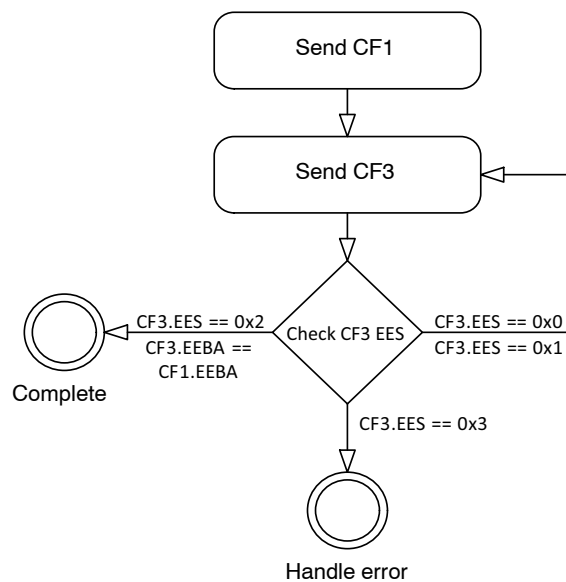


Figure 25. EEPROM Write Operation

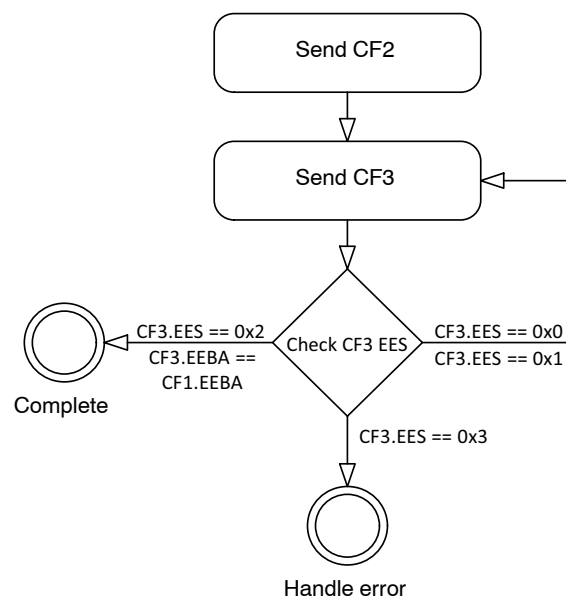
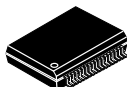


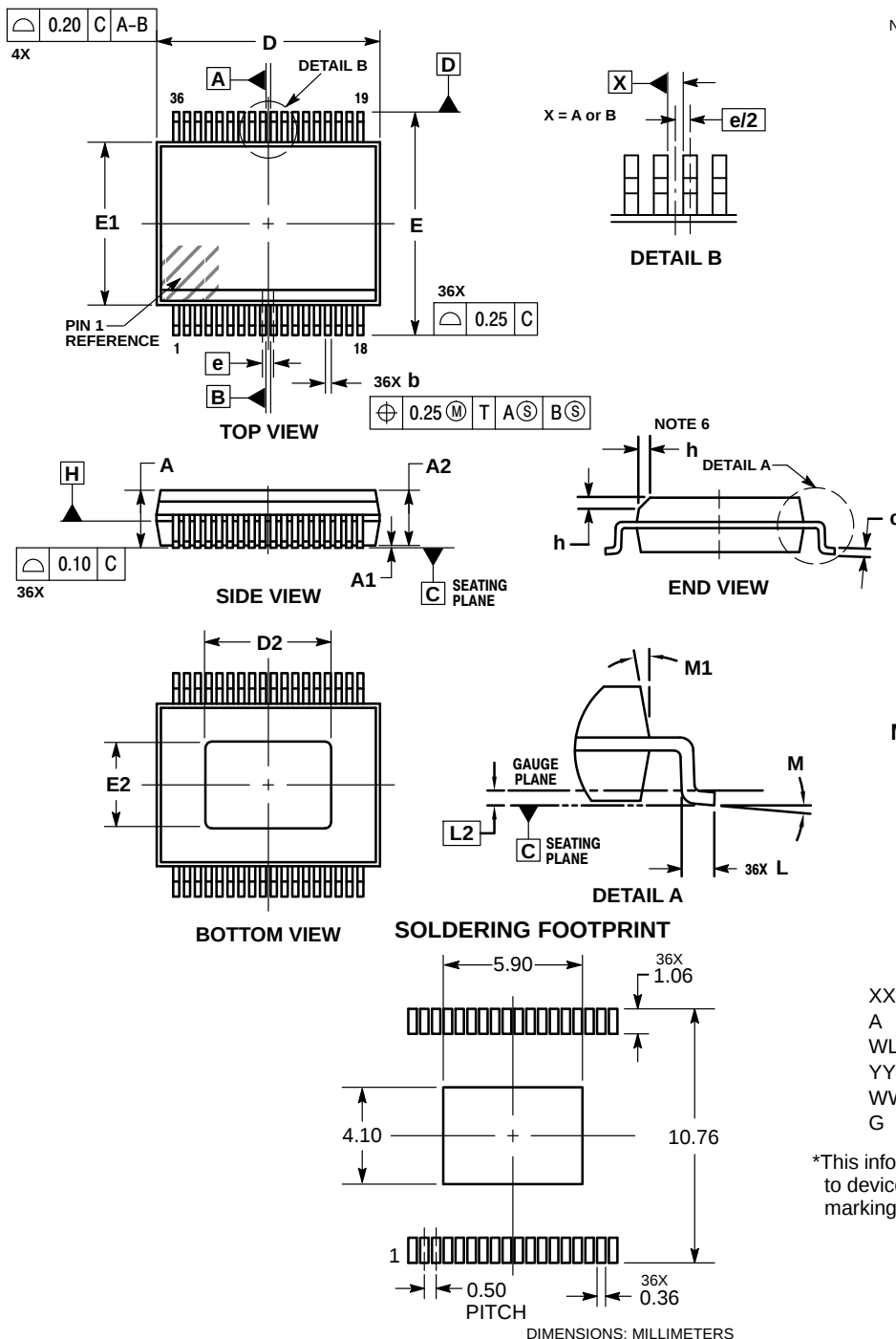
Figure 26. EEPROM Read Operation



SCALE 1:1

SSOP36 EP
CASE 940AB
ISSUE A

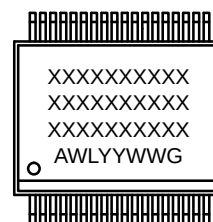
DATE 19 JAN 2016



NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.13 TOTAL IN EXCESS OF THE b DIMENSION AT MMC.
4. DIMENSION b SHALL BE MEASURED BETWEEN 0.10 AND 0.25 FROM THE TIP.
5. DIMENSIONS D AND E1 DO NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. DIMENSIONS D AND E1 SHALL BE DETERMINED AT DATUM H.
6. THIS CHAMFER FEATURE IS OPTIONAL. IF IT IS NOT PRESENT, A PIN ONE IDENTIFIER MUST BE LOCATED WITHIN THE INDICATED AREA.

MILLIMETERS		
DIM	MIN	MAX
A	---	2.65
A1	---	0.10
A2	2.15	2.60
b	0.18	0.30
c	0.23	0.32
D	10.30 BSC	
D2	5.70	5.90
E	10.30 BSC	
E1	7.50 BSC	
E2	3.90	4.10
e	0.50 BSC	
h	0.25	0.75
L	0.50	0.90
L2	0.25 BSC	
M	0°	8°
M1	5°	15°

GENERIC MARKING DIAGRAM*


XXXX = Specific Device Code
A = Assembly Location
WL = Wafer Lot
YY = Year
WW = Work Week
G = Pb-Free Package

*This information is generic. Please refer to device data sheet for actual part marking.

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